BALANCE AND DISTORTION TRADEOFFS

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SpeakerBuilder

THE LOUDSPEAKER JOURNAL

de LANGE:
TIME AND CROSSOVER FILTERS

CARRINGTON:
MODS FOR RADIO SHACK'S SL METER

WASLO:
MANAGING MIKES WITH IMP

PIKE:
BUILD A SIXTEEN Hz WOOFER

AUGSPURGER:
BANDPASS & BIRDHOUSE

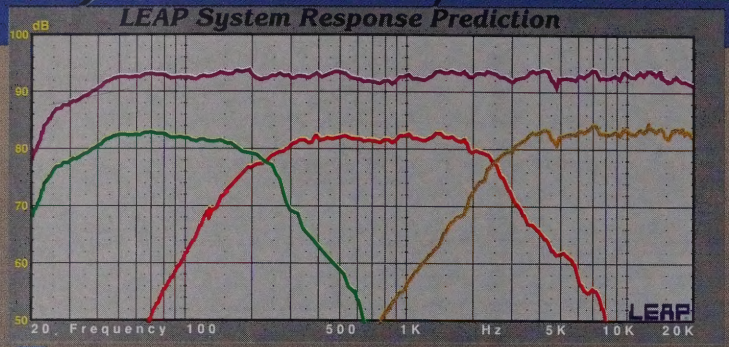
WAYLAND:
BISCUITS FOR BETTER JOINTS

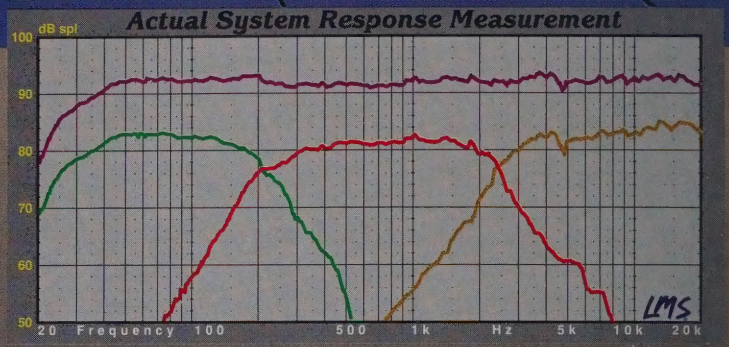
DOES U.S. PATENT LAW NEED REVISION?

New!
LEAP
Version 4.5

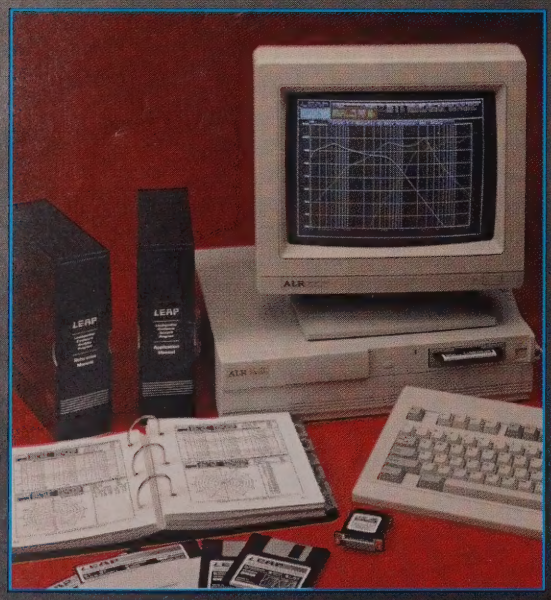
LEAP

Precision Development Tools for Precision Loudspeaker Designs





The Art and Science... of loudspeaker system development today has become more complex than ever before. Competition is tough, and to compete each design must perform to the best of its ability, and make the most out of every dollar's worth of transducer cost. The simple approach of choosing a combination of seemingly appropriate transducers coupled with ordinary networks and filters, has given way to a painstaking process of meticulously blending selected transducers in combination with carefully devised and matched crossover designs.



LEAP (Loudspeaker Enclosure Analysis Program) is a complete full range analysis package which provides virtually *all* of the tools necessary to develop precision loudspeaker systems, for today's demanding audio markets. Whether your applications are consumer audio, car stereo, professional audio, or custom esoteric marvels, LEAP provides the power, flexibility, and accuracy to investigate every possible design permutation. The open architecture and broad spectrum of features provided will dramatically reduce your development time, while improving the quality of the final result... and demonstrates why LEAP has become the #1 choice of professional loudspeaker designers world-wide!

Advanced System Analysis Features

- ✔ Use simulation or imported actual measured SPL/Z data.
- ✔ 5-Way crossover system modeling, and more.
- ✔ Time offset between transducers.
- ✔ Active or Passive based crossovers.
- ✔ Hilbert-Bode transform for deriving phase.
- ✔ 22 Passive components per xover section.
- ✔ 16 Active filter blocks per xover section.
- ✔ Passive Network Optimizer for single/system response.
- ✔ Active Filter Optimizer for single/system response.
- ✔ Frequency ranges from 1Hz to 100kHz.

Printer/Output Formats

When you wish to produce a hardcopy output of your finished designs and graphical data, LEAP supports a large number of printer standards, and even supports numerous desktop publishing graphic formats in both black & white and color! Portrait/Landscape orientations in any custom size and aspect ratio are user controllable.

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- ✔ PostScript EPS/TIF- B&W,Color
- ✔ AI-Adobe Illustrator B&W,Color
- ✔ NEC 24 Dot Matrix
- ✔ TOSHIBA 24 Dot Mat
- ✔ HP DeskJet 500C
- ✔ PostScript Printers
- ✔ DXF AutoCAD
- ✔ TIFF, BMP, PCX Plots

Extensive Documentation

The two volume manual set comprises almost 1,000 pages of documentation which thoroughly covers the operation of the program- and provides numerous examples of how to maximize your use and understanding of the program's many features. The Reference Manual describes all graphs, menus, commands, and their operation. This manual explains the unique and special non-linear speaker and port models, as well as proper use of the optimizers, importing data, and the many other utilities. The Application Manual provides many exciting examples showing how to use the powerful features of the system in a combined manner to perform both simple and complex design tasks. Both novice and experienced users alike will find this information invaluable for exploiting the full power of the system. Additional information is also provided on loudspeaker measurements, design tips, filter calculations, and complete crossover system development for both passive and active based systems.

- ✔ 502 Page Reference Manual
- ✔ 436 Page Application Manual

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Good News

➲ CIRCUIT GUARDIAN

Protect your circuit boards and electronic parts from moisture, static, corrosion, and pollutants with a new nontoxic spray from Circuit Guard International (Ontario, CANADA). Circuit Guard™ protects, without altering resistive, capacitive or inductive components. Contact Circuit Guard International at (905) 509-2647, or (800) 365-5030.

Reader Service #104

■ THE CHAMELEON HF

Galaxy Audio designed the Chameleon to blend into your home audio surroundings. This compact, yet full-range, monitor speaker is compatible with virtually all amps and is available from Galaxy Audio, 625 East Pawnee Avenue, Wichita, KS 67211, (316) 263-2852, FAX (316) 263-0642.

Reader Service #117

■ GENELEC GOES HOME

Genelec, long recognized by audio producers and engineers for its professional studio products, introduces a line of loudspeakers for the custom home theater market. The nine models range from a compact, biamplified, two-way design to a large three-way design with a massive triamplification system. Genelec, distributed by QMI, 25 South Street, Hopkinton, MA 01748, (508) 435-3666, FAX (508) 435-4243.

Reader Service #109

➲ CANON FIRES SALVO

Canon has recently showcased two new speakers, the S-30 and S-35. Designed for use with audio and audio/video systems, including home theater, these models feature curved acoustic mirrors for what Canon claims is a stereo image up to six times larger than a conventional speaker. Canon, One Canon Plaza, Lake Success, NY 11042, (516) 328-5145.

Reader Service #102

➲ STICKY BUSINESS

Hernon React 727 Acrylic Structural Adhesive bonds to a variety of materials, including metals, glass, and ceramics, in many applications. You can apply this easy-to-use, fast-acting thixotropic paste, which you can use with any of three Hernon activators. No curing action takes place until the parts are mated. Hernon Manufacturing, 121 Tech Drive, Sanford, FL 32771, (407) 322-4000, (800) 527-0004, FAX (407) 321-9700.

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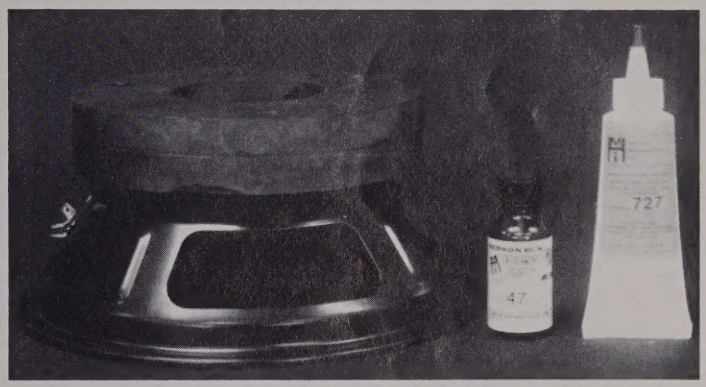
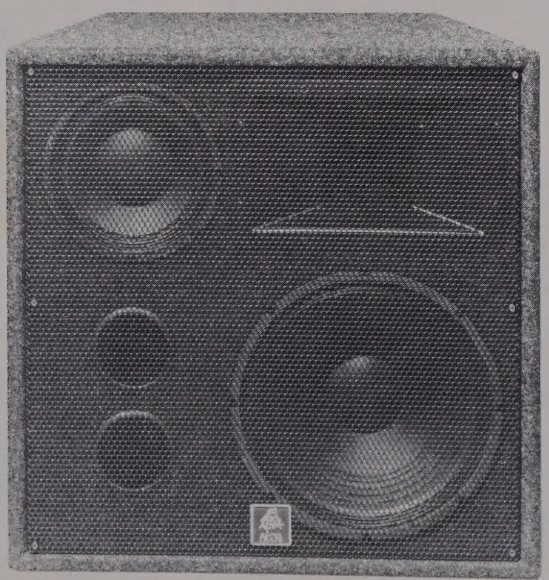


■ THE SIBLINGS' SPEAKER

Clair Brothers introduces the R-4 Series III three-way speaker system, featuring a directional mid-range baffle system. The new design, according to the company, delivers high intelligibility and transparent sound. Clair Brothers Audio Systems, PO Box 396, Lititz, PA 17543, (616) 695-5948, FAX (616) 695-7623.

Reader Service #115



■ LOSING A VOICE

By purchasing ServoDrive design and manufacturing rights, Quantum Sound hopes to change the face of low-frequency audio reproduction. Replacing the voice coil with the servomotor drive increases linear excursion and the overall efficiency of the loudspeaker system, thus enabling the ServoDrive to handle large low-frequency peaks without damage or sonic compromise. Quantum Sound, 3453 Commercial Ave., Northbrook, IL 60062, (616) 695-5948, FAX (616) 695-7623.

Reader Service #114

➲ NO FLIES ON THIS SPEAKER

Bullfrog introduces the PR1510, a three-way trapezoidal loudspeaker system designed for main PA use and full-range instrument applications. The Bullfrog cabinet houses a double-ported 15″ bass speaker and two woofers. Bullfrog, 1503 Prairie Ave., South Bend, IN 46613, (616) 695- 5948, FAX (616) 695-7623.

Reader Service #105

Speaker Builder (US ISSN 0199-7920) is published every six weeks (eight times a year), at **$32** per year, **$58** for two years; Canada add $8 per year; overseas rates **$50** one year, **$90** two years; by Audio Amateur Publications, Inc., Edward T. Dell, Jr., President, at 305 Union Street, PO Box 494, Peterborough, NH 03458-0494. Second-class postage paid at Peterborough, NH and an additional mailing office.

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← ***Reader Service #36***

Good News

B&W LAUNCHES NAUTILUS

Featuring a radical snail-like design, the Nautilus speaker boasts a high price tag ($35,000/pair) as well as high-quality music reproduction. The manufacturer claims its smooth contours produce the most diffraction-free baffle surfaces yet devised. B&W, 54 Concord St., N. Reading, MA 01864, (508) 664-2870, (800) 370-3740, FAX (508) 664-4109.

Reader Service #118

■ FREE FOR THE ASKING

Simply by calling (800) 833-8575, you can receive a free brochure, "The Power of the Platform," which gives prospective users a detailed overview of Techron's TEF 20 Systems for sound system test and analysis. For more information about TEF applications, contact Techron, PO Box 1000, Elkhart, IN 46515, (219) 294-8300, FAX (219) 294-8329.

Reader Service #116

■ PCB DESIGN AID

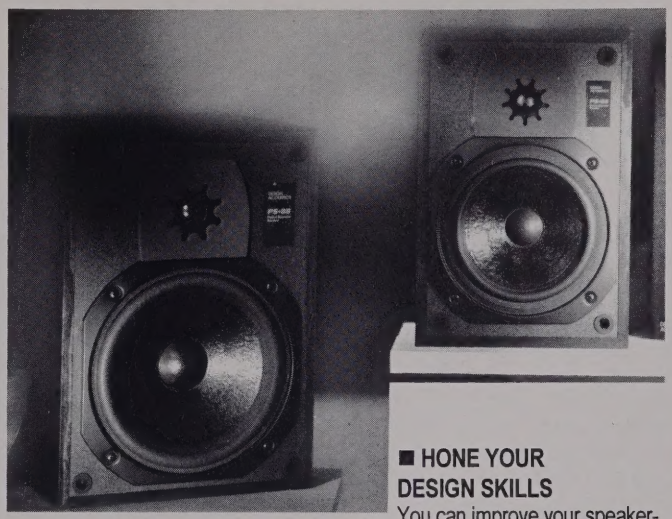
Baas Electronics of The Netherlands has released version 4.92 of its popular, low-cost PCB design package Layo1, which speeds up off-line design rule checking, component dragging, and printing check plots. Compatible with many schematic capture packages, such as OrCAD, Layo1 is available from Old Colony Sound Lab, PO Box 243, Peterborough, NH 03458, (603) 924-6526, FAX (603) 924-9467.

Reader Service #112

■ CABLE SANS DISTORTION

Alpha-Core offers a new speaker cable, the Goertz MI (matched impedance), which eliminates distortion—regardless of run length—between amplifier and speaker. The compact cable design features low inductance and high capacitance for a difference anyone can hear in a comparison test, according to Alpha-Core, 915 Pembroke St., Bridgeport, CT 06608, (203) 335-6805, FAX (203) 384-8120.

Reader Service #110



BOOKSHELF SPEAKERS

Design Acoustics presents two new full-range video-shielded loudspeakers, the PS66 and PS88, featuring a compact size and high performance well-suited for home theater applications. Design Acoustics, 1920 Enterprise Parkway, Twinsburg, OH 44087, (216) 425-8222, FAX (216) 425-9339.

Reader Service #106

■ A POWERHOUSE AMP

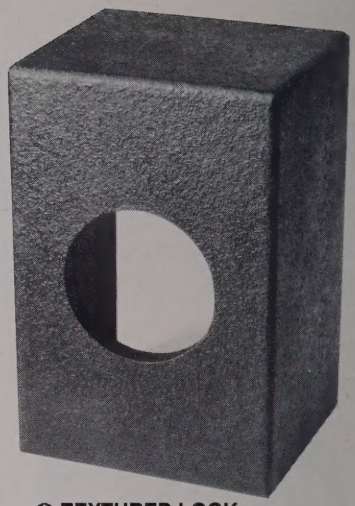
The PA9 monoblock power amplifier is available from Sentec America (Route 9, Stockport, NY 12172, (518) 828-8490). This Class A/B amplifier, developed with an optimized signal path which the manufacturer claims produces the most coloration-free signal possible, is compatible with a variety of speakers.

Reader Service #108

■ A SOUND TEST

The Sheffield/Coustic Test and Demonstration Disc is an audio reference compact disc to help you test and fine-tune both home and mobile audio systems. You can evaluate system sonic performance and accuracy and make appropriate adjustments with the specially designed test tracks. Coustic, 4260 Charter St., Vernon, CA 90058, (213) 582-2832, (800) 227-8879, FAX (213) 582-4326.

Reader Service #113



TEXTURED LOOK

Give your speaker boxes a finished look of leather, stucco, stone or pebble with the Texturelac paint finish from Abilene Research and Development Corp., PO Box 294, Hewlett, NY 11557, (516) 791-6943, FAX (516) 791-6948. You can apply your choice of textured pattern—which dries to a hard, mar-resistant finish within 15–30 minutes—with a conventional spray gun.

Reader Service #101

■ HONE YOUR DESIGN SKILLS

You can improve your speaker-building skills with the Loudspeaker Design Information Series (LDIS) beginning June 1994 and published eight times a year. The course, targeted for beginners and audiophiles, teaches the latest in speaker-building technology and sound reproduction techniques. For more information, contact Technologie MDB, 5445 de Lorimier, Suite 108, Montreal, Quebec H2H 2C3, (514) 526-8851.

Reader Service #107

TWEETER ON TOP

The B&W Matrix HTM is designed for several home theater applications: as a center channel, left or right front monitor, or as a surround speaker. The cabinet design, which incorporates B&W's zero magnetic field technology, eliminates stray magnetic flux that could induce video-image distortion in monitors/TVs. B & W, 54 Concord St., N. Reading, MA 01864, (508) 664-2870, (800) 370-3740, FAX (508) 664-4109.

Reader Service #111

Reader Service #34 ➡

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Arnie Nudell,** President **Genesis

*"**CFAC Inductor** is a product that have advanced the state-of- the-art, providing me with a product that is the **most sonically pure component** available. They are **simply, the best...."***

Robert Grost,** President **Unity Audio

PRODUCT HIGHLIGHT

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Made in USA.

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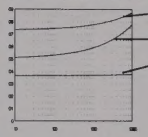

Skin Effect Resistance
Skin Effect will cause the inductor resistive load increase, power loss, non-uniform current density in the conductor, signal amplitude decrease and phase delay. Sound quality will be degraded. The ultra-thin structure (0.003" thick) of the **CFAC INDUCTOR** is specially designed to minimized the Skin Effect.

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The peculiar evil of silencing the expression of an opinion is, that it is robbing the human race; posterity as well as the exisiting generation; those who dissent from the opinion, still more than those who hold it.

--JOHN STUART MILL

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About This Issue

Is existing patent law stifling the entrepreneurial spirit in the US? Editor/Publisher **Ed Dell** ascends his editorial soapbox to address this question and summons audiophiles to become involved.

Dewald de Lange shares with *SB* readers his extensive crossover network time studies and concludes that, in some circumstances, time responses enjoy an advantage over frequency responses when considering crossover filters (p. 10).

C.L.P. Carrington's clever circuit modification shows how to expand the usefulness of a Radio Shack SLM (sound level meter). By adding a switch you can choose between regular operation and an accurate millivolt testmeter to measure speaker response (p. 12).

Similar speakers fed the same inputs will deliver the same performance, right? Not necessarily, according to *SB* author **Don Jenkins**. After putting different speakers through their paces, this super sleuth reveals the results of his investigation in "Acoustic Distortion and Balanced Speakers," p. 14.

Measuring gear doesn't have to be expensive to be accurate...especially if you use **Bill Waslo**'s versatile brainchild, IMP (Impulse response Measurement and Processing). In their most recent adventure, Bill Waslo uses the IMP to computer-correct an inexpensive microphone for unerring speaker results.

After reading our 2/94 article about Matt Federoff's "Birdhouse" design, **George Augspurger** was incredulous. So, Augspurger, a noted loudspeaker authority, decided to look closer at this unusual bandpass design and shares his findings in "More About the Birdhouse Bandpass," p. 22.

Charles Pike always dreamed of achieving "just clean, pure bass that I could feel as well as hear." He recently realized this goal with a 16Hz subwoofer solution detailed on p. 26.

Visiting **Bob Wayland**'s Woodworking World, we discover the secrets of biscuit joinery, an easy way to make strong, eye-pleasing joints for more creative projects.

In the Tools, Tips & Techniques department, airman **Steven Crosby**, recently involved in the airdrop of supplies over Bosnia, shares his interesting design of a dual-volume test box.

Also, contributing editor **Gary Galo** examines a new line of high-performance speaker cables from D.H. Labs.

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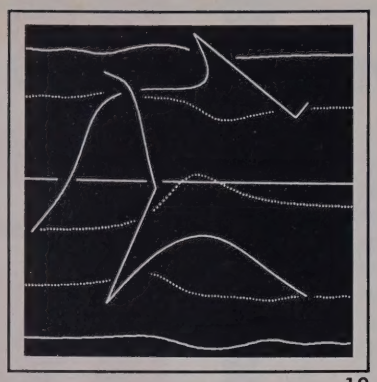
THE LOUDSPEAKER JOURNAL
VOLUME 15 NUMBER 4 MAY 1994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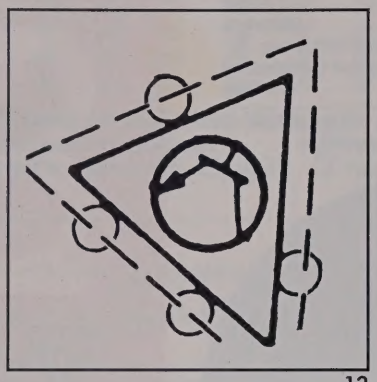
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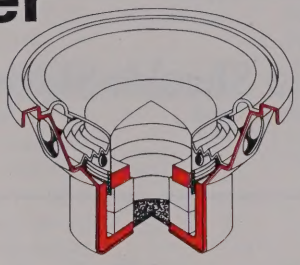
MW 114-S

Neodymium Magnet DPC Cone 4" Woofer



Specification

Overall Dimensions	Ø118mm (4.64") x 58mm(2.29")
Mounting Baffle Hole Diameter	Ø95mm (3.75")
Magnet System	Pot Type, Vented, Neodymium Magnet
Nominal Power Handling (Din)	150W
Transient Power - 10ms	800W
Voice Coil Diameter	54mm (2.125")
Voice Coil Type/Former	Hexatech Aluminium
Frequency Response	55-7000 Hz
FS - Resonant Frequency	65 Hz
Sensitivity 1W/1m	87 dB
Z - Nominal Impedance	8 ohms
RE - DC Resistance	5.6 ohms
LBM - Voice Coil Inductance @ 1kHz	0.47 mH
Magnetic Gap Width	1.25mm (0.050")
HE - Magnetic Gap Height	6mm (0.236")
Voice Coil Height	12mm (0.472")
X - Max. Linear Excursion	3mm
B - Flux Density	0.88T
BL Product (BXL)	6.75
Qms - Mechanical Q Factor	2.32
Qes - Electrical Q Factor	0.36
Q/T - Total Q Factor	0.31
Vas - Equivalent Cas Air Load	3.18 litres (0.113 cu. ft.)
MMS - Moving Mass	7.00gm
CMS	807µm/n
SD - Effective Cone/Dome Area	53cm² (20.86 sq. in.)
Cone/Dome Material	DPC (Damped Polymer Composite)
Nett Weight	0.500 kg

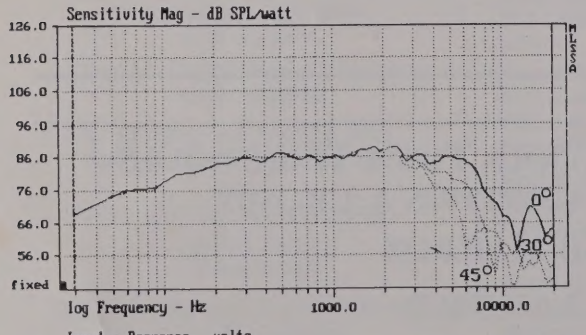
Specifications given are as after at least 45 minutes of high power, low frequency running, or 24 hours normal power operation.

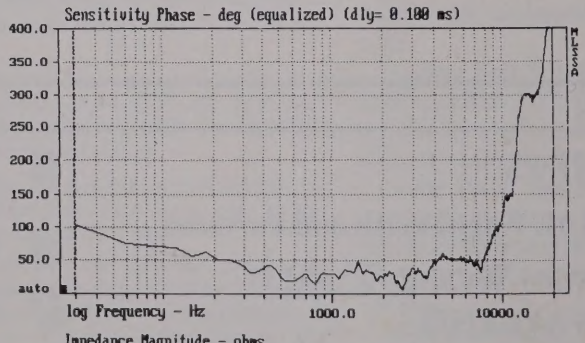
The 114-S is the first of Morel's new generation of woofers, featuring a powerful Neodymium magnet system which provides increased sensitivity, lower Qt and reduced distortion. For a 4" driver it is unique in having a large 54mm (2.125") diameter Hexatech aluminium voice coil.

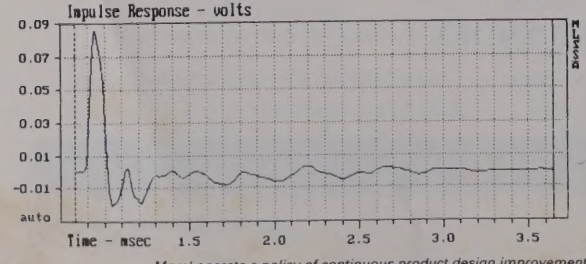
Benefits of this large voice coil diameter include a very high power handling capacity and lack of sound level compression. In addition, it allows the use of a very shallow cone profile. Coupled with the use of Damped Polymer Composite cone material and a rubber surround, this provides excellent dispersion (off-axis response), resistance to cone break-up (even at high sound pressure levels) and lack of colouration.

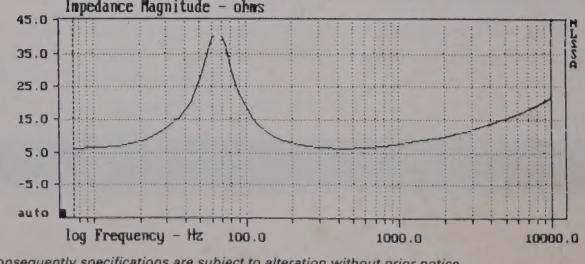
Frequency and phase response are very flat, while the roll-off is very smooth. The MW 114-S may be used either as a bass-mid range in 2-way systems, or as a mid-range in multi-way systems.

The vented magnet system is encased within a steel chassis, which improves efficiency and shields the magnet, virtually eliminating stray magnetic fields. The MW 114-S is ideal not only for high quality hi-fi, but also TV, video and surround-sound applications.









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Editorial

Patent Perils and Purviews

The United States patent statutes have, in my opinion, a serious case of schizophrenia. On the one hand, a patent must fully describe the device itself and also explain clearly how it works. The patent is in public domain, that is, it is not copyright-protected and must be, and is, available to the public. Copies of any US patent are readily available from the patent office for $1.50. (The Commissioner, Office of Patents and Trademarks, Class 179 and 181, Acoustical Devices, Washington, DC 20231.)

On the other hand, any individual who builds the actual patented device, for his own use only, is considered by the law to be infringing the patent. Strictly speaking the builder need not sell a copy or market it, or make one for his neighbor to run afoul of the letter of the present law.

Considerable case law, however, has held that experimental construction of a patented device is not an infringement until it is put to some commercially valuable use. As early as 1850 a case was decided stating that use of a patented machine to satisfy the curiosity of the builder does not constitute infringement.

This matter came up in early April when I received George Augspurger's response to Matt Federoff's article, "The Birdhouse," which appeared in *SB* 2/94. I was more than a little pleased to read that author Augspurger considers Federoff's work "...an original design ideally suited to the author's requirements."

I have included Mr. Augspurger's commentary and findings in his article beginning on page 22. Augspurger also noted the relationship of Federoff's device to a patent held by Bose Corporation (4,549,631), although Federoff's chambers are tuned exactly opposite to those in most commercial systems, including those manufactured by Bose Corporation.

Two weeks later I received notification from a Boston law firm that the Bose Corporation considered the Federoff article to infringe their patent and that by publishing the article I, as publisher, was "actively inducing infringement" of it. Our legal counsel have replied citing one of a long list of case law decisions.

"(p)ublication of information about a patented product or method does not constitute active inducement unless accompanied by other activity such as sale of material capable of infringing use. The patent itself is published and must contain full disclosure of how to make and use the invention." Chisum, Patents, Sec. 17.04(4) (footnotes omitted).

The issue of experimental use of a patented device goes much deeper than being able to build something out of curiosity. If such restrictions are enforced, or are legally actionable, then they inevitably have a chilling effect on innovation, technical exploration, and discovery of further device variations. A patent may not fully explore all that is possible with a specific, legally protected device. Federoff's Birdhouse is dubbed "unique" by one of the most respected authorities on loudspeaker patents in the United States. Is the purpose of a patent to prevent further exploration of the possibilities of a formula or an invention?

The United States has, for over a decade, been falling far behind both Germany and Japan in the number of individuals who apply for and hold patents. One reason is that both countries actively encourage and sometimes fund the inventive efforts of individuals. In the US most of the patents are granted to corporate owners whose R&D produce the discoveries.

At this time a Senate Committee is actively considering an overhaul of the patent law. This might be a good time for amateurs of all stripes to make known their views on experimental use of patented devices. If you agree with me that a clarification of this issue is an important matter, write to: Senator DiConcini, Chairman, Senate Judiciary Subcommittee on Patents, Copyrights, and Trademarks, Washington DC 20510, (202) 224-8178.

A special notice concerning the Federoff article appears in a box below. I commend it to your attention.

The patent mentioned in that notice, 4,549,631, covers forms of multiple port, two-chamber loudspeaker systems. If you'd like a copy of it, send us a 9 x 12 self-addressed envelope with 50¢ US postage (or three International Postal Coupons) and we'll supply one. I have consulted with a number of experts in the loudspeaker field about this patent and some are convinced that others may have published such designs before the Bose patent was granted on October 29, 1985. Our attorneys tell us that if such prior art exists it may be presented to the Patent Office with a request for a reexamination of the originality of the claims.

Speaker builders will certainly hold, as I do, that any patent holder is entitled to the full benefits of his or her discovery. But the patent system was founded not only to protect the rights of the inventor, but to make the benefits of the invention available to all citizens. This patent use issue has hung, like a sword of Damocles, over the heads of any amateur who loves to explore technologies. I believe its revision is in the best interests not only of amateurs like ourselves, but of America's economic health and well being, as well.—E.T.D.

NOTICE

Bose Corporation of Framingham, Massachusetts, have notified Audio Amateur Publications, Inc. of Bose's position that the loudspeaker system described in *Speaker Builder*, Issue 2, 1994, pages 36–40, titled "The Birdhouse: A Sound-Reinforcement Subwoofer" by Matthew Federoff, infringes U.S. Patent No. 4,549,631, which Bose Corporation owns.

TIME RESPONSE OF CROSSOVER FILTERS

By Dewald P. de Lange

If the individual outputs of the drivers in a two- or three-way system were to be combined, the result should be a replica of the original input signal. In this article the combined response of the crossover filters only is considered, i.e., driver response is taken as ideal (although the distortion it causes is probably the overwhelming factor). Speaker-system designers appear to exclusively use SPL-frequency response to analyze crossover networks.

FREQUENCY RESPONSE

This type of analysis relies on the fact that any arbitrarily shaped signal can be built up from sinusoidal waves of various amplitudes and frequencies. If sine waves of all frequencies proceed through a crossover network with a constant amplitude and time delay, it follows that other, more complex input signals will retain their original shape. The SPL-frequency response consists of magnitude and phase.

The phase is usually converted to a quantity more directly related to time, namely phase delay (phase over frequency) or, more commonly, group delay (the rate at which the phase changes with frequency). Designers then aim for an SPL-frequency response that has a flat magnitude and a flat group delay. This method is fine if these quantities are indeed flat. When they are not, however, it is hard to imagine how seriously a real signal will be affected.

TIME RESPONSE

Time responses provide a more direct indication of what a crossover filter does to real signals. Of the two common time test signals, impulse and unit-step, the latter is used here. The step response indicates what the edges of a low-frequency square wave would look like.

A two-way crossover network is used for simplicity in this article. The responses shown are all derived by computer simulations. The SPL graphs are "on axis," being the response straight in front of the speaker system.

The simplest crossover filter is a first-order one, as illustrated in *Fig. 1*. As can be seen from the middle graph, the combined frequency response has a flat SPL magnitude with zero group delay, which is exactly what we want. (Zero group delay implies a constant phase shift against frequency, which would not be a constant time delay, unless the phase is zero, which it is). On the right the output step response is also a perfect replica of the input step.

In *Fig. 2* we have a second-order low- and high-pass filter, both with the same cutoff frequency. At this cutoff frequency the out-

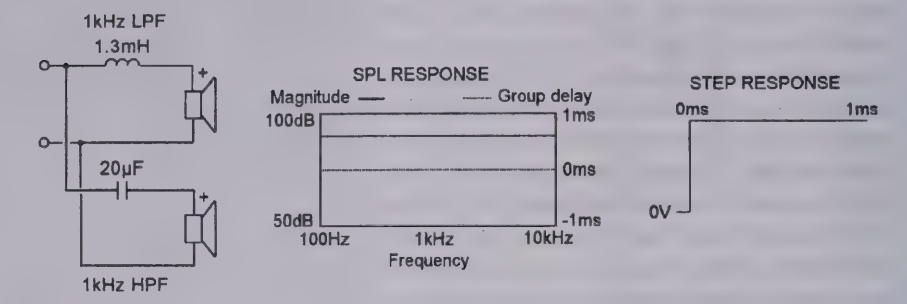

FIGURE 1: First-order crossover filter response.

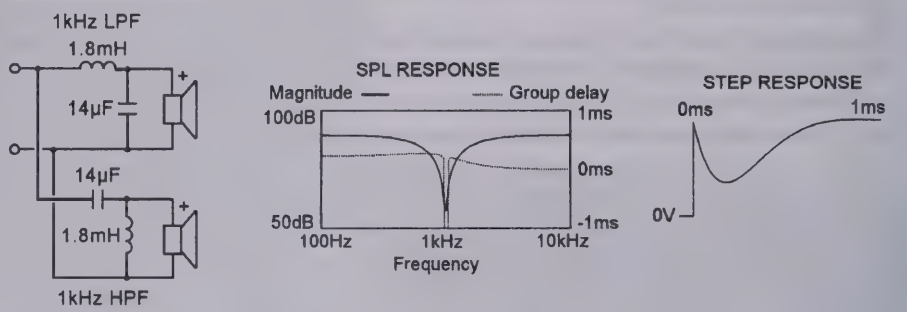

FIGURE 2: Second-order Butterworth crossover filter response.

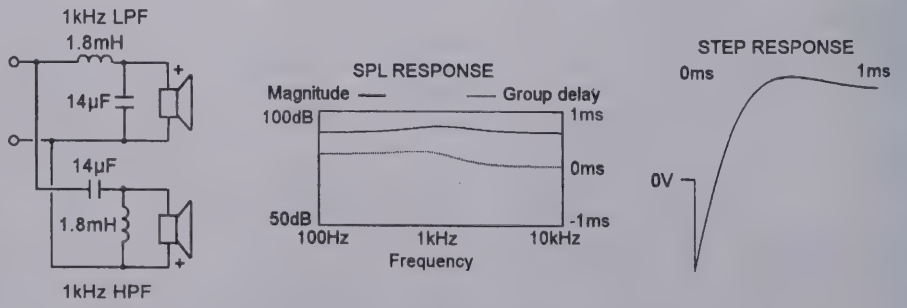

FIGURE 3: Second-order Butterworth crossover with reversed tweeter polarity.

ABOUT THE AUTHOR

Dewald de Lange resides in Cape Town, South Africa. He obtained a B. Eng. (electronics) degree from the University of Stellenbosch. He has mostly worked in the radio frequency field. He recently wrote a basic speaker system design program.

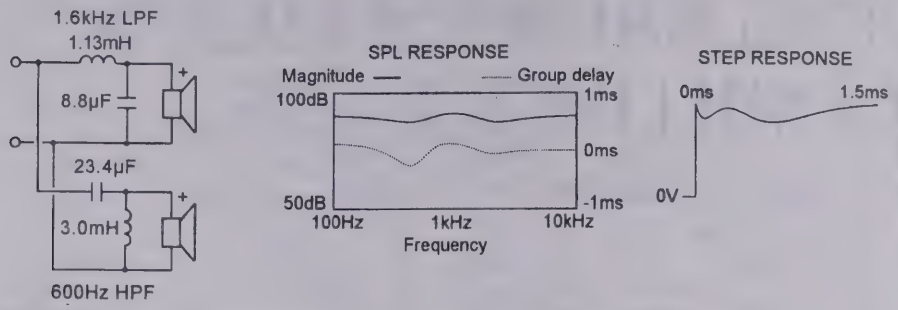

FIGURE 4: Second-order Butterworth crossover with overlapping cutoff frequencies.

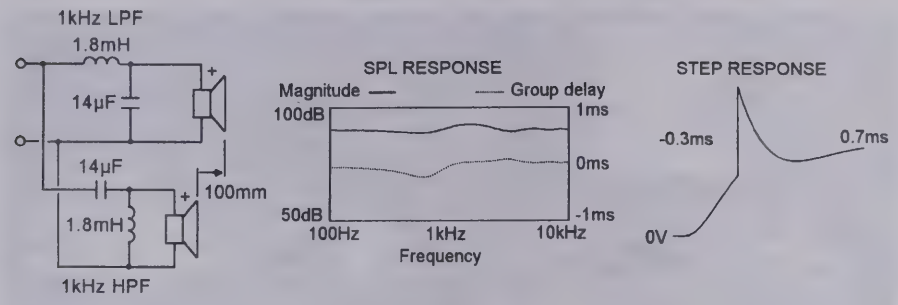

FIGURE 5: Second-order Butterworth crossover with woofer offset.

put from the low-pass filter is out of phase with that from the high-pass filter. The result is a combined output of zero, as can be seen from the dip in the amplitude response. At this frequency there is also a sharp dip in the group delay response, because the phase changes from –90° below the cutoff frequency to +90° above it.

In principle the dip in the amplitude response means the speaker system would produce no sound at 1kHz and the filters must be modified in some way. Before we take a look at possible mods, note that the step response, instead of a flat line after the step, has a dip. The initial spike comes from the high-pass filter. A slight delay occurs before the low-pass filter comes into action. The combined result is the distorted step response shown. The literature contains a suggested method for eliminating the null in the frequency response of a second-order filter, namely to reverse the polarity of one driver (e.g. the tweeter). The two outputs should now be in phase. *Figure 3* indicates the result.

The bulge in the SPL magnitude at 1kHz is compensated for by moving the cutoff frequencies of the two filters apart. However, the step response indicates a fundamental problem, not apparent from the frequency response. As the polarity of tweeter is now reversed, it produces a negative spike.

Another way of addressing the null in the frequency response would be to overlap the low-and high-pass filters (*Fig. 4*). The SPL response is now continuous, except for ripples of a few decibels. The step response also shows an improvement. However the overlapping slopes now provide little improvement in attenuation over a first-order filter.

Finally, one might consider moving the woofer forward, which largely removes the dip in the SPL magnitude (*Fig. 5*). The step response is not totally unexpected, with the low-pass filter output appearing before the spike from the high-pass filter.

Third-order Butterworth crossover filters have the advantage that their outputs sum to a flat magnitude response, as shown in *Fig. 6*. The uneven group delay indicates that the

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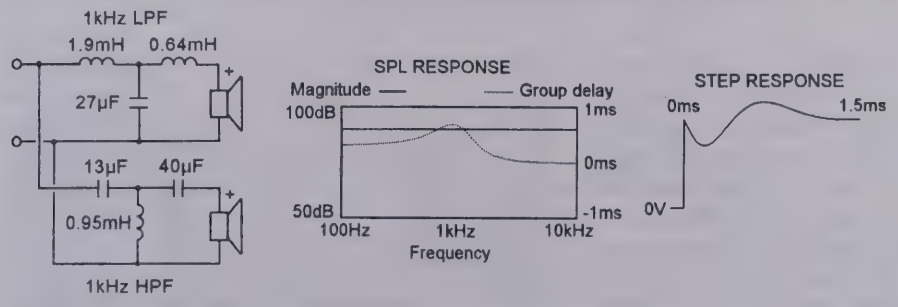

FIGURE 6: Butterworth third-order crossover filter response.

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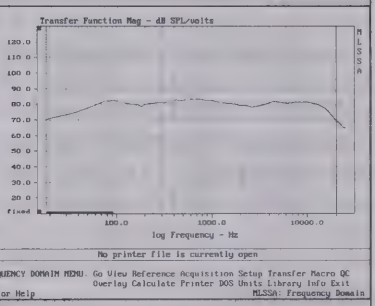

These measurements are taken without any form of filtering. Frequency Range: -6 dB at 70 Hz, from 150 Hz to 16 kHz ±2,5 dB, -3,5 dB at 16 kHz and above.

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CONVERTING RADIO SHACK'S SLM TO MILLIVOLT USE

By C.L.P. Carrington

Until I read Joseph D'Appolito's article "Mitey Mike" (*SB* 6/90, p. 10), I used a borrowed Shure Equalisation Analyser to make measurements on the speakers I built. However, the one-octave bandwidth of the analyzer and the 3dB incremental display limited the equipment's usefulness for the analysis I wanted.

Mitey Mike proved highly satisfactory when used in conjunction with Dick Crawford's Warbler (*TAA* 1/79, p. 22). The results were so good, in fact, that I built another measurement system for a friend who is equally enthusiastic about speaker building (probably more so; he has embedded concrete LF horns in the cellar of the home he constructed!).

Having built the Mitey Mike and Warbler, I looked around for an economical mV testmeter to use with them, but without success. My friend used a Radio Shack sound level meter (Cat. No. 33-2050) for balancing his system, which I reckoned could be modified to provide the mV testmeter we sought. Radio Shack includes a schematic diagram in the instruction booklet, which facilitated the modifications.

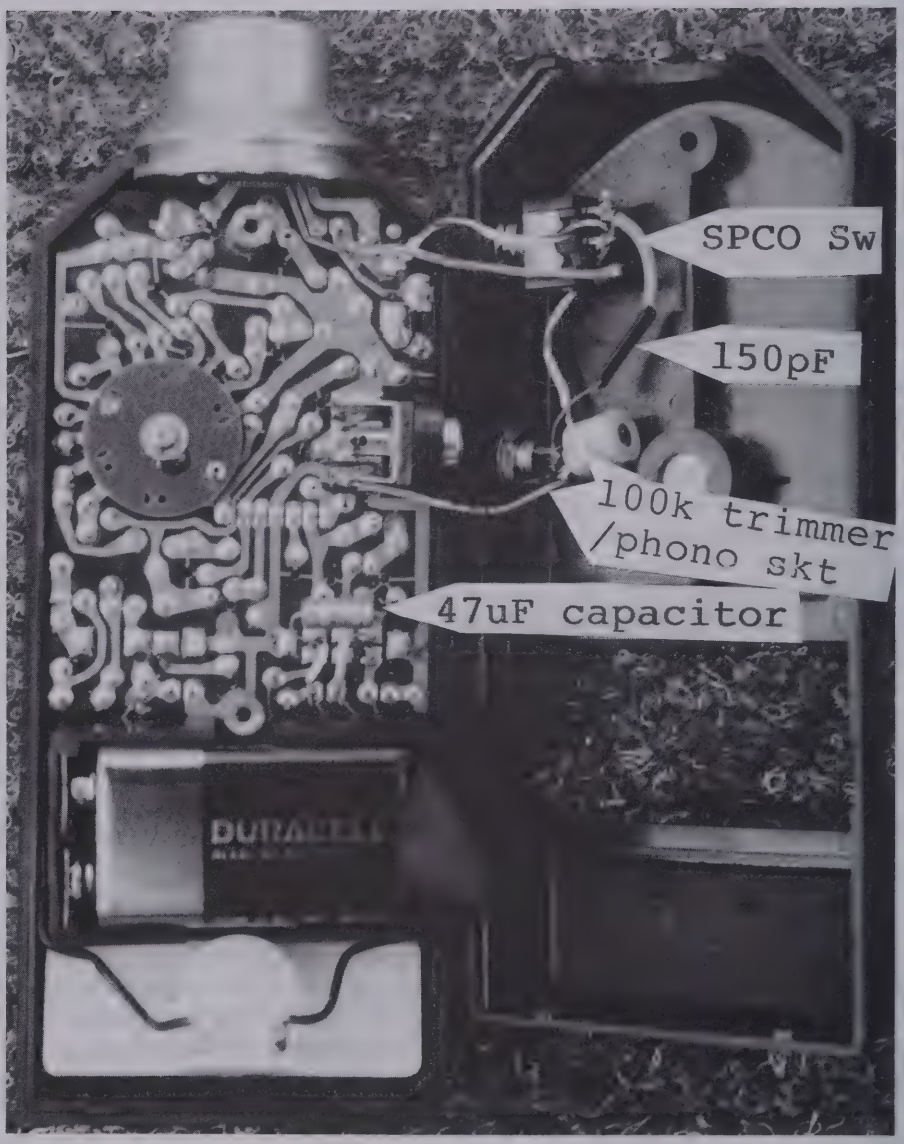

PHOTO 1: The modified Radio Shack sound level meter.

MINIMAL MOD

My aim was not to rebuild the sound level meter but to modify it, minimally, in such a way that it could still be used as originally intended. Accordingly, I provided a line input to Q1 through an SPCO switch that would select either the SLM microphone or the line input. I achieved this by cutting the printed circuit between C1 and the "CAL" potentiometer VR1; *Photo 1* shows the location of the track to be cut.

With the meter switched to "C" weighting, measurements made by connecting an audio oscillator to the top of VR1 indicated that for 0dB on the 120 scale to correspond to 0.775V (i.e., 0dBm), some attenuation would be required. On this basis, the most sensitive position (60dB) would equate to 0.775mV (–60dBm). I detected a fall-off in response at the low and high ends of the 20Hz–20kHz passband.

To carry out the modifications, mount an SPDT miniature toggle switch (RS 275-625) and RCA phono socket (RS 274-852) in the side of the rear housing—the switch ½" from the cutout for the "output" socket in the direction of the microphone and the socket ½" on the other side of the cutout. Both should be mounted halfway up the side of the housing (*Photo 1*).

Using a scalpel or craft knife, cut the copper track leading from C1, in line with the metal pillar. (With the sound level meter lying face down, as shown in *Photo 1*, this track is just to the right and below the microphone; it is the fourth track to the right of the pillar.) Solder a wire to the track near C1 (the half of the track nearest the microphone) and connect this to one side of the SPDT switch. Connect the other end of the severed track to the center pole of the switch so that with the switch thrown in the direction of the SLM microphone, the meter will operate as before, i.e., as a sound level meter.

Connect a wire from the solder tag, retained by the nut of the phono socket, to the

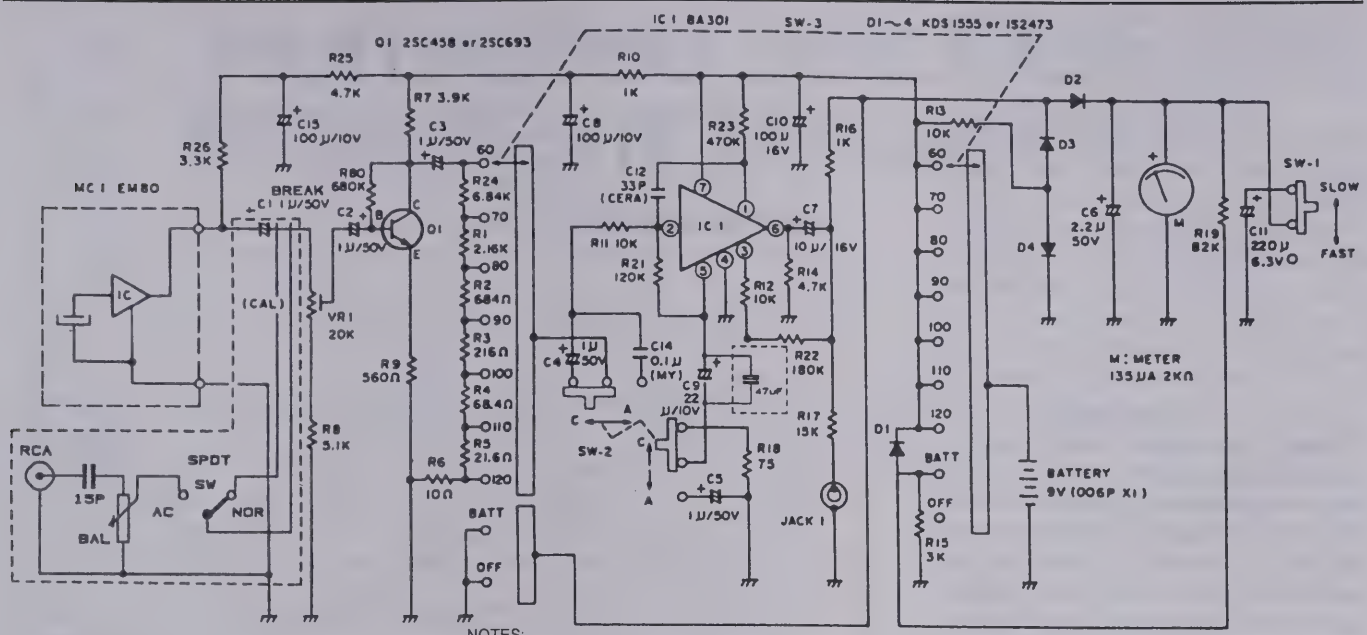

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FIGURE 1: Schematic for the Radio Shack sound level meter, with modifications.

U-shaped piece of metal grounding the "output" socket. Connect a 100k trimmer resistor (RS 271-284 would probably do) across the RCA phono socket, and connect the wiper of the trimmer to the free side of the SPDT switch. When the switch is thrown in the direction away from the microphone, the "input" line socket should now be connected to the amplifier, via the trimmer.

DECLINE AND FALL-OFF

An accurate audio-signal generator will facilitate the calibration. Connect the generator to the "input" socket with the switch thrown to put this in circuit. Feed a 0dBm (0.775V) signal at lkHz into the input, switch the meter to the 120 scale, adjust the trimmer for 0dB (at about the midpoint [electrical] of the trimmer), and switch to 20kHz. If you see any fall-off in reading, shunt the top half of the trimmer by a small capacitor to compensate.

The two meters I modified differed in sensitivity; you may need to proceed by trial and error. Start by connecting a 150pF capacitor from the input socket to the wiper of the trimmer (or to the side of the SPDT switch connected to this point). If you're lucky, the 20kHz level will now correctly read 0dB. If it reads high, reduce the capacitor to, say, 120pF. If it reads low, increase the capacitor. It may be worthwhile to buy a trimmer capacitor (RS 272-1336) and adjust it for the correct reading. By judicious mounting, this could be accommodated within the enclosure.

Switch to 20Hz. You will note a minor fall-off in reading on the meter; you can correct this by shunting C9 with a 47µF capacitor. Six soldered connections are in line to the

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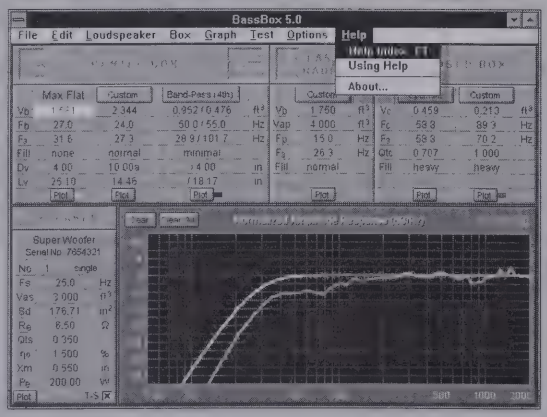

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ACOUSTIC DISTORTION AND BALANCED SPEAKERS

By Donald Jenkins

Designers of sound systems pay a great deal of attention to the matching, balancing, and phasing of components in their quest to achieve optimal performance. They will measure and analyze electrical characteristics of the components and give careful thought to the design and construction of the speaker enclosure.

But what about the question of speaker-to-speaker acoustic balance? Do supposedly identical speakers actually deliver the same performance when energized with the same inputs? The results of my investigation show that in some cases what are considered equal acoustic systems may in fact display large acoustic dispersion over a small frequency range.

The test that I devised measured speaker (that is, voice-coil) current, which I took to be a linear indication of acoustic power. If the two speakers are constructed in the same way and mounted in similar enclosures, an identical input voltage to each should produce identical currents at a given frequency; if the speaker currents are equal, the speakers' outputs are considered equal. Of course, a better way to assess speaker output is to measure the actual cone displacement, but for speakers installed in enclosures that is hard to accomplish.

So for purposes of this test I have assumed that equal currents in the same design produce equal displacements.

In this steady-state test, I measured the RMS value of the current over a range of frequencies. I used a ten-second linear frequency sweep from 20–175Hz at a constant input voltage to the speaker, keeping the sweep-frequency range intentionally narrow.

Figure 1 shows a standard plot of imped-

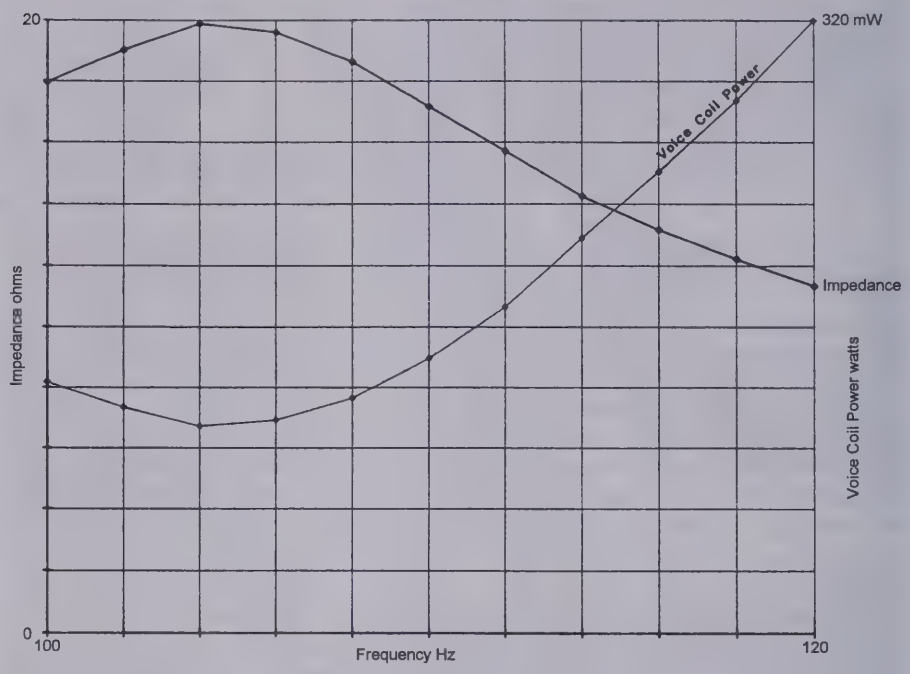

FIGURE 1: Impedance versus frequency; Sony APM-215/Heath W-5M.

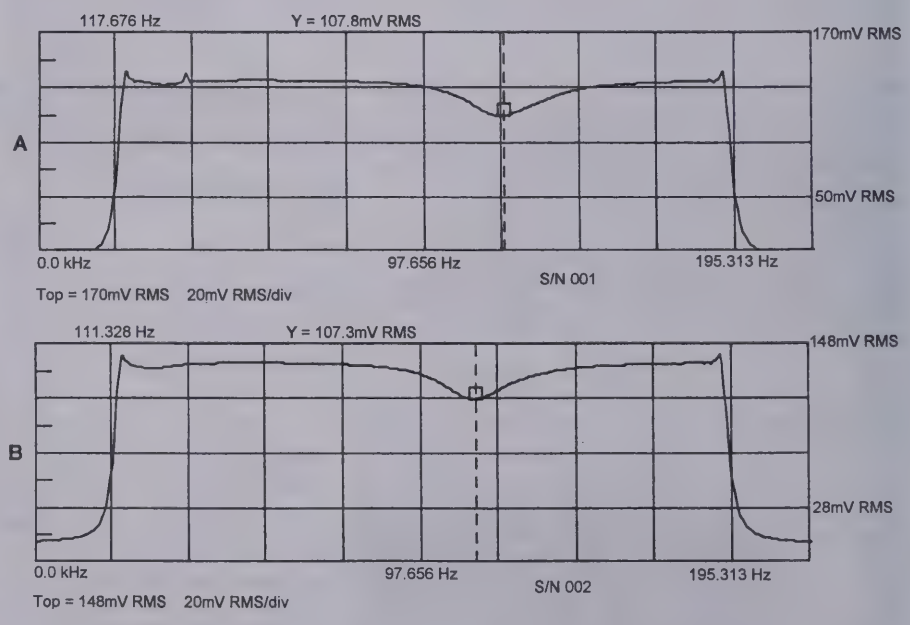

FIGURE 2: Slow-sweep spectra of a "matched" set of Sony APM-215 speakers: *a.* speaker A; *b.* speaker B.

ABOUT THE AUTHOR

Don Jenkins is a retired engineer from the aerospace business, and is now consulting part time. He considers the study of amplifiers strictly on the hobby side of his interests. The "professional" side is concerned with rocket propulsion system design and development, with some activity on the projection technology committee of the SMPTE.

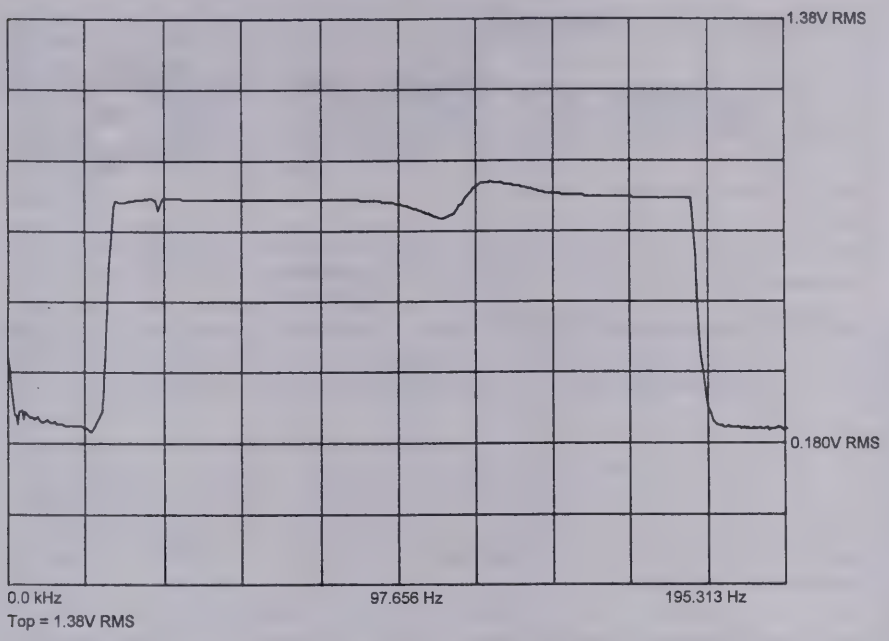

FIGURE 3: Ratio of the two curves, S/N 001 to S/N 002, for the Sony APM- 215s. The plot is derived by line-by-line division across the test spectrum.

ance versus frequency for a typical set of speakers mounted in an enclosure. At greatest impedance, speaker current is lowest. I have defined acoustic dispersion as the difference in output currents at the same input frequency.

The speaker unit I used in this test was a SONY APM-215. It has three small drivers built into a bookshelf-type cabinet and is designed to be used with its companion amplifier, the Model TA-215. Note that the speaker current is lowest at about 104Hz.

Figures 2a and *2b* show the slow-sweep spectra of a "matched" set of these speakers. The spectra are similar. *Figure 3* shows the ratio of the two curves, S/N 001 to S/N 002, derived by line-by-line division across the test spectrum. The resultant curve is the difference, therefore the acoustic distortion, that results from equal inputs to the two speakers. It reaches a maximum of about 6% in the 100–120Hz frequency range.

This means the same input to these speakers would result in as much as a 6% difference in acoustic output at those frequencies. It also means that a small frequency shift could actually shift the relative amplitude from one set of speakers to the other. As a result, any stereo separation in this range would not only be lost but would actually be channel-reversed.

This explains why I used such a narrow band of frequencies in the test. The mismatch phenomenon is greatest at the resonant points of the speaker/enclosure/crossover network assembly, since this is the frequency range in which the rate of change in the impedance is greatest. Achieving a true match of all these elements is clearly quite difficult.

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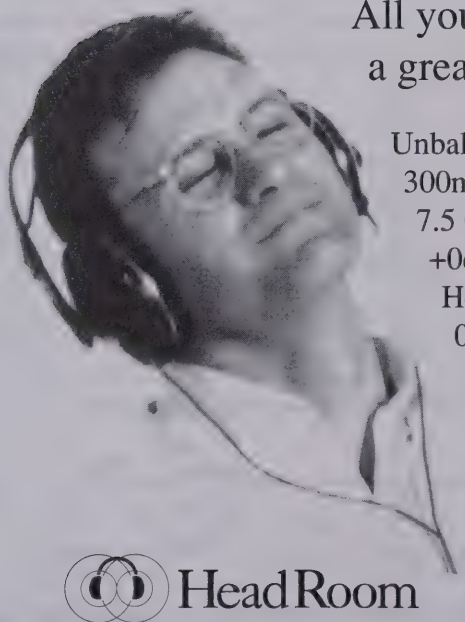

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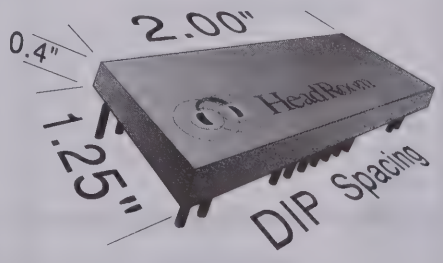

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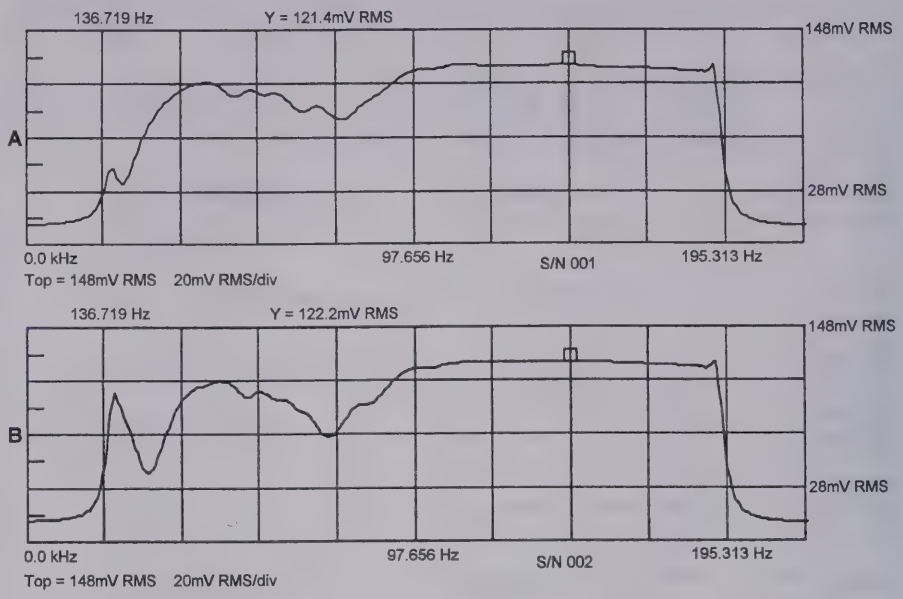


FIGURE 4: Slow-sweep spectra of a "matched" set of Heath SS-1/-1B speakers: *a.* speaker A; *b.* speaker B.

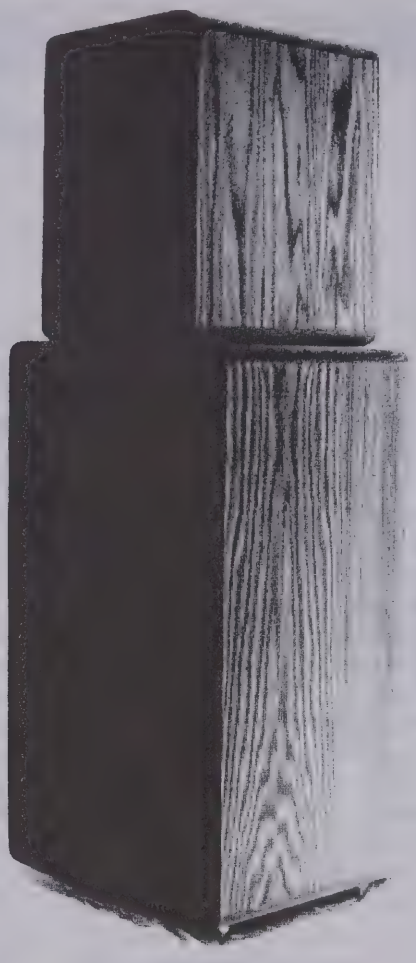
To see whether this response was unique to my Sony speakers or is typical of any two matched speaker units, I compared them with two Heath SS-1 and SS-1B units. Each of these units consists of four speakers in two cabinets with crossover networks.

These two assemblies proved to be even more divergent in input. *Figures 4a* and *4b* are the two sweep spectra, and *Fig. 5* is the ratio plot. For these units, peaks in acoustic distortion are in the 20Hz range and have an amplitude as high as 100%; there is also distortion of about 32% near 75Hz. Here too, a slight shift in frequency will result in apparent channel shift.

The results of these tests show that this type of unsymmetrical acoustic distortion will probably occur with any two speaker sets used for discrete channel reproduction. Whether or not this unbalance is discernible to the listener is the next question. Critical listening tests might provide an interesting answer.

FIGURE 5: Ratio of S/N 001 to S/N 002 for the Heath SS-1/-1Bs.

NEW FOR 1994

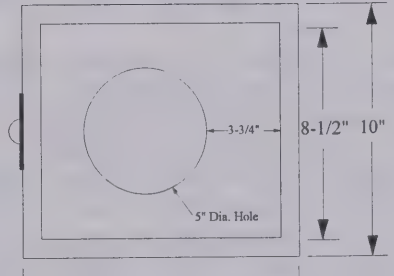


The *J-3* incorporates the proven Vifa P17WJ with the new 26mm D27TG silk and fabric dome tweeter. The D27 has a lighter and better controlled diaphragm than typical cloth domes. A tuned cavity, and linear response faceplate add to the technology Vifa designed into this driver to compete with other fine Danish tweeters. Response has been favorable to this new addition to the Vifa line.

The crossover is computer optimized as a forth order Linkwitz-Riley type. Text-book formulas could not produce the values used in this effective circuit. The designer and Brian Smith of A&S worked to overcome problems such as the nasality and "honk" associated with some bass-mids, and the brightness and sibilants associated with many dome tweeters. The use of LMS and LEAP together, with careful listening, have refined the *J-3* into a very musical two-way speaker with excellent bass, mid-range tonal balance and smooth, extended highs.

The *J-3* overcomes many of the problems associated with DIY projects and less sophisticated commercial systems. The enclosure features a computer-optimized, third-order (slightly modified QB3) alignment. In this design, both enclosure tuning and the natural resonance of the bass-mid are below the -3dB point of the system. This alignment provides better transient response and less colorations than conventional forth-order vented designs. To reduce cabinet resonances, and further improve bass clarity, the inside walls of the enclosure are covered with two layers of Black Hole Pad. Careful use of other materials completes the damping of the cabinet.

The *J-3* is -3dB at 49 Hz and -10dB at 35 Hz., but may be combined with matching subwoofers. These computer-optimized modules incorporate the new Vifa M26WR 10 inch woofer and will extend the system response down to 26 Hz.



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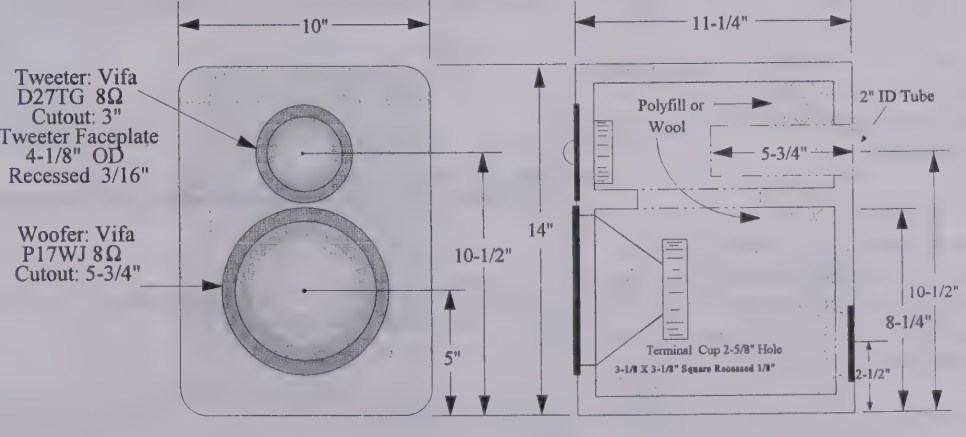

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Reader Service #31

MICROPHONE RESPONSE CORRECTION WITH IMP

By Bill Waslo

Loudspeaker measurement accuracy is greatly affected by your microphone. A cost-effective microphone of known characteristics can achieve very accurate results. It can also remove any requirement for an expensive, flat measurement microphone. One advantage to using a computer system for measuring and processing is that known, linear effects, like microphone response and sensitivity, may be digitally corrected with little effort.

Even without correction the little microphone capsule used with IMP (as well as in Mitey Mike and others) has quite good characteristics. But better results can be automatically obtained when correction data is provided to the IMP or IMP/M software. (Note: the P9932 capsule used in IMP and Mitey Mike was discontinued. The P9939 (WM60) capsule has been found to be a good substitute.)

It is easy to get confused with terminology here, so let me make a few distinctions. "Cal" in the IMP system refers to the normalization process that eliminates response errors resulting from the electronic test equipment. "Correction" refers to the process that eliminates response errors due to the microphone—usually accomplished with the help of microphone calibration data (as it is called in the world beyond IMP). Beware of confusing "cal" with "calibration." Within this article, the microphone calibration data is called "correction data."

JUST SAY WHEN

There are several ways to determine how and when to apply microphone correction. An obvious approach is to apply the correction curve whenever you desire. But this adds yet another thing for you to keep in mind during a system measurement. If your lab is like mine, you probably don't need more complication in your work.

The alternative IMP provides is saving a record of the input from which time data was acquired. If the data is taken from the microphone input, IMP assumes the measurement is acoustically based and that the correction should be applied. This tag, which defines the source as a microphone, is also attached to any frequency responses transformed from the time data. Of course, when exceptions to these rules are desired, there are menu options under [* Display Mic_correction]. These options allow you to force a time or frequency domain data set to be treated as either acoustic or nonacoustic.

The correction is actually applied only to frequency-response data, and then only when a "cal" is, or has been, performed on the data. Remember: an IMP "cal" refers to the process of normalizing the spectrum at a speaker's output with what is at its input. The frequency response yielded is independent of the IMP pulse shape or amplification/filter effects. Microphone correction is done only with cal as a practical matter. If you need rapid, noncritical results, you may not want to wait the extra time required to apply the correction to each FFT result; and, the response rolloff from pulse shape and anti-aliasing filters is usually greater than that of the IMP microphone. Therefore, microphone correction is of little value without cal application.

Although this arrangement is complicated to describe, it is easy to use. Just have the microphone correction data available and loaded into the IMP software. IMP will apply it when appropriate. When both microphone correction and system cal are applied, a small "m"

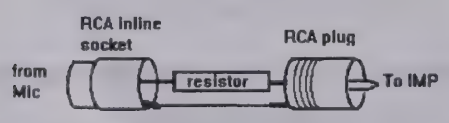

FIGURE 1: Simple attenuator for Mitey Mike to reduce sensitivity with IMP.

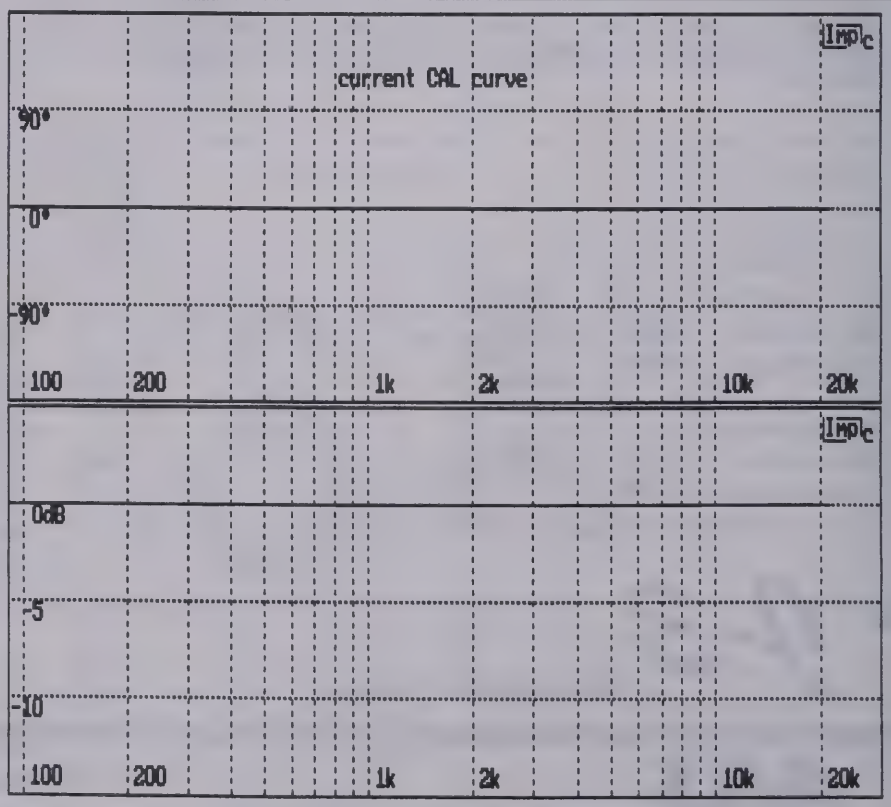

FIGURE 2: A "flat" dummy cal curve.

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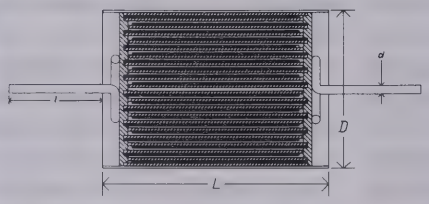

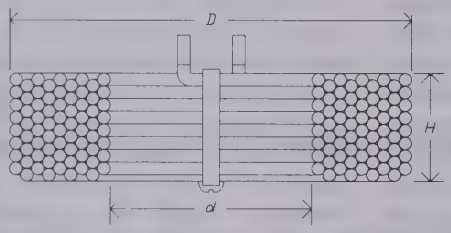

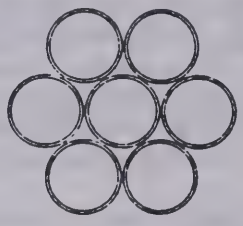

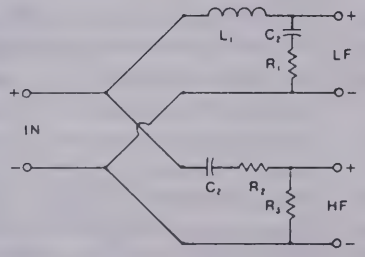

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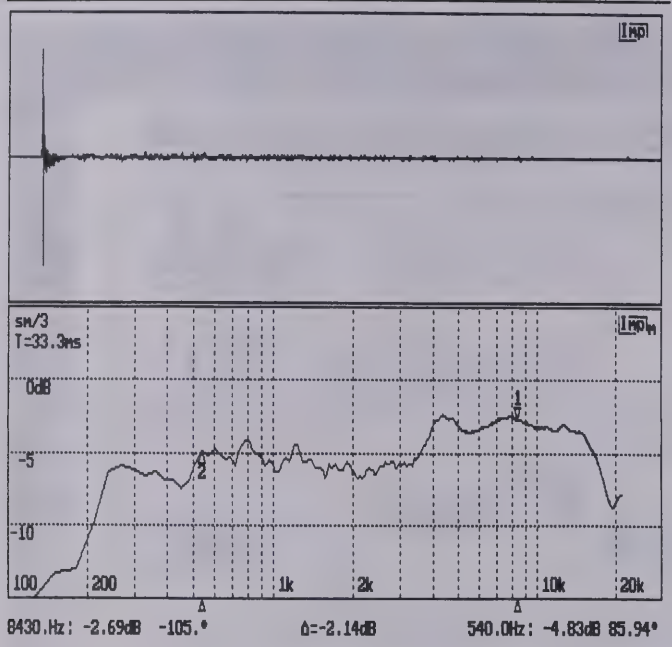

FIGURE 3: Corrected for response for Mic only (using flat Cal).

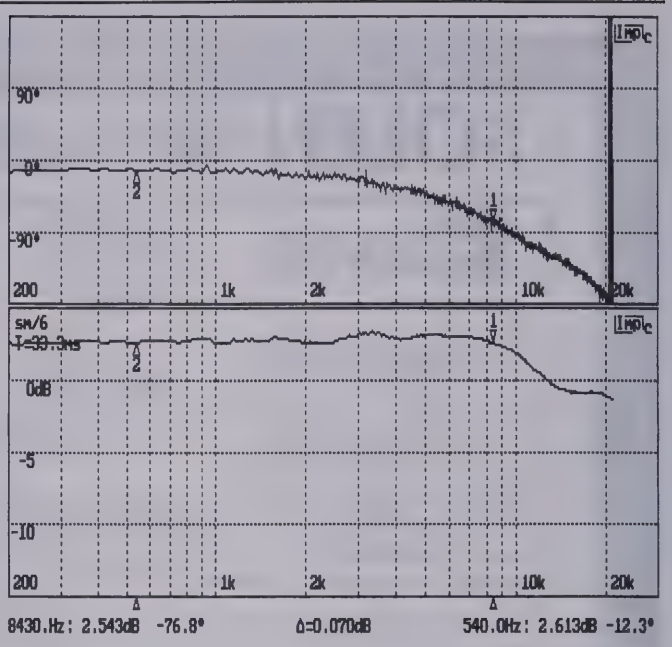

FIGURE 4: The correction data (response curve) for the WM60 sample mike.

will appear by the IMP logo. A small "c" by the logo indicates that only cal has been applied.

One further note: in IMP, the microphone correction is used only with the higher sample rate. The capsules commonly used are much better for low-frequency performance than for the high-frequency end of the audio spectrum. In addition, conventional memory use in IMP is already approaching maximum. Therefore, correction data below approximately 1kHz is not usually needed.

NAILING DOWN A REFERENCE

Of course, the correction cannot be applied at all if it is not available and loaded into the computer in a compatible way. Several options are available, depending on cost, effort, and the level of precision you require.

The easiest and cheapest route is to use the "TYPMIC.DAT" correction file provided with the IMP software. This will generally get you very close, within about a decibel flatness, according to a quick comparison of five P9932 capsules. If you wish to have more accurate results, or if absolute sensitivity levels are an issue, you will need measured data for the individual capsule you use.

Several microphones with calibration data from sources such as Old Colony and Josephson Engineering are available. Use an electret condenser type, not a dynamic microphone. The bias current supplied from the IMP microphone input is undesirable in a magnetic circuit. You may, with some microphones, have to adapt connectors to match IMP's input. If your microphone's output impedance is not very low, you may have to modify the correction data sensitivity to account for the 2.21kΩ loading of IMP's microphone input as well. (This should not be a problem with Mitey Mike.)

The cal data on your microphone might be in disk or paper form. The format required by IMP for microphone disk-based data is described in two places. Refer to the "IMP-READ.ME" file of version 1.21 or to the Help files in version 2.00 and above. Don't despair if you have only a paper graph or frequency list. You can enter the data into an ASCII file using any of a number of common text editors. Just follow the specified format. IMP interpolates its required values from the frequency points provided—so the correction resolution is determined by the number of points you wish to enter. If phase data is not provided, just neglect it. IMP will assume the phase correction is zero—not likely to be very far off below about 15kHz.

When you are finished, name the new file with a ".DAT" extension (example: MYMIC.DAT). Then copy it into the directory from which you run IMP. Go to [* File Micdat], load, and you are ready to go.

DIY CORRECTION FILES

Another approach is to borrow a calibrated microphone and use IMP to generate your own correction data for your capsule. A microphone calibrated this way will be less direct in its calibration pedigree. (It's one more step away from the initial standard, and some error can be expected to be introduced at each step.) It does have the advantage of providing do-it-yourself satisfaction, however. This can be done efficiently, but will require some care, a little trickery, and some explaining.

One problem that arises in this process is gross differences between microphone output levels. If absolute sensitivity is not required, just use IMP's microphone level control for compensation. But, if you need to determine sensitivity also, you must keep track of the attenuation. The easiest method is to attenuate the higher-output microphone down close to the level of the lower-output microphone.

For example, to attenuate Mitey Mike down near the sensitivity of the bare IMP capsule (as powered from the IMP module), a resistor of 10.0kΩ can be installed. Positioned in series with the hot lead at the IMP module microphone input, this provides 15dB attenuation. This resistor should be installed with connectors, so that it can be included without permanent modification to either IMP or the Mitey Mike (*Figure 1*).

MITEY MIKE ATTENUATOR

If you use such an attenuator, be sure to adjust the sensitivity value of the correction data for the high-output microphone accordingly. That way, IMP is aware of the change. (Make a separate file of the correction data for such use.) For instance, for a 10.0k resistor used with a Mitey Mike, divide by 5.656 the number preceding the expression "mV/Pa" in the first line of the correction data file. In this instance, 15dB = 20*log(5.656).

You'll need a speaker with reasonably flat response, preferably within 10dB over the desired correction frequency range. IMP will measure the speaker with both microphones and will use the data from the corrected microphone. Then IMP will calculate the correction curve for the unknown microphone based on the differences in the measurements.

This step-by-step procedure will enable you to generate a correction file. Menu directions apply to the IMP/M version of the software:

1. Make a dummy cal file. That way, you can trick the program into applying only the microphone data to a curve. (Microphone correction can only be applied when cal-ing.) Take any non-cal'd impulse response of high RATE and large size, such as the simulated pulse the IMP program shows when first started. A size of 1,024 or 2,048 should be OK. FFT all of it. (In the time curve before FFTing, put marker 1 at 1 and marker 2 at SIZE.) Declare the curve as "cal" using [* Transform Set_cal]. Then, by the key sequence [* Display Mic_correction Freqdata No], tell the program the response is nonacoustic. Set delay to zero (via F9), and the frequency response GAIN (F8) to 0dB. Do [* Transform Cal]; i.e., cal it with itself. You should get a straight line of 0dB and 0° phase (*Fig. 2*). Declare this line as your new cal data by [* Transform Set_cal] again. You might want to save this cal curve ([File Save fCal_file]) under the name "Flat," so you don't have to go through this again.

2. Make sure that the proper Microphone Correction file for your calibrated microphone is loaded in via [File Micdat].

3. Use your calibrated microphone to make an acoustic-only measurement of a loudspeaker that has reasonable response. (MLS is definitely preferred.) This measurement may be anechoic (echoes edited out) or not, as you like. The valid correction curve will extend lower if echoes are included. Including echoes, however, makes repeatability of microphone placement more critical. If you include echoes, you may want to deaden the area around the speaker and microphone by spreading blankets or pillows. Doing so minimizes high-frequency reflections and gets a smoother curve.

Do not use a probe "cal" procedure with this measurement. FFT and then cal this curve using the dummy cal data. This ensures that only the microphone correction is applied to its frequency response when the "Flat" cal is used. Declare this modified curve as the new CAL using [* Transform Set_cal]. This is the speaker's response under arbitrary but controlled conditions—corrected for microphone response, but *not* for IMP pulse shape or filtering (which need not be dealt with in this procedure). This curve will show significant rolloff above 15kHz because of the lack of a proper cal (*Fig. 3*).

Carefully note the exact position of the calibrated microphone capsule in space. You will need to position your new microphone, as accurately as possible, in this same spot.

4. Replace the microphone with the new mike for which you wish to derive a response curve. Mount it in the same way and at the same point as the reference microphone. Then measure the impulse response of the same speaker under the same conditions of amplifier gain and IMP control settings.

5. Tell IMP not to use microphone correction data by the command sequence [* File Micdat]. Then type "none." FFT and "Cal" the new response. This divides the curve using the new microphone with the curve using the corrected reference microphone. The result is the correction curve for the new microphone.

6. The true differences between the responses should tend to be relatively wideband (i.e., no hash or fine ripples), because the microphone diaphragm lacks the high Q for narrow-band effects. Any such narrow ripples you see are apt to be due to differences in microphone placement affecting echo-induced cancellations—usually narrow-band effects. You will induce little error and get more coherent results by just smoothing these

Continued on page 56

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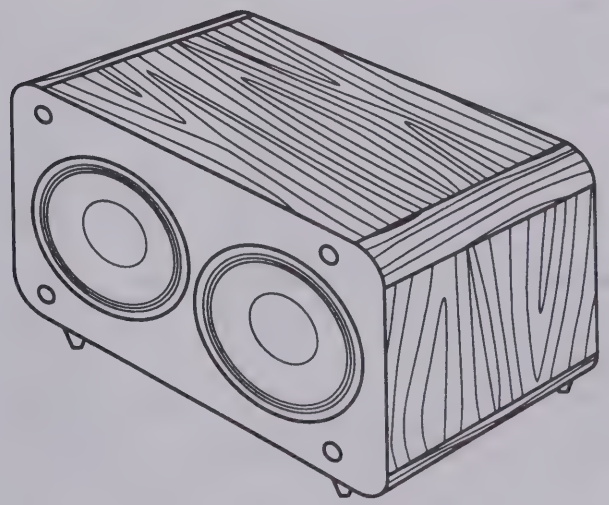

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Reader Service #12

MORE ABOUT THE BIRDHOUSE BANDPASS

By G.L. Augspurger

In *SB* 2/94 ("The Birdhouse: A Sound Reinforcement Subwoofer," p. 36) Matt Federoff details an unusual bandpass design. After a quick glance at the article my reaction was, "This can't be right."

The article describes a tandem, double-tuned bandpass system with a 105-liter chamber tuned to 90Hz and a 65-liter chamber tuned to 30Hz. To the best of my knowledge this is a first. Commercial tandem designs always tune the smaller chamber to the higher frequency. The original Bose patent (4,549,631) sets forth a preferred range of resonance frequencies as 1.5:1 to 2:1, with corresponding subchamber volumes of 1:2 to 1:4.

The author apparently obtained his published response curve from Bullock's bandpass program. I ran a quick simulation with my own computer model and got an almost identical curve (*Fig. 1*). My model also confirms that output in the passband is almost 5dB greater than the rated sensitivity of the woofer.

WHAT'S GOING ON HERE?

The overlaid impedance and cone excursion curves give a hint, but it took several additional computer comparisons before the situation became clear.

Let's start with the basics. A good first step in synthesizing a double-tuned bandpass system is to find a practical vented box alignment for the loudspeaker in question. *Figure 2* shows 1W, 1m response of the PS-15C woofer in a 250-liter box tuned to the speaker's free-air resonance of 35Hz. Low frequency output is about 95dB with the –3dB frequency at vent resonance.

To transform this box into a double-tuned bandpass system with a comparable low-fre-

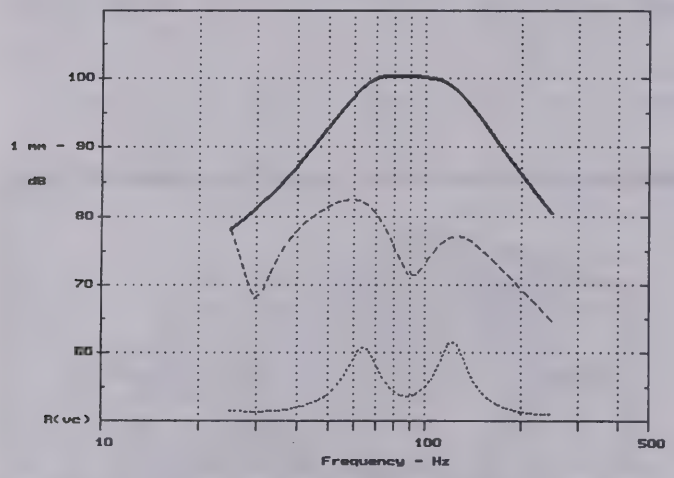

FIGURE 1: PS-15C in 65/105 liter Federoff bandpass box.

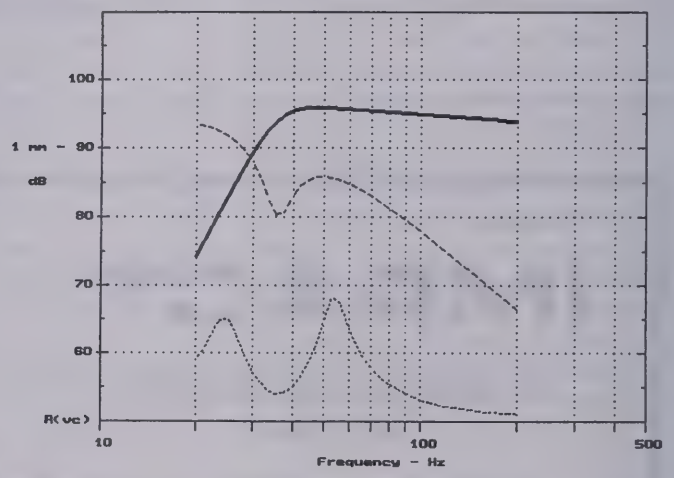

FIGURE 2: PS-15C in 250-liter, 35Hz vented box.

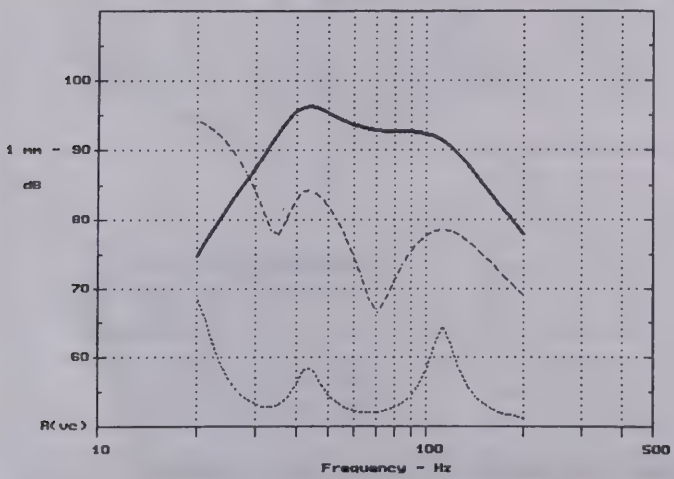

FIGURE 3: PS-15C in 65/185 liter double-vented bandpass box.

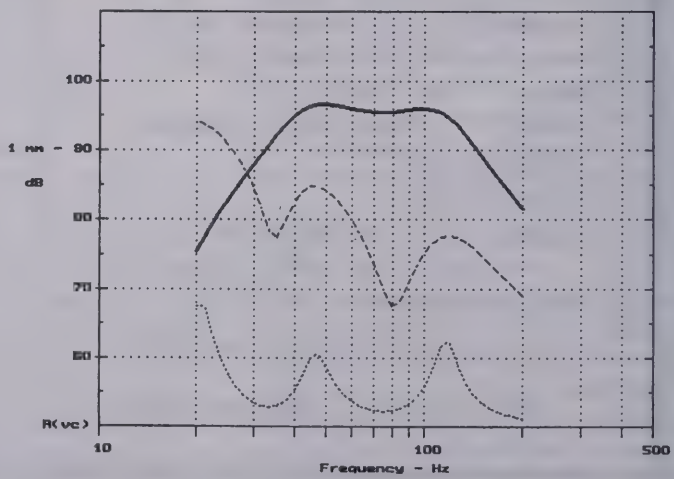

FIGURE 4: PS-15C in 70/180 liter double-vented bandpass box.

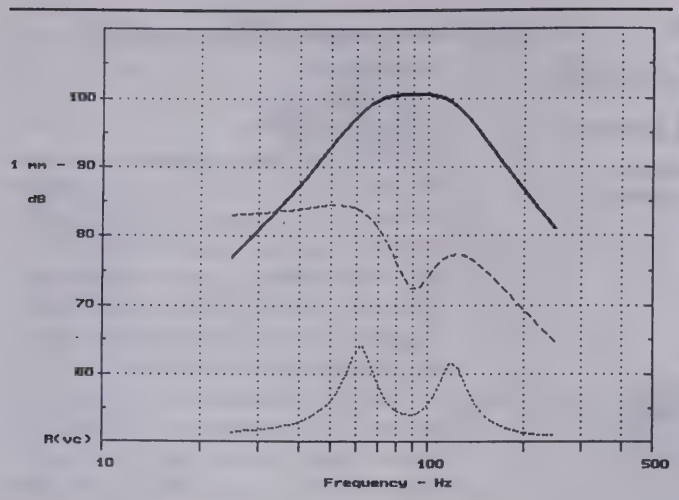

FIGURE 5: PS-15C in 65-105 liter sealed-vented bandpass box.

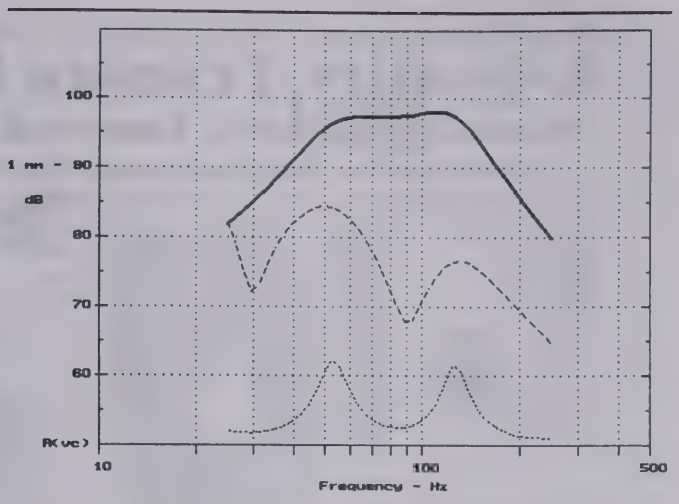

FIGURE 6: PS-15C in 65-105 liter reverse-tuned Federoff bandpass box.

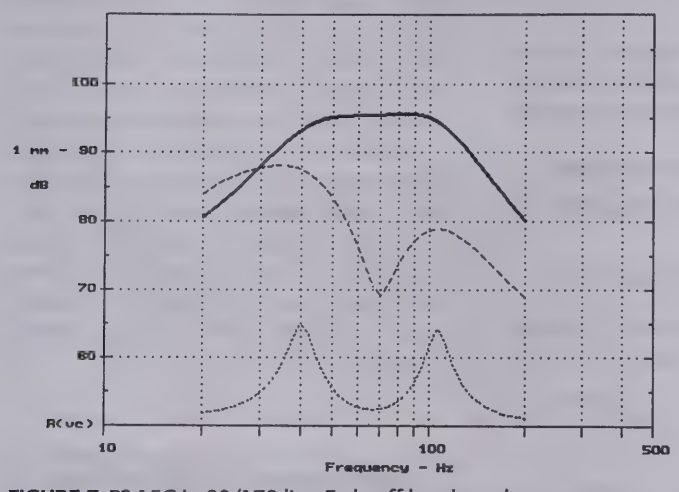

FIGURE 7: PS-15C in 80/170 liter Federoff bandpass box.

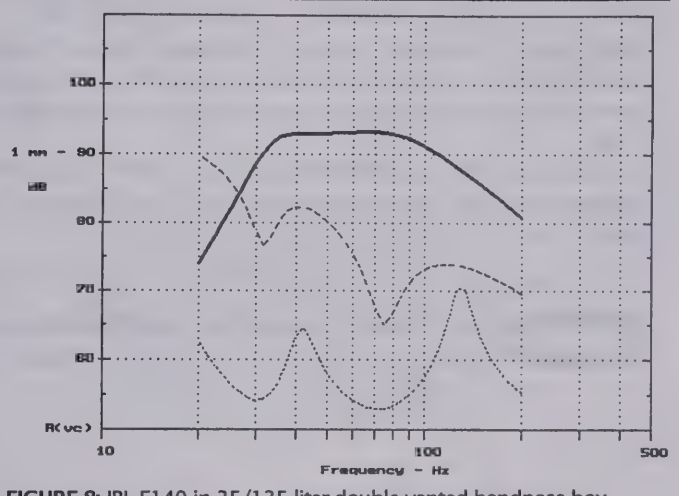

FIGURE 8: JBL E140 in 35/135 liter double-vented bandpass box.

quency band edge, we can divide the box into 65- and 185-liter subchambers, a ratio of roughly 3:1. Keeping the lower vent resonance at 35Hz, you can then tune the smaller chamber an octave higher. *Figure 3* shows the predicted result...not great, but not bad for starters. A little more twiddling with computer models suggests that a better subchamber ratio is 70 liters and 180 liters, which delivers a nearly symmetrical bandpass curve about 1.5 octaves wide centered near 70Hz (*Fig. 4*).

Not satisfied with these figures, Mr. Federoff wished a super-punchy system with response maybe an octave wide centered at about 90Hz. His starting point was the single-tuned, sealed-vented bandpass system analyzed by Margerand. With a 65-liter sealed chamber and a 105-liter vented chamber tuned to 90Hz, such a system delivers the performance of *Fig. 5*. Hmm, the only real difference between this and *Fig. 1* is reduced cone amplitude below 60Hz because of the clamping effect of the second vent.

THE DESIGN IN QUESTION

The Federoff "Birdhouse" functions as a hybrid system, with one port delivering useful sound and the other controlling out-of-band cone excursions. In other words, an original design ideally suited to the author's requirements.

His system has another interesting property. The computer model predicts that somewhat broader bandwidth with correspondingly less efficiency will result from simply reversing the tuning frequencies, as in *Fig. 6*.

Although Federoff's design offers the advantage of controlled cone excursion, anyone wishing to build such a system is advised to omit the second vent for two reasons: First, you can use Margerand's published tuning alignments (*SB* 6/88 and 1/89, "The Third Dimension: Symmetrically Loaded, Parts I and II"); and second, the sealed-vented bandpass configuration is in the public domain. A double-vented system, on the other hand, no matter how the vents are tuned, may be an

About the Figures

Acoustic output, cone excursion, and relative impedance are plotted logarithmically on each graph.

The dark, heavy curve represents one-watt, one-meter response with standard (hemispheric) acoustic loading.

The dotted curve near the bottom of each graph is impedance relative to voice-coil resistance. For example, a point 6dB above the bottom of the graph is twice the DC resistance of the voice coil, a point 10dB above the bottom is 3.15 times DC resistance, and so on.

The dashed curve shows one-watt cone excursion. 90dB is equivalent to one millimeter peak excursion, or 2 mm peak-to-peak. A point at 82dB is therefore 6dB below 1 mm, or 0.5 mm.

ABOUT THE AUTHOR

George L. Augspurger is the author of numerous articles and technical papers. He also contributes regularly to the *Patent Reviews* published by JASA and reprinted by JAES. Mr. Augspurger is a member of the Acoustical Society of America, United States Institute for Theatre Technology, and a fellow of the Audio Engineering Society. Formerly with James B. Lansing Sound, Inc., Mr. Augspurger devotes his full time to Perception Inc., a consulting office specializing in architectural acoustics and sound system design.

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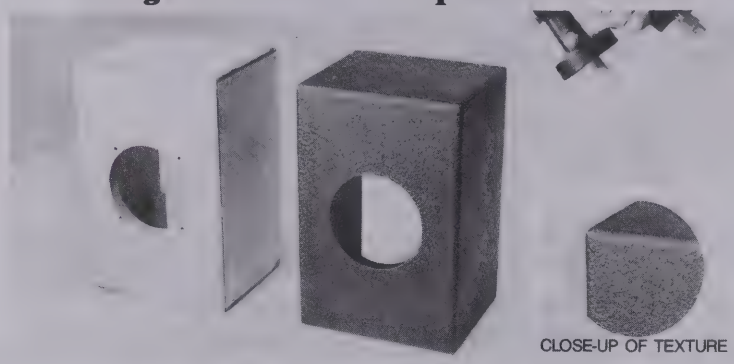

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Reader Service #33

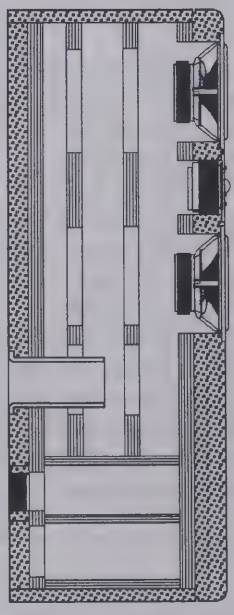
infringement of the Bose patent (Page 9 for a related editorial.)

A HAPPY WOOFER

For hi-fi or home theater applications, the low-frequency cutoff should be lowered still farther to about 35Hz. To keep the PS-15C happy, a larger box is required. *Figure 7* shows a 250-liter hybrid system with an 80-liter chamber tuned to 70Hz and a 170-liter chamber tuned to 15Hz. Again, the lower vent resonance only affects cone excursion and can be omitted.

As a useful comparison, *Fig. 8* shows the theoretical performance of a "Bose-type" double-vented system operating over roughly the same range from 35–100Hz. A more efficient loudspeaker such as a JBL E140 can deliver similar bandwidth and efficiency in a 170-liter box because both vents contribute useful output. The situation is analogous to the efficiency difference between corresponding sealed and vented box alignments.

As Matt Federoff emphasizes, if you want to play with bandpass response analysis you need a good computer modeling program. However, by using the information in his manuscript and then interpolating between the systems described here, you should achieve a variety of useful bandpass outputs from the PS-15C woofer.



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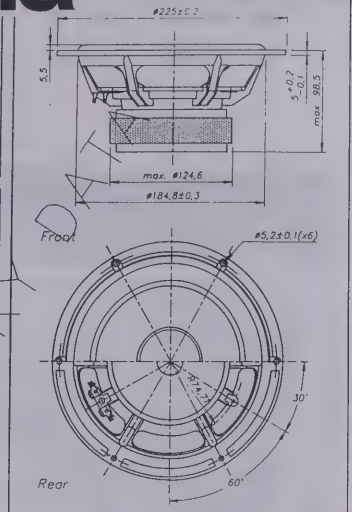
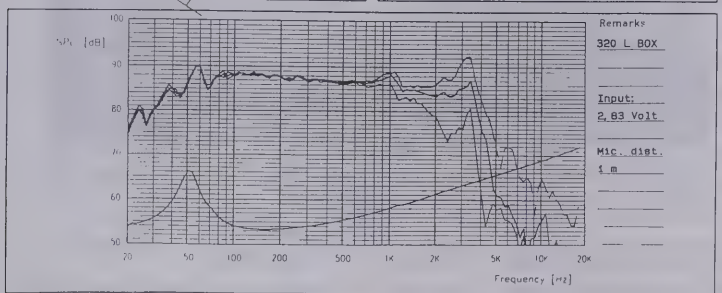
Effective Cone Area [sq.cm]	224
Voice Coil Resistance [ohm]	3,2
Voice Coil Inductance [mH]	0,9
Operating Power [W]	12,6
Voice Coil Diameter/Height [mm]	50/19
Air Gap Height [mm]	8
Moving Mass (incl. Air) [g]	35
Free Air Resonance [Hz]	48
Force Factor, B*l [T*m]	6,4
Magnet Weight [g]/[oz]	1062/37,4
Qms	3,44
Qes	0,87
Qts	0,70
Vas [ltr]	20





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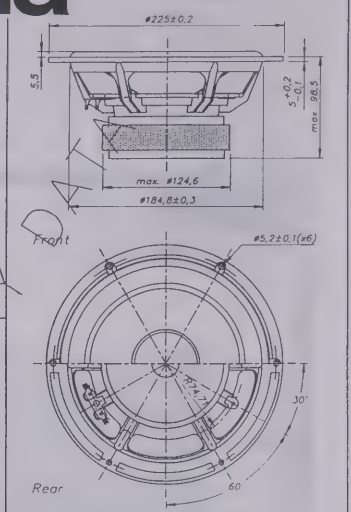
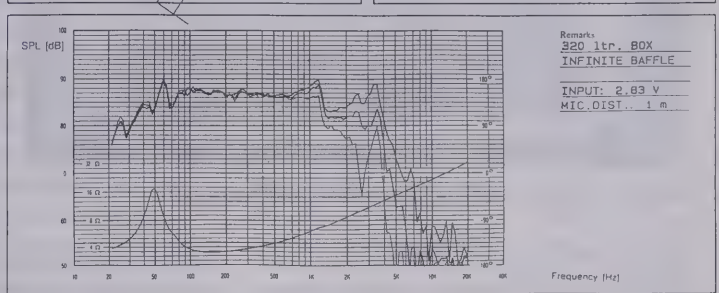
Nominal Power (IEC 268-5) [W]	150
Music Power (DIN 45500) [W]	200
Frequency Range [Hz]	45-500
Sensitivity (1W,1m)/(2,83V,1m) [dB]	84/87
Effective Cone Area [sq.cm]	224
Voice Coil Resistance [ohm]	3,1
Voice Coil Inductance [mH]	0,85
Operating Power [W]	16
Voice Coil Diameter/Height [mm]	50/19
Air Gap Height [mm]	8
Moving Mass (incl. Air) [g]	42
Free Air Resonance [Hz]	47
Force Factor, B*l [T*m]	6,9
Magnet Weight [g]/[oz]	1062/37,4
Qms	3,70
Qes	0,81
Qts	0,66
Vas [ltr]	19





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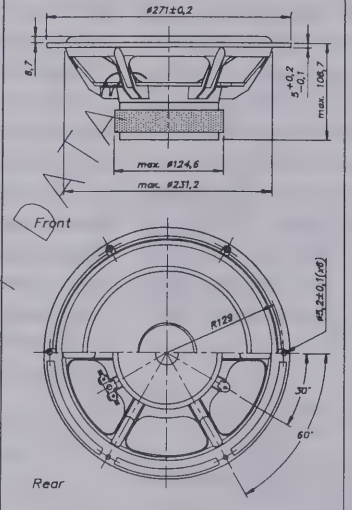
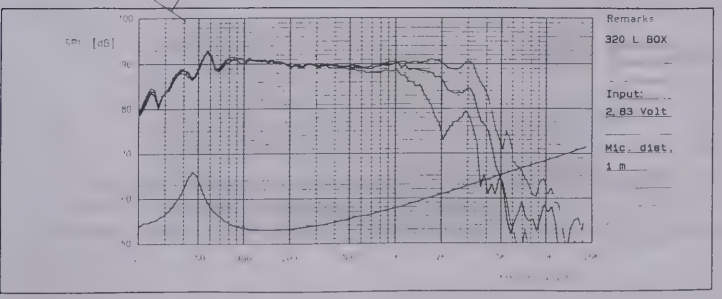
Nominal Impedance [ohm]	4
Nominal Power (IEC 268-5) [W]	150
Music Power (DIN 45500) [W]	200
Frequency Range [Hz]	40-1000
Sensitivity (1W,1m)/(2,83V,1m) [dB]	87/90
Effective Cone Area [sq.cm]	337
Voice Coil Resistance [ohm]	3,2
Voice Coil Inductance [mH]	0,95
Operating Power [W]	8,1
Voice Coil Diameter/Height [mm]	50/19
Air Gap Height [mm]	8
Moving Mass (incl. Air) [g]	41
Free Air Resonance [Hz]	44
Force Factor, B*l [T*m]	6,4
Magnet Weight [g]/[oz]	1062/37,4
Qms	3,65
Qes	0,84
Qts	0,69
Vas [ltr]	53





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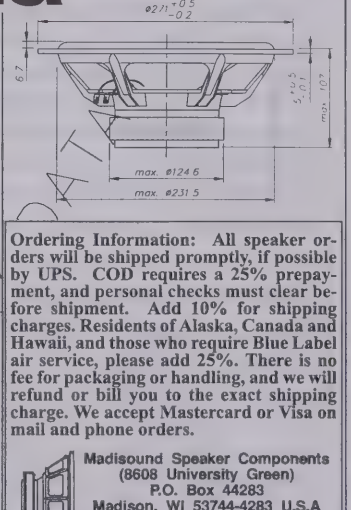
Nominal Impedance [ohm]	4
Nominal Power (IEC 268-5) [W]	150
Music Power (DIN 45500) [W]	200
Frequency Range [Hz]	40-500
Sensitivity (1W,1m)/(2,83V,1m) [dB]	86/89
Effective Cone Area [sq.cm]	337
Voice Coil Resistance [ohm]	3,1
Voice Coil Inductance [mH]	0,88
Operating Power [W]	10
Voice Coil Diameter/Height [mm]	50/19
Air Gap Height [mm]	8
Moving Mass (incl. Air) [g]	50
Free Air Resonance [Hz]	40
Force Factor, B*l [T*m]	6,9
Magnet Weight [g]/[oz]	1062/37,4
Qms	4,08
Qes	0,80
Qts	0,67
Vas [ltr]	55



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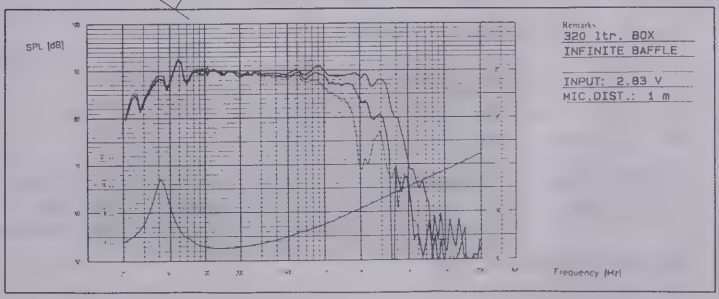

PRELIMINARY DATA

A 16HZ SUBWOOFER

By Charles T. Pike

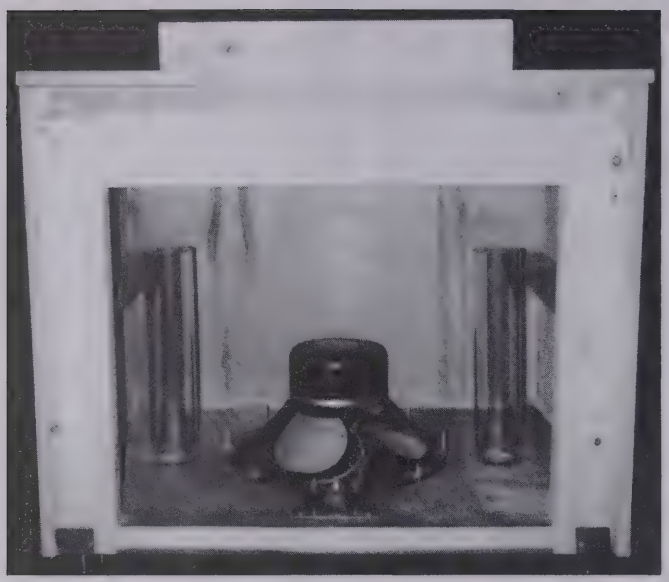

PHOTO 1: Loudspeaker box with back panel removed.

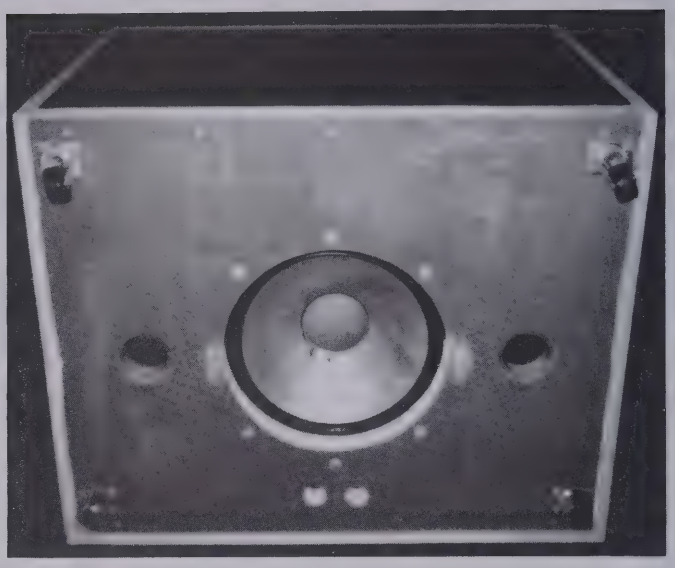

PHOTO 2: Bottom view of enclosure (mounted on casters) showing the placement of two ducts in relation to driver.

Ever since I became interested in audio, I've wished to build a speaker system to realistically reproduce the lowest notes (16Hz frequency) of the 32-foot stop of a pipe organ. Using loudspeaker drivers mounted in an infinite baffle or sealed enclosure, you would require too many 15″ or 18″ drivers capable of large excursions, however, to achieve flat response to 16Hz at realistic volume levels. I needed an alternative to this conventional, but expensive and bulky, approach.

If you tune a vented enclosure to the lowest frequency to be reproduced, the port will radiate most of the sound, and cone excursion will be greatly reduced. The price you'll pay for arbitrarily choosing the box tuning frequency is uneven frequency response.

Figure 1 shows a 15″ woofer's frequency response (tuned to 13Hz and calculated with the QUICK BOX computer program) in a 10 ft^3 vented enclosure. The frequency response rolls off at about 3dB/octave below 300Hz. Since this rolloff can be electronically corrected, however, the maximum SPL produced is more important. *Figure 2* shows the maximum SPL calculated using the BOXRESPONSE computer program and predicts that an SPL of 100dB can be produced at 16Hz (which is adequate for satisfying reproduction).

A SOLUTION UNFOLDS

I used a 15″ Goodmans Audiom 955 woofer, which I purchased 30 years ago. At the end of the article I suggest a currently available alternative driver which should work as well or better than the Goodmans. *Figure 3* depicts an impedance curve of the Goodmans driver, while *Table 1* lists the Thiele/Small parameters for this woofer calculated using the formulas in Vance Dickason's *The Loudspeaker Design Cookbook (LDC)*.[1]

The V_{AS} was calculated from the compliance, which I determined by measuring the displacement of a small weight on the speaker cone. The measured frequency of the 32Hz impedance peak in the final enclosure and the BOXRESPONSE calculation confirmed the accuracy of these numbers. The speaker's free

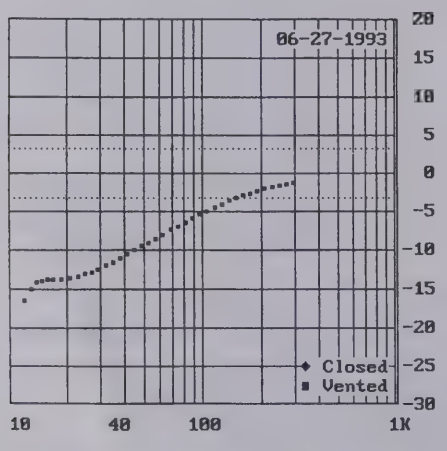

FIGURE 1: Frequency response of 15″ woofer in 10 ft^3 box.

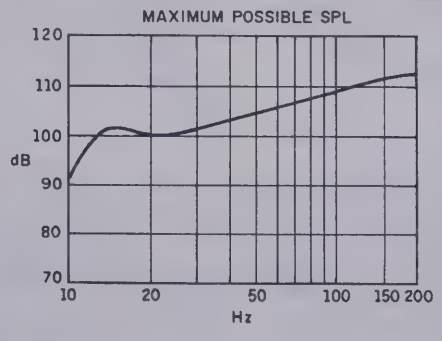

FIGURE 2: Maximum SPL produced by 15″ woofer in 10 ft^3 box.

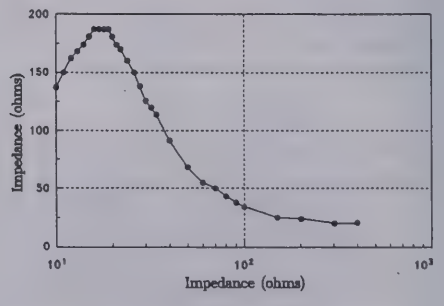

FIGURE 3: Impedance curve for Goodman's 15″ woofer.

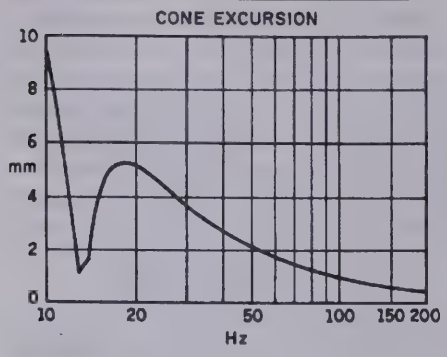

FIGURE 4: Cone excursion for 15" woofer.

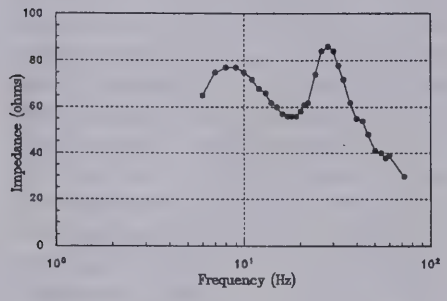

FIGURE 5: Impedance curve for 15" woofer in lossy box.

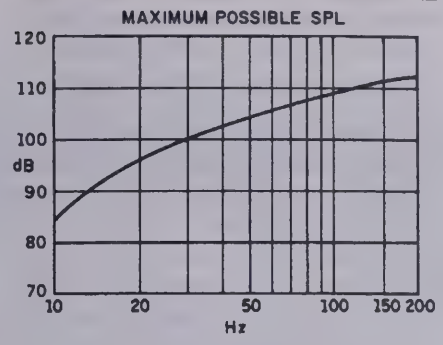

FIGURE 6: Effect of losses on maximum SPL.

air resonant frequency was originally 25Hz, but the suspension has loosened over the years and reduced it to the present value of 16Hz.

I used these parameters to determine the results of *Figs. 1* and *2*. In *Fig. 4*, which shows the BOXRESPONSE cone excursion results, the peak cone excursion of about 5mm is well within the limits of the cone suspension.

I originally mounted the woofer in a cabinet formed by enclosing the space under the stairs leading to my family room. After I installed the woofer, I measured the impedance as a function of frequency, but no peak due to the vent tuning was evident. Air leaks severely reduced the Ql of the cabinet. I crawled inside the cabinet with a caulking gun and a roll of duct tape to seal every crack and opening I could reach. *Figure 5* shows the improvement.

Using the formulas in the *LDC* and the Mathcad computer program, I calculated the value of Ql as a disappointing 1.4. *Figure 6* shows the maximum SPL reduced by about 9dB at 16Hz from the value in *Fig. 2*, which was calculated assuming a combined Q of 33. I realized that the only way to get good performance was to build a separate enclosure for the subwoofer.

PARTS LIST

SUBWOOFER ELECTRONIC CROSSOVER

REF	DESCRIPTION
Capacitors	
C1–C8	100N 5%, 50V film
C9	2,200µF, 35V aluminum electrolytic
Resistors	
R1, 4, 5, 10, 13	82.5k 1% ¼W metal film resistor
R2, 11	7.5k 1% ¼W metal film resistor
R3, 6, 7, 12, 15, 16	39.2k 1% ¼W metal film resistor
R8, 17	49.9Ω 1% ¼W metal film resistor
R9, 18	1k 1% ¼W metal film resistor
R19	1.6k 5% 1W carbon
R20	100k 1% ¼W metal film resistor
Semiconductors	
U1	TL074CN
U2	TIP120
D1	1N4148
D2	1N4001
Z1	1N5338
Miscellaneous	
K1A,B,C	Relay, 2PDT, 24V coil (see text)
	Perforated circuit board
	Gold-plated phono connectors
	±15V regulated power supply
	Chassis box

THE PLAN

To achieve a high Ql, your loudspeaker enclosure must be free of air leaks, the walls

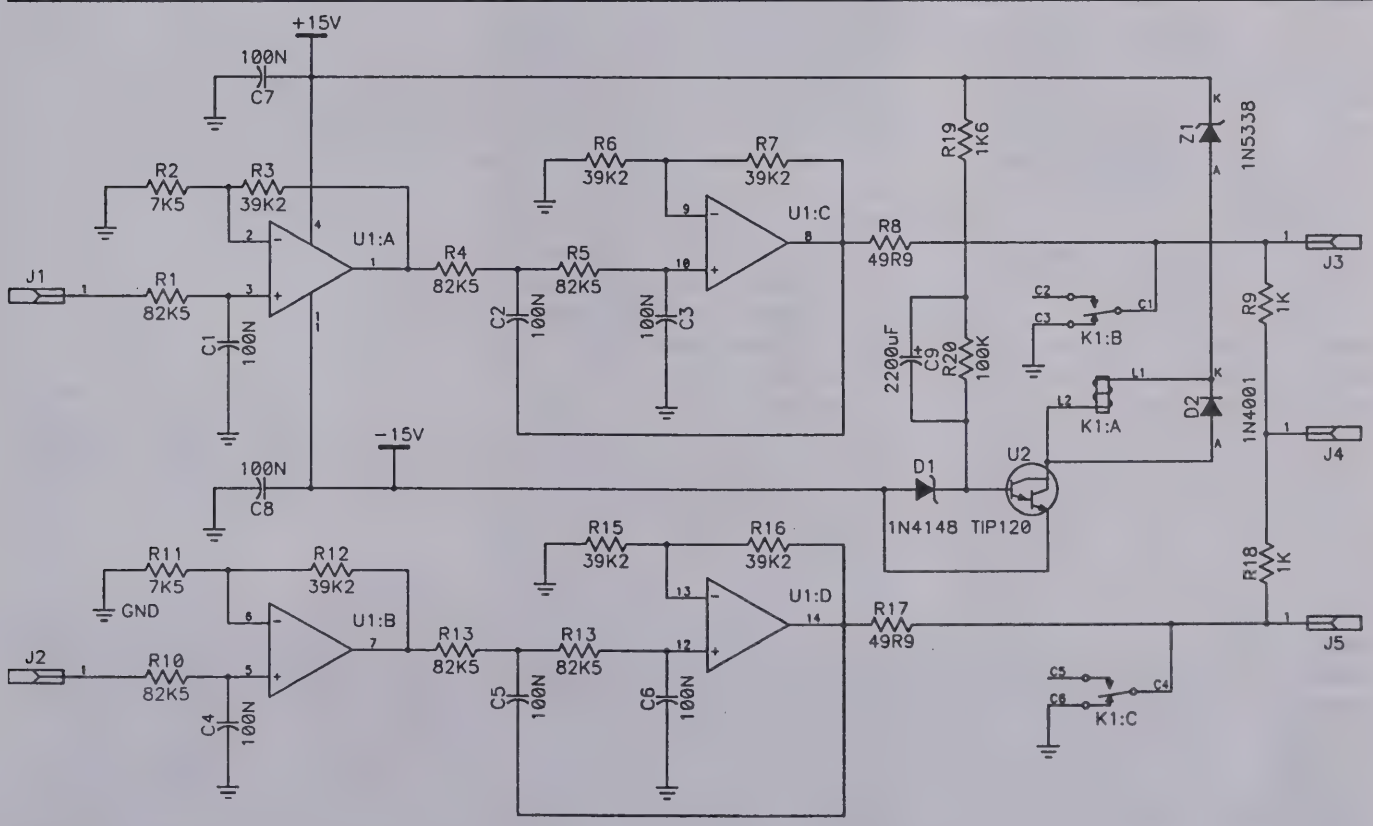

FIGURE 7: Schematic of electronic crossover.

must be rigid, and no damping material should be used. The lowest major box resonance occurs at a frequency with a wavelength twice the longest box dimension. In my design the lowest resonance happens at about 200Hz, which is well above the recommended crossover frequency. Therefore, I don't need to include damping material to control cabinet resonances.

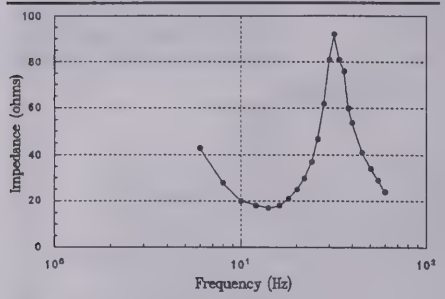

FIGURE 8: Impedance of final system.

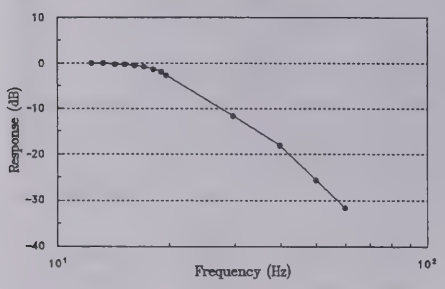

FIGURE 9: Frequency response of electronic crossover.

Using the "alignment" suggested in this article, you will reduce system efficiency when you decrease the volume of the enclosure. Also, you will require a longer tuning duct, which will be more difficult to fit inside the box. I chose a 10 ft^3 volume as a compromise between efficiency and size. The overall outside dimensions of the cabinet are 33.5" long by 27.5" deep by 26.5" high—small enough to fit through a normal doorway and, since it is mounted on 2" casters, easy to move. The casters space the bottom of the cabinet 2.5" from the floor, making the total height 29"—a convenient level for use as a table.

To ensure rigidity, I constructed the top, bottom, and front of the cabinet from a 1.75" thick exterior or "solid core" door, which had served as our kitchen table for many years. The other three walls are constructed of 0.75" particleboard braced with an edge-mounted 2 × 3 across the middle of the ends and a similarly mounted 2 × 4 on the back. Since the door's core is a sawdust and glue mixture, which has a porous surface, I used RTV silicone rubber to cement all joints for air tightness. Also, to get a good grip in the door, I used 3.5" deck screws to hold the joints together.

Use 2 × 4s to frame the opening for the removable back panel as shown in *Photo 1*. The back panel on top of the enclosure in *Photo 1* shows the position of the 2 × 4 stiffening cleat. I centered the driver face down in the cabinet's bottom panel to take advantage of room boundary coupling. You can do this without worrying about diffraction because the frequencies the loudspeaker is handling are larger than the thickness of the panel. At 40Hz the sound wavelength (at sea level) is about 28', or almost 200 times the panel thickness.

To tune the cabinet to 13Hz, I installed two ducts—with an inside diameter of 2.75" and a 0.25" wall thickness—on each side of the driver (*Photo 2*). I rounded the ends of each 16"-long plastic tube to a radius equal to the tube's wall thickness. The subwoofer is free of any "chuffing" noises even when driven at high levels at subaudible frequencies.

The input connectors are Superior Electric Co., 50 amp Supercon connectors, which some readers may consider overkill, but unlikely to ever cause any problems. I finished the enclosure by covering it with charcoal gray cabinet carpet cemented in place with a multipurpose floor covering adhesive.

A HANDY SAFETY PRECAUTION

I designed the electronic crossover (*Fig. 7*) with the aid of the *Active-Filter Cookbook*.[2] It is an 18dB/octave low-pass filter, which is 3dB down at 20Hz. Since the first op amp stage in most designs serves as a buffer for a passive 6dB/octave filter section, you can easily modify this design by adding R2, R3, R11, and R12 to bring the subwoofer output up to the level of other loudspeakers. I've provided two inputs, J1 and J2, for the right and left channel stereo signals. The corresponding outputs for driving a pair of subwoofers are J3 and J5. The outputs are summed using the 1k resistors, R9 and R18, and connected to J4 for driving a single subwoofer.

Since operational amplifiers' output can fluctuate widely on turnoff, I have provided a simple protection circuit to prevent damaging your amplifiers or speakers. When you turn on the power, the Darlington transistor, U2, turns on by the current through the 1.6k resistor, R19, and the capacitor, C7. When the voltage is high enough, relay K1 is pulled in and the outputs are active. Capacitor C7 then charges up almost to the 30V difference between the supply rails, and U2 stays turned on by the current through the 100k resistor, R20.

When you turn off the power, the rail voltage drops, and U2 turns off almost immediately by C7, which has a time constant of 220s in parallel with R20. Relay K1 then drops out and the outputs are shorted. I recommend, however, that you turn on the crossover before the subwoofer power amplifier and turn it off after the power amplifier is shut off.

Because of the protection circuit's sensitivity to voltage fluctuations, the power supply should be regulated. I have not included a schematic of the ±15V power supply, since many suitable designs have been presented in *TAA* and other resources. The electronic crossover should be preceded by a buffer with an output impedance of about 1k or less to avoid affecting the electronic crossover performance.

CROSSOVER CONSTRUCTION

I built the electronic crossover on a Radio Shack or Vector predrilled board with solder pads for mounting components. The latest version uses ¼W 1% metal film resistors for all resistors except R19, which is a 1W 5% carbon resistor. The capacitors, C1 to C8, are Panasonic P-series 2% tolerance polypropylene capacitors, and C9 is a 35V electrolytic. The quad op amp, U1, is a Texas Instruments TL074CN. The relay, K1, is 4PDT with gold-plated contacts and a 24V DC coil with a resistance of 650Ω (many manufacturers offer similar units). The input and output connectors are also gold-plated to avoid oxidation. The straightforward layout should present no problem for experienced amateurs.

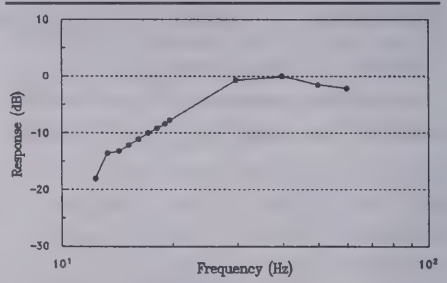

FIGURE 10: Frequency response of final system without electronic crossover.

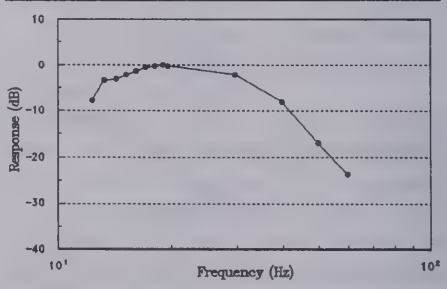

FIGURE 11: Frequency response of final system with electronic crossover.

TESTING

I used the following test equipment: a modified Hewlett-Packard Model 200 audio oscillator calibrated with a Continental Specialties Model 5001 counter-timer in the period mode; the AC millivoltmeter section of a Heathkit Model IM-22 Audio Analyzer; a Triplett Model 650-SC AC voltmeter, which was calibrated at 60Hz by a Keithley Model 130 digital multimeter; a Radio Shack sound level meter (cat. no. 33-

Reader Service #45

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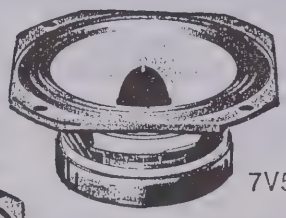
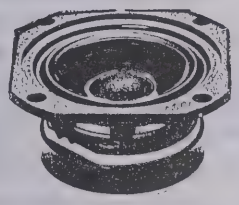
- **4" FOCAL midbass:** 4C212, cast frame, semi exponential coated paper cone, negative rubber surround, sharp impulse here also, and real muscle too! Will work in 2-way + sub. Nice roll-off around 2-3 KHz.
- **6" FOCAL midbass:** 60211S, new low profile 6" frame, latex damped paper cone, Neoprene surround, 4 layers VC. Ideal for potent 2-way, easy to cross over.
- **7" FOCAL High efficiency midrange/midbass:** 7V513, cast frame, Polyglass cone with phase plug, coated foam surround. Edge-wound flat copper wire VC on Kapton former, massive magnet structure. An explosive high power midrange, more than 93 dB/W/m. Detailed and elegant it is the rare find for tube amp. officionados!

4C111

6C211S



7V513



4C212

T120ZLT

	4C111	4C212	6C211S	7V513	T120ZLT
DCR	6.16	6.78	3.10	6.83	6.0
Fs	81.72	72.80	41.37	46.81	–
Qes	0.482	0.58	0.389	0.250	–
Qms	2.851	2.33	6.114	2.91	–
Qts	0.412	0.47	0.362	0.28	–
Vas (L)	4.14	4.82	29.92	33.81	–
L (mH)	0.24	0.48	0.5	0.55	–
dB/W/m	88.56	86.8	89.2	93.25	93.5
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TABLE 1

T/S PARAMETERS FOR GOODMANS AUDIOM 955 WOOFER

PARAMETER	VALUE	UNIT
f_S	16	Hz
V_{AS}	900	L
R_E	15	Ω
Q_E	0.1	
Q_M	1.2	
Q_{TS}	0.09	
Cone Area	730	cm^3
Efficiency	97.5	dB/W
Power Rating	50	W

2050); and a Panasonic WM-063T microphone cartridge in a home-built microphone.[3] (Panasonic has since replaced the WM-063T with the WM-60AT.)

Figure 8 illustrates an impedance curve of the completed system. The 92Ω impedance peak at 32Hz and the 17Ω null (which approaches the driver's 15Ω DC resistance) at the box tuning frequency of 13Hz indicate that the Ql of the cabinet is high. Using BOXRESPONSE and inserting different values until I matched the measured impedance of 17Ω at the box resonance, I estimated the value of Ql to be 20. The formulas in *The Loudspeaker Design Cookbook* produced a value of about 15, so I was able to confirm my results. *Figure 9* is the frequency response of the electronic crossover, which agrees with the design predictions.

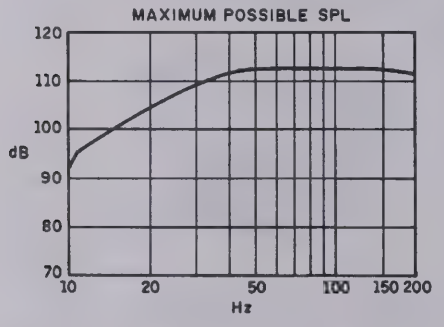

FIGURE 12: Maximum SPL for Radio Shack woofer in 10 ft^3 enclosure.

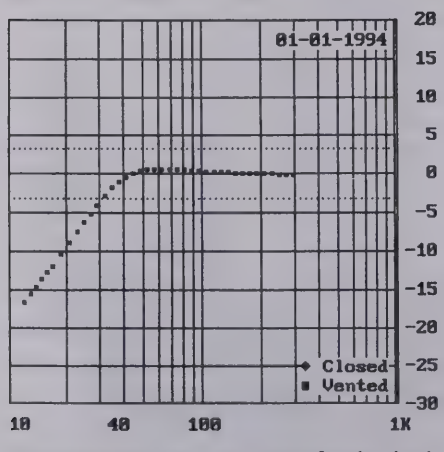

FIGURE 13: Frequency response of Radio Shack woofer in 10 ft^3 enclosure calculated by QUICK BOX.

I measured the frequency response with both the home-built microphones using the Panasonic WM-063T cartridge and the Radio Shack sound level meter. The speaker input voltage was 5.7V RMS maximum, which is equivalent to 2W RMS into the loudspeaker's nominal 16Ω impedance. I placed the microphones on the floor under the center of the speaker system. The Radio Shack meter response rolled off below 20Hz relative to the Panasonic cartridge.

After examining the Radio Shack meter schematic, I noted that a 1μF capacitor is used for interstage coupling to the 10kΩ resistance of the attenuator. This results in a high-pass filter characteristic, which is 3dB down at 16Hz and explains the difference in measurements.

Therefore, my measurements using the Panasonic cartridge are plotted in *Fig. 10*, which shows the system's frequency response without the electronic crossover. Although the slope of the rolloff is steeper than predicted, the total 13dB drop in output is about the same. When combined with the crossover response, the system is 3dB down at 15Hz and 35Hz, as shown in *Fig. 11*.

FEEL THE BEAT

While some may argue that it is impossible to reproduce frequencies with a wavelength longer than twice the largest room dimension, I have not found this to be true. Room modes will not enhance the output at these frequencies, but reproduction is still possible. During testing I found a 14Hz resonance in my house's structure, which caused the walls to shake and an outside door to rattle. Since the floor of my family room is a concrete slab, the energy which caused the walls to vibrate had to be acoustically coupled.

My listening material for testing included the CDs on the Dorian label by J. Guillou and on Telarc by M. Murray, and the recording of Saint-Saens' Symphony No. 3 on Philips #412 619-2 conducted by Edo de Waart with J. Guillou as organist. The subwoofer extended my system response to the lowest CD frequencies with room-shaking authority. No audible evidence of distortion was apparent; just clean, pure bass that I could feel as well as hear. The subwoofer also enhanced my enjoyment of laserdisc films such as "Last of the Mohicans," "The Empire Strikes Back," and "The Adventures of Baron Munchhausen."

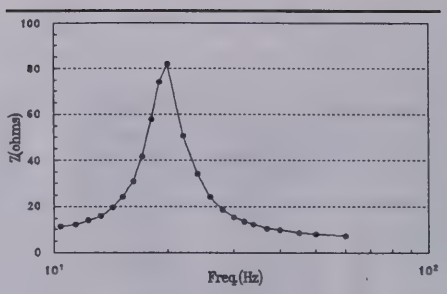

FIGURE 14: Measured impedance of Radio Shack #40-1301 woofer without enclosure.

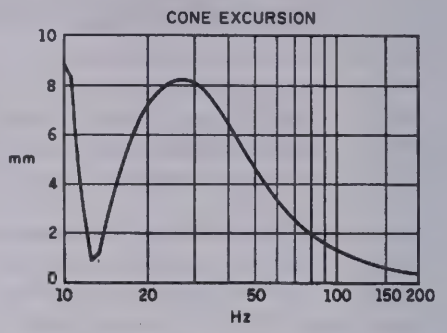

FIGURE 15: Peak cone excursion required by Radio Shack woofer calculated by BOXRESPONSE.

AN EXCELLENT ALTERNATE

Based on availability and the specifications listed in *Table 2*, I selected the Radio Shack 15″ woofer catalog no. 40-1301 to test as an alternate to the Goodmans driver. I was particularly impressed with the X_{MAX} of 18mm (peak-to-peak), which is one of the highest values I have encountered. The maximum SPL this woofer generates using the suggested approach is approximately the same as the Goodmans woofer as calculated with BOXRESPONSE (*Fig. 12*). As shown in *Fig. 13*, the QUICK BOX-calculated frequency response—and, consequently, the perform-

TABLE 2

RS 40-1301 15″ WOOFER SPECIFICATIONS

PARAMETER	DATA SHEET	MEASURED	UNITS
f_S	20	20	Hz
V_{AS}	594		
R_E	5.6	5.6	Ω
Q_E	0.485	0.45	
Q_M	6.45	6.1	
Q_{TS}	0.45	0.42	
X_{MAX}(peak-to-peak)	18		mm
Cone Area	856	856	cm^3
Efficiency	90	92 (calculated)	dB/W
Power Rating	100		W

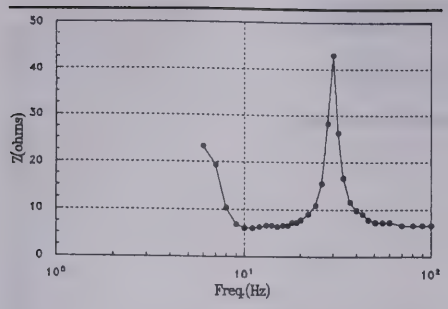

FIGURE 16: Measured impedance of Radio Shack woofer in subwoofer enclosure.

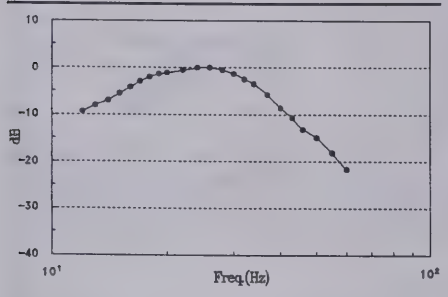

FIGURE 17: Measured frequency response of Radio Shack woofer in subwoofer enclosure.

ance—is similar to the measured response for the Goodmans woofer.

After purchasing the woofer, I measured the impedance as a function of frequency, as shown in *Fig. 14*. I then calculated the Q_{MS} and Q_{ES} and present them in *Table 2*. The similarity between my measured values and the Radio Shack values is excellent and probably within the experimental error of my measurements. Radio Shack has done an excellent job of quality control on this unit. *Figure 15* shows that the required peak excursion is less than the 9 mm peak excursion limit of the Radio Shack woofer.

Since the Radio Shack woofer has a slightly larger cone than the Goodmans woofer and requires a 14″ baffle hole rather than 13″, I fabricated an adapter flange with a 14″ hole from ¾″ plywood. *Figure 16* depicts an impedance curve of the subwoofer with the Radio Shack woofer installed. By substituting values into the BOXRESPONSE until I matched the 43Ω impedance peak, I estimated a Q of 8 for the enclosure. Although the Q is lower than with the Goodmans woofer, the maximum SPL of 100dB at 16Hz matches.

I measured the frequency response, shown in *Fig. 17*, using the same conditions and equipment as used for the Goodmans woofer. At 16Hz the response is down 4dB, which is excellent performance considering that the Radio Shack woofer costs less than what I paid for the secondhand Goodmans woofer 30 years ago.

CONCLUSION

I have presented a general approach to the design of subwoofers capable of reproducing the lowest octave of the audio spectrum at realistic levels in a home environment. The design includes an enclosure tuned to the lowest frequency to be reproduced and an electronic crossover to flatten the frequency response. Although I refer to two specific design examples, you could use this approach to extend the bass response of a variety of drivers.

SOURCES

Mathsoft, Inc.
201 Broadway
Cambridge, MA 02139
Mathcad

Old Colony Sound Lab
PO Box 243
Peterborough, NH 03458
QUICK BOX, *BOXRESPONSE*, The Loudspeaker Design Cookbook, Active-Filter Cookbook, *Mitey Mike test microphone*

REFERENCES

1. V. Dickason, *The Loudspeaker Design Cookbook*, (The Marshall Jones Co., 1987; Audio Amateur Press, 1991).

2. D. Lancaster, *Active-Filter Cookbook*, (Howard W. Sams & Co., 1975).

3. J. D'Appolito, "Mitey Mike: For Loudspeaker Testing," *SB* (6/90): 10ff.

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Reader Service #27

Wayland's Wood World

BISCUIT JOINTS

By Bob Wayland

All too often, we try to make screws correct for poor craftsmanship and planning when we make enclosures. Shoulders aching and hands sore, we leave unsightly screw heads and plugged holes, most evident in exposed corner joints. As I mentioned in my column on power tools (*SB* 2/93, p. 54), biscuit joiners are a wonderful and easy way to make joints. Let's explore this topic in a bit more detail.

A biscuit joiner is essentially a small motor with a right-angle drive coupled to a 4" saw blade, which makes a shallow cut on each side of the pieces to be joined. This cut accepts a flat, football-shaped biscuit of compressed birch. When the biscuit is inserted into the glue-filled cut, it quickly swells to form an incredibly strong bond. If you start gluing soon after making the biscuit cut, your joint be will stronger ("Sticking Together," *SB* 2/94, p. 46). You can easily make adjustments for perfect alignment before the glue sets.

When purchasing a biscuit joiner, consider one of the newer designs featuring a rugged front fence that can be accurately aligned and will hold your settings without the need of extra clamps. Expect to pay about $200. A useful additional feature is a dust/chip catcher. These little devils can produce prodigious quantities of chips. A number of woodworker supply houses offer a 12-tooth blade to replace the standard 6-tooth ones supplied with the joiners. The claim is that you get a cleaner cut.

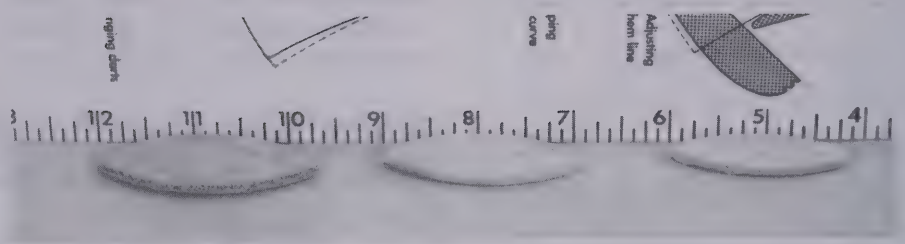

PHOTO 1: Standard-size biscuits for use with biscuit joinery.

The biscuits come in three different sizes (*Table 1*; *Photo 1*). If you're building several enclosures (or a large one), buy the biscuits in 1,000-piece lots. For the same price—about $20 plus S&H—you can buy a mixed bag of 250-#0, 250-#10 and 500-#20. (You can also buy exotic special-order biscuits at a premium, without gaining any appreciable advantage.)

All the biscuits fit into the standard 5/32-inch-wide slot cut by the joiner. If the relative humidity in your area is consistently high, store the biscuits in zip-lock bags to slow the absorption of moisture, and add activated charcoal for extra protection.

PHOTO 2: Adjusting the depth of cut on the biscuit joiner.

Before you start, set the depth of cut on the joiner. Then cut along the long dimension of

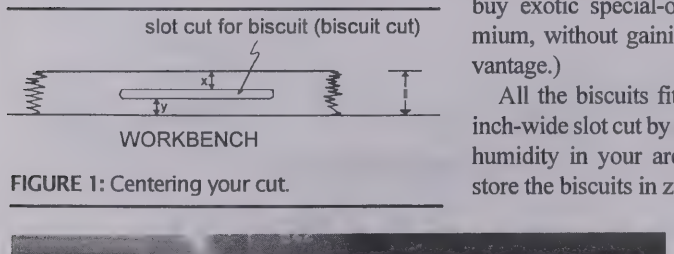

FIGURE 1: Centering your cut.

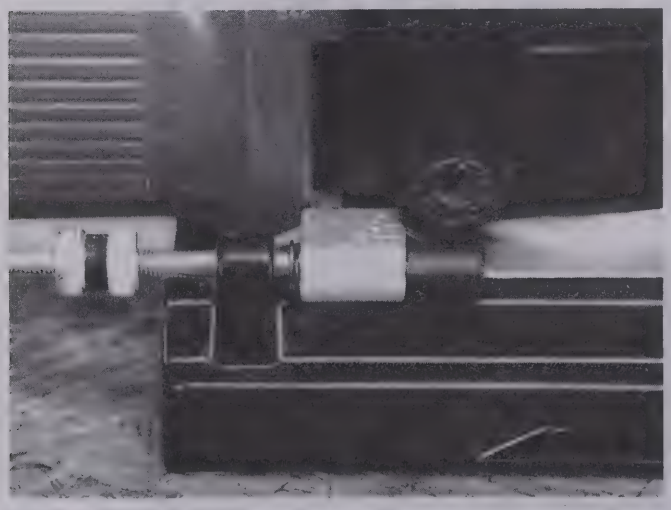

PHOTO 3: Biscuit-size adjustment on Freud JS100 Joiner.

PHOTO 4: Adjusting to center the slot cut. Notice the clamps holding the fence firmly in place.

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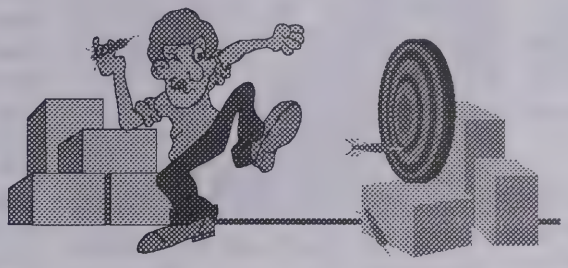

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Developed by Marc Bacon and published by Technologie **MDB** the LDIS is *the only speaker design course in the world that combines theory and hands-on practice in a structured series of topics.* It uses a combination of easy-flowing text, cassettes, FOCAL drivers, AXON crossover components, software, and experiments to teach the latest in speaker-building technology and sound reproduction techniques, yet requires only a basic knowledge of algebra as a prerequisite. Beginning in June 1994, the LDIS will be published eight times each year for a total of 24 units and will be revised every three years to ensure that it incorporates the industry's most recent developments. The writing style is aimed at beginners, yet the course contains a wide enough base of knowledge to satisfy the most seasoned audiophile. Course contents are given below:

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1	Music, Physics, and the Perception of Sound	13	Determining Driver Parameters
2	Dynamic Loudspeakers, part 1 -- Woofers	14	Objective Measurement of Loudspeakers
3	Dynamic Loudspeakers, part 2 -- Midranges and Tweeters	15	Subjective Measurement of Loudspeakers
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9	Cabinet Construction and Damping	21	Woodworking and Cabinetmaking
10	Mid and High-Frequency Enclosure Design	22	Advanced Speaker Design -- An introduction to active crossover theory
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12	Passive Crossovers and Response Shaping Circuits	24	Advanced Speaker Design -- Putting it all together. A rational design approach to achieving optimum performance in hobbyist or low volume production.

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Reader Service #18

PHOTO 5: Proper amount of glue for a biscuit slot.

a #20 biscuit so that exactly one-half is left. Using this test biscuit as a gauge, adjust the joiner's depth of cut just a bit deeper than the half biscuit (*Photo 2*). Most machines have a stepped index to allow switching from one biscuit size to another without resetting the depth each time (*Photo 3*). Make a test cut and insert the test biscuit to make sure it just fits.

A FEW GOOD JOINTS

This column covers the two most commonly used joints for speaker builders: the edge-to-edge and butt joints. Once you have mastered these techniques, you'll be able to handle even the most difficult biscuit joinery.

All biscuit joiners have an indexing system that allows you to center your cut precisely. This is usually a clearly defined mark on the housing, both top and bottom and on the fence. To quickly cut only the slots needed, simply align these marks with marks on your work. The cutting is made by pushing the cutting head into the edge of the material. Quick and simple!

First, decide on the biscuit spacing along the joint. Place a biscuit at each end with its center line about 2″ from the end, then space the remainder about every 6–8″ between the two. For normal cabinet construction, this is good practice. Since the pressure levels generated within an enclosure impose special requirements, however, I normally leave a biscuit length between each.

Usually, there is never a whole number of biscuits' lengths along any joint and you will need to expand or contract the spacing as appropriate. Using a try-square or a framing square, draw a line perpendicular to the glue line at the midpoint of each biscuit. Don't fret about where the midpoint is; the main thing is that the line is in the same place on both sides of an aligned joint.

EDGE-TO-EDGE JOINTS

Set the fence to place the biscuit slot on the center line of your board's thickness. Here is another of woodworking's little secrets: if the cut is always relative to the same surface,

PHOTO 6: Spreading the glue in a slot with a palette knife.

TABLE 1

BISCUIT SIZES

SIZE	DIMENSIONS (in inches)
#0	1¾ × 5/8
#10	2⅛ × ¾
#20	2⅜ × 1

centering won't have to be exactly "dead on." (Some woodworkers don't worry about centering the cut. I think this introduces unnecessary potential for stress buildup.) As a good approximation for centering your cut, consider *Fig. 1*. For the cut to be centered x = y and

$2x + 5/32 = t$ (thickness of board),
or for t = ¾ ″, x = 19/64 or about 5/16″.

Now adjust the fence so the position of the cut will be in the center of the board's thickness (*Photo 4*). Many biscuit joiners are made with the distance y fixed, however. For edge-to-edge joints, centering the cut is worth the effort (this isn't possible with butt joints). Support the joiner's base firmly with shims, to ensure getting the correct distance x for a centered cut. If your fence is strong enough, and will accurately remain in adjustment, let the fence support the joiner without shims.

As an alternative, set your fence for the thickness t with the base as one reference surface and the position of the fence determined by the board's thickness. Your slots will be off center, but you'll be OK as long as you follow the above simple rule. Narrower boards present a problem: the distance x should always be at least ¼″. Problems with swelling can also occur, with subsequent dimple formation if you sand or plane the surface before the swelling has subsided.

To ensure that everything goes together as you wish, dry clamp before beginning the gluing; never put glue on all surfaces before assembly. It is far simpler to make corrections before gluing.

First put glue into the slots on one of the boards and insert the biscuits (*Photo 5*), then spread the glue in the slot with an artist's palette knife (*Photo 6*). Special glue dispensers perform the whole operation in one step. Woodworkers often claim the only glue needed for the other piece is in the biscuit slots. This doesn't work for speaker enclosures—unless, of course, you want to produce a whistle. Put glue on the board face of the glue joint and in the biscuit slots of the other piece.

Next, quickly assemble the pieces and align them accurately before the glue starts to set (usually within five to ten minutes). The best guides will be the centerline guides you marked for the slots; be sure you have clamped properly (see "Putting On the Pressure," *SB* 1/94, p. 58). When gluing up a whole panel of boards, glue in pairs from the outside edges inward. The biscuits set up

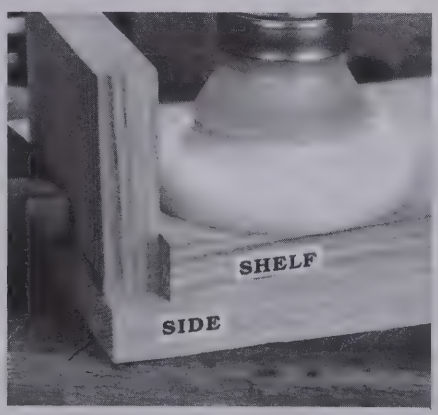

PHOTO 7: Aligning and clamping the pieces together: corner joint.

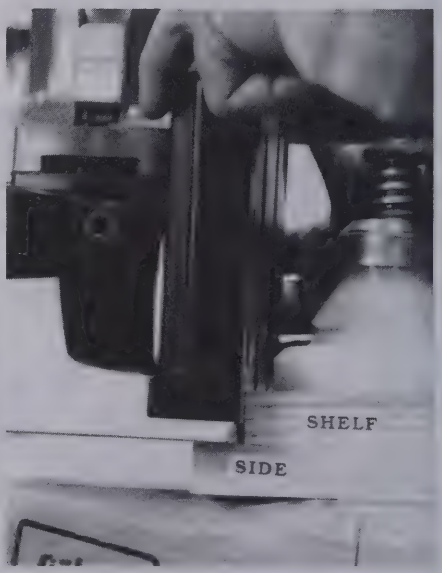

PHOTO 8: Making the vertical cut into the side piece: corner joint.

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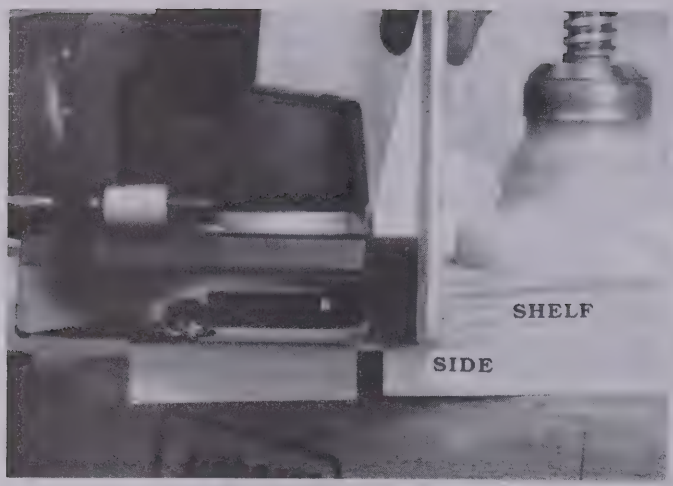


PHOTO 9: Making the horizontal cut into the shelf piece: corner joint. The extra board, which acts as a support for the joiner, should be of the same thickness as the other pieces.

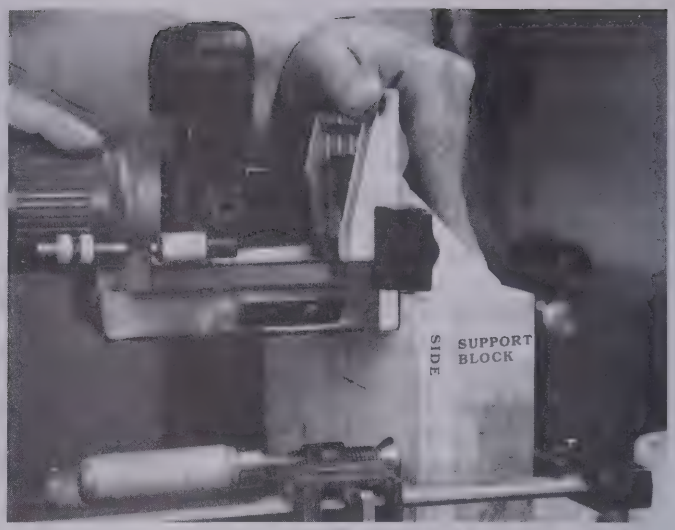


PHOTO 10: Using a support block to cut the slots into the side piece.

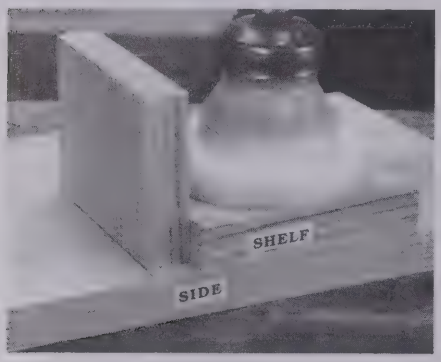


PHOTO 11: Aligning and clamping the pieces together: interior butt joint.

much faster than normal joints; with most glues, you can probably remove the clamps in about 10 to 15 minutes (see *SB* 2/94, p. 46). If time isn't critical, it's better to leave the clamps on overnight.

BUTT JOINTS

Though not very pretty, the butt joint is the simplest, and fine if your finish is to be veneer, texture, or paint. As the name implies, you just butt one piece against another, usually at right angles, either at a corner where the sides meet the top and bottom or as a shelf within the cabinet. The corner and interior butt joints are basically made the same way, with only small differences.

The corner butt joint requires the two pieces to be exactly the same thickness. With particleboard and medium density fiberboard (PB/MDF) this should not be a problem. First, set up the side and shelf so that the cuts can have the same configuration. Using a piece of material of the same thickness as the shelf and side, clamp the two together (*Photo 7*). You shouldn't be able to feel any offset where the spacer and side meet. Be sure the shelf piece is clamped flat to the side. At each of the marked locations, align the joiner in a vertical position and make the cut (*Photo 8*). Rotate the joiner to the horizontal position, align with the mark, and make the cut into the shelf (*Photo 9*). This procedure ensures that all cuts are made relative to the outside (exposed) faces.

Another method for making the side cut is to use a square support block (*Photo 10*). Be sure to set the fence, using the shelf as a guide.

An interior butt joint (for a shelf) is made almost the same way. Carefully mark the shelf's location, then clamp it to the side so it is flat and spaced its exact thickness below the reference line. Either use a piece of material of the same thickness as a guide (*Photo 11*), or align one edge with the reference line and then carefully rotate the shelf downward onto the side piece. Make the vertical and horizontal cuts as before (*Photos 12* and *13*).

Good joints which are not at right angles to each other are also possible. Unfortunately, most fences have fixed angles built in (usually 45° and 90°). If you wish to use these joiners, you must cut angled shims and attach them to the fence. If you expect to be making many non-right-angled joints, consider a biscuit joiner with an adjustable fence.

Biscuits having a self-clamping action used to be available, but lately I haven't been able to find a source for them. If any reader knows of one, please let me know by writing to me in care of *Speaker Builder*.

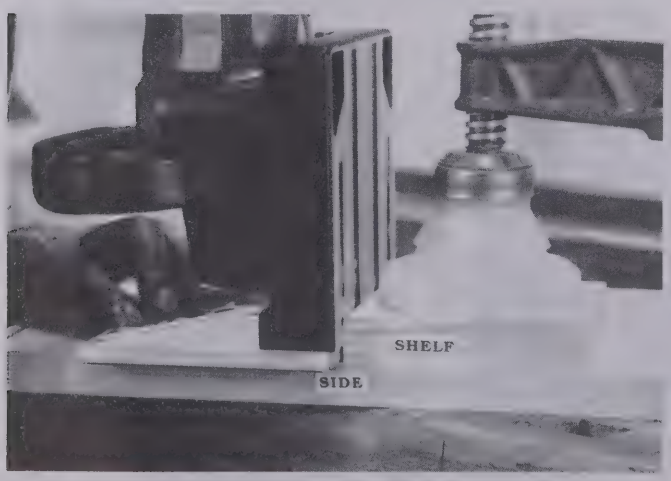


PHOTO 12: Making the vertical cut into the side piece: an interior butt joint.

PHOTO 13: Making the horizontal cut into the shelf piece: interior butt joint.

SYNCHRON.....Coincidental Loudspeakers

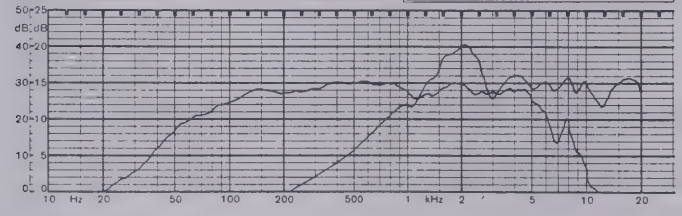
The dream of every loudspeaker designer is to build a speaker with a smooth frequency response and good imaging. A smooth frequency response can be obtained by careful crossover design and use of quality drivers. Good imaging poses a problem that is not so easily solved, for the sound of the woofer and the tweeter should reach the listener at exactly the same point in time. Phase correction circuitry can be added to the crossover in order to achieve better imaging, but this type of design creates a precise image only in small areas of the listening room. Now a much improved design technique is possible due to the development of neodymium tweeter magnets. Their small mass allows a tweeter to be mounted at the base of the woofer cone, very close to the woofer voice coil.

Close positioning of the driver voice coils creates true imaging for any listening position in the room, because the sound generated by the woofer and tweeter originates from a single point source. The listener hears the music synchronized, in phase, and the result is unmistakably realistic.

The woofers in Synchron loudspeakers feature cast baskets, mineral filled polypropylene cones, PVC surrounds, Kapton formers, and ferrofluid cooling. The tweeters have aluminum domes with supronyl surrounds (3/4" on the SYN-519A and SYN-619A, 1" on the SYN-825A), neodymium magnets, diffuser/dome protectors, Kapton formers and ferrofluid cooling.

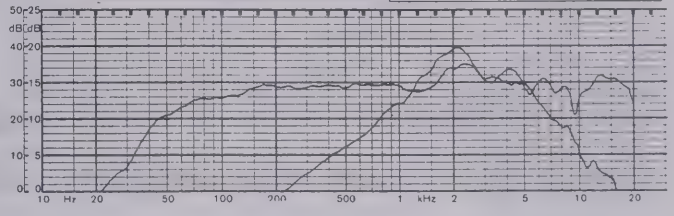
SYNCHRON SYN-519A 5.25"			
Technical Data	**Symbol**	**Value W/T**	**Unit**
Nominal Impedance	Z	8/8	Ω
Resonance Frequency	Fs	58/1500	Hz
Power Handling (IEC)	P	25/30	W
Sensitivity (1W/1m)	E	90/90	dB
Voice coil Diameter	Ø	33/19	mm
DC Resistance	Re	6/5.6	Ω
Voice Coil Inductance	Lbm	50/-	μH
X-Max peak		2/-	mm
Magnet Weight	m	.33/.008	kg
Force Factor	BL	6.5/2.5	TM
Suspension Compliance	Cms	1018/-	mN^{-1}
Mechanical Q Factor	Qms	1.8/-	
Electrical Q Factor	Qes	0.39/-	
Total Q Factor	Qts	0.32/-	
Equivalent Air Volume	Vas	14	Ltr
Moving Mass	Mms	7.8/0.19	g
Effective Piston Area	SD	.01/-	m^2
Price Each		**$63.00**	

Suggested Box Alignments		
	Sealed	Vented
VB Ltrs	3.7	6
F3 Hz	125	85
Fb Hz	-	72
Vent Ø	-	1.5"
Vent L	-	3.3"

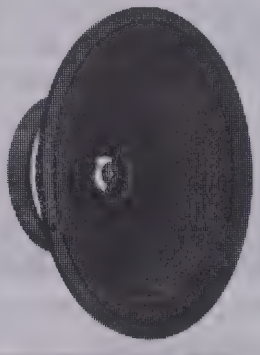


SYNCHRON SYN-619A 6.5"			
Technical Data	**Symbol**	**Value W/T**	**Unit**
Nominal Impedance	Z	8/8	Ω
Resonance Frequency	Fs	55/1500	Hz
Power Handling (IEC)	P	30/30	W
Sensitivity (1W/1m)	E	90/90	dB
Voice coil Diameter	Ø	33/19	mm
DC Resistance	Re	6/5.6	Ω
Voice Coil Inductance	Lbm	47/-	μH
X-Max peak		2/-	mm
Magnet Weight	m	.33/.008	kg
Force Factor	BL	7/2.2	TM
Suspension Compliance	Cms	803.3/-	mN^{-1}
Mechanical Q Factor	Qms	2.1/-	
Electrical Q Factor	Qes	0.44/-	
Total Q Factor	Qts	0.36/-	
Equivalent Air Volume	Vas	20	Ltr
Moving Mass	Mms	10.3/0.19	g
Effective Piston Area	SD	.013/-	m^2
Price Each		**$68.00**	

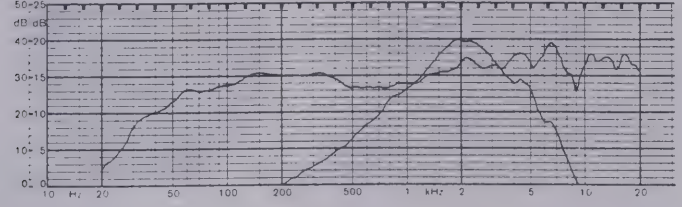
Suggested Box Alignments		
	Sealed	Vented
VB Ltrs	7.5	12
F3 Hz	105	65
Fb Hz	-	61
Vent Ø	-	2"
Vent L	-	3.8"



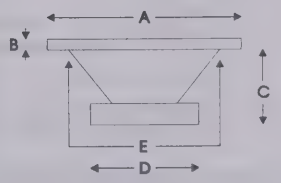
SYNCHRON SYN-825A 8"			
Technical Data	**Symbol**	**Value W/T**	**Unit**
Nominal Impedance	Z	8/8	Ω
Resonance Frequency	Fs	40/1500	Hz
Power Handling (IEC)	P	60/40	W
Sensitivity (1W/1m)	E	91/90	dB
Voice coil Diameter	Ø	42/25	mm
DC Resistance	Re	6/5.6	Ω
Voice Coil Inductance	Lbm	58/-	μH
X-Max peak		4/-	mm
Magnet Weight	m	.567/.013	kg
Force Factor	BL	9.1/3.18	TM
Suspension Compliance	Cms	808.4/-	mN^{-1}
Mechanical Q Factor	Qms	1.74/-	
Electrical Q Factor	Qes	0.36/-	
Total Q Factor	Qts	0.30/-	
Equivalent Air Volume	Vas	50	Ltr
Moving Mass	Mms	19.7/0.49	g
Effective Piston Area	SD	.021/-	m^2
Price Each		**$78.00**	



Suggested Box Alignments		
	Sealed	Vented
VB Ltrs	11	16
F3 Hz	94	60
Fb Hz	-	54
Vent Ø	-	2"
Vent L	-	3.6"



Dim. mm	SYN-519A	SYN-619A	SYN-825A
A	154.5	180.3	217
B	5	5	5.5
C	77	83.3	98
D	90	90	110
E	119	145.4	187



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Dealer inquires welcome. Supronyl and Titanium domes available at 200 piece quantities.

Reader Service #1

A DUAL-VOLUME V_{AS} TEST BOX

One of the ways to measure a woofer's V_{AS} is to mount it in a sealed enclosure of known volume and measure the shift in its resonant frequency (f_S) and electrical Q (Q_{ES}). *The Loudspeaker Design Cookbook*, by Vance Dickason (Audio Amateur Press, 1991), calls for a $1ft^3$ enclosure for drivers up to 10″ and a $2ft^3$ box for larger ones. To avoid building and storing two separate boxes, I designed and built a dual-volume test box. The basic idea is a $2ft^3$ box with a removable divider. With the divider installed, the internal volume loading the woofer is reduced to $1ft^3$.

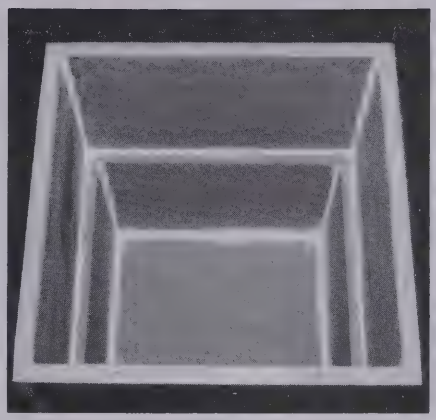

PHOTO 1: Looking down into the test box.

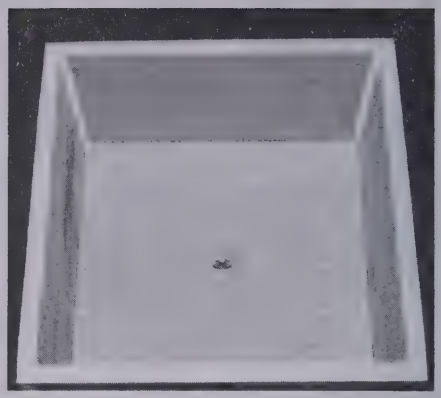

PHOTO 2: Top section of box with divider in place.

DESIGN

I made five top pieces, one each for 6″, 8″, 10″, 12″, and 15″ woofers. For increased strength, I added an inner half-wall to support the divider, rather than using shims (*Fig. 1*); I didn't want to worry about breaking the shims in when I pushed the divider into the box. I used ¾″ thick wood (MDF) because the internal volumes came very close to their targets and I wished to minimize wall vibration. You can use the same procedures to calculate the correct volumes with other wood thicknesses.

The first requirement was to make the top of the box large enough to mount a 15″ woofer. A 16″ × 16″ top would accomplish this, and a single sheet of wood yielded three top pieces.

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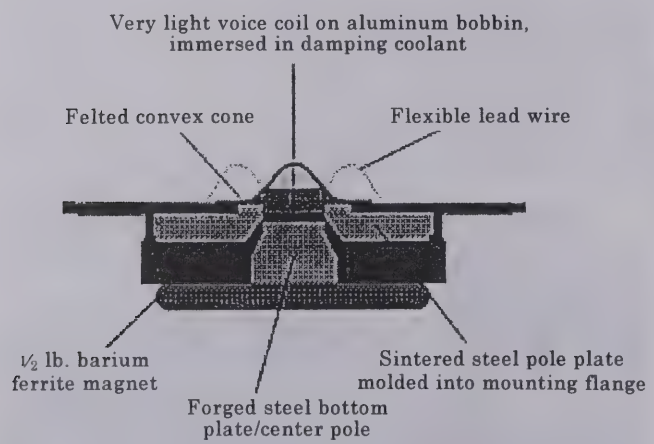
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Reader Service #39

BLACK HOLE

High efficiency dampening products from ORCA

A better speaker damping material...

If you've been building speakers for some time, you know how much guesswork goes with speaker damping and stuffing. The choices seem endless: fiberglass, wool, Dacron, flat foam, convoluted foam, felt, tar, plus various "magic" compounds that you're invited to brush or pour into your new cabinets. Everyone has their own recipe, and who knows if it's a recipe for disaster? Or what effects the vapors emitted by these chemicals might have on the glues that bond your woofer surround to its cone and chassis? In this era of costly, space-age drivers and computer-assisted design, we think such risks are totally unacceptable. So we went to work to find the ideal solution.

The problems are fairly well-known: a driver transforms electrical energy into mechanical energy. This mechanical energy is transformed into acoustical energy which is radiated to the outside of the cabinet - the useful front wave - and to the inside - the sometimes-useful back wave. Unfortunately, it is also transmitted though the frame of the driver to the cabinet itself, which acts as a very large "cone" of very small excursion. This means that the spurious resonances and vibrations of the cabinet have to be controlled in a predictable and reproduceable way. That's how we came to BLACK HOLE 5 and the BLACK HOLE PAD.

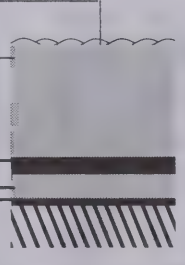
First, THE PAD. It's a thin (1/16 inch) black flexible viscoelastic damping material (filled vinyl copolymer) with maximum performance between 50 and 100 degrees F (we hope that that covers the temperature range of your listening room) and excellent flame resistance - it meets UL94 V-O. Thanks to its outstanding damping characteristics, THE PAD will dramatically reduce the vibration energy stored in the walls to which it is applied.

Easy to cut and apply, THE PAD has a pressure-sensitive adhesive back: simply peel off the release paper and press hard onto a clean surface. You can use THE PAD on just about anything you suspect of vibrating: driver frames, thin panels like car doors, and, of course, the walls of your speaker cabinets. And it can be used to recess a driver without using a router: just laminate enough layers to match the thickness of the driver frame and apply to the front baffle. Finally, it is the ideal material for "constrained layer" wall construction, where two panels are laminated on each side of a damping material for optimum transmission loss. Because THE PAD has a fine grain leather finish, you can wrap an entire cabinet exterior and give it an attractive appearance at the same time!

For applications which require **maximum damping, isolation and absorption,** we've developed BLACK HOLE 5. One and 3/8" thick, BLACK HOLE 5 is a high-loss laminate that provides optimum acoustical damping performance. It consists of five layers:

Thin diamond-pattern embossing, densified with a polyurethane film surface. This unique surface layer dramatically improves the performance of the whole acoustical system, especially the lower mid-range and mid-bass frequencies where simple acoustical foam loses its effectiveness.

One-inch deep polyester urethane foam, structurally optimized for acoustical damping. Highly effective at "soaking" maximum sound energy with minimum thickness.



Barrier septum, 1/8 inch thick. Made of limp flexible vinyl copolymer loaded with non-lead inorganic fillers, it is a "dead wall" that isolates the vibrations in the walls of your cabinet from the vibrations created inside the enclosure.

Polyester urethane flexible open-cell foam, 1/4 inch thick. Thanks to special vibration-isolation characteristics, it decouples the vibrating structure (the wall) from the rest of the damping system, thus optimizing performance.

High-loss vibration damping material, same as The Pad. It is strongly bonded to the cabinet wall with pressure sensitive adhesive.

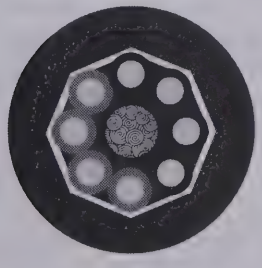
These layers are laminated using an adhesive-free mechanical and thermal process, thus optimizing performance and eliminating the risk of solvent fume damage. BLACK HOLE 5 can be used in any enclosure, as well as for acoustical panels to improve the characteristics of your listening room. *YOU PROVIDE THE MUSIC; BLACK HOLE FIVE WILL TAKE CARE OF THE NOISE!*

orca 1531 Lookout Drive Agoura, CA 91301 U.S.A 818-707-1629 FAX 818-991-3072

AXON Cables

New from ORCA!

AX-ON (Greek axon, axis): that part of a nerve cell through which impulses travel away from the cell body. AXON 8 speaker cable combines outstanding design features with component quality usually associated with the most expensive cable. With eight AXON 1 solid-core conductors and utilizing mylar/ polypropylene construction, AXON 8 offers outstanding performance for amp-speaker connections and perfectionist internal speaker wiring. Our superb AXON 1 AWG 20 solid core conductor is also available separately. Oxygen-free and 99.997% pure, it is ideal for most internal wiring applications.



Outer insulation: UL approved TPE
Cable geometry: non interleaved spiral
Individual conductor insulation: 105 degree Celsius, UL approved PVC
Cable equivalent gauge: total - AWG 11, 2 conductors - AWG 17, 4 conductors - AWG 14
Individual conductors: solid core AWG 20 copper, long-grain and ultra-soft, free of all contaminants and oxygen.
Cable core: crushed polypropylene
Inner envelope: mylar film

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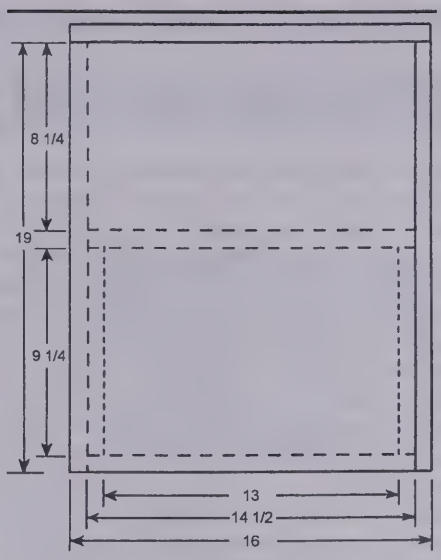

FIGURE 1: Side view of test box.

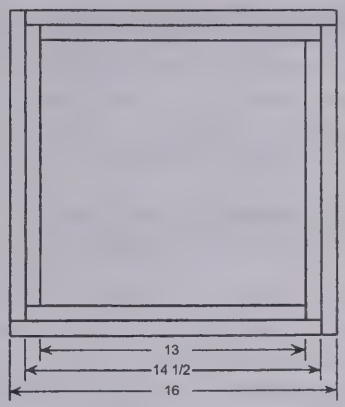

FIGURE 2: Top view with top and divider removed.

Given the size of the top piece and the thickness of the sides, I knew two of the top section's internal dimensions (14½" × 14½"). It remained only to figure out the height for a 1ft³ volume. The internal dimensions of the top section ended up being 14½" × 14½" × 8¼". The target volume for the top section is 1,728 in³ and these dimensions give you about 1,735 in³—within 0.4% of the target.

To use the full 2ft³ volume, you simply remove the divider. So when figuring the dimensions for the rest of the enclosure, remember to include the divider's volume. With the 1" divider removed, the dimensions of the top section are 14½" × 14½" × 9". The top section volume now equals 1,892 in³. The bottom section volume then needs 1,564 in³ more to equal 3,456 in³ (2ft³). The dimensions I used were 13" × 13" × 9¼" (*Fig. 1*), which gives you 1,563 in³ for the bottom section and 3,455 in³ total.

TABLE 1

TEST BOX PARTS LIST

QTY	PART	DIMENSIONS
5	lids	16" × 16"
4	outside sides	19" × 15¼"
2	bottom and divider	14½" × 14½"
4	half-sides	9¼" × 13¾"

PHOTO 3: My dual-purpose test box.

CONSTRUCTION

Construction of the box is rather straightforward. I used an overlapping butt joint technique (*Fig. 2*). The nice thing about this method is that all four side pieces are the same size. All pieces are glued and screwed together using countersunk flat-head wood screws, with the exception of the divider and top(s), which to facilitate removal are mounted with only pan-head wood screws (no glue).

The first step is to glue and screw together the half-sides and the bottom. Temporarily mount the divider, using screws only, to get the proper alignment of the holes; then remove it. Attach the outside sides with more glue and screws and let the glue dry overnight. Then seal all the joints in the lower section and the corners of the upper section (I used silicone caulk). Attach some sort of seal (I used weather-strip foam) along the top edge of the half-sides and outer sides. You want to avoid any leaks from the outer box or into the lower section while the divider is installed.

Finally, mount a handle (I used an eyebolt) in the center of the divider. Also round the corners of the divider so it doesn't scrape off the top section's caulking as you install or remove it.

With the divider installed, the woofer is loaded by the top 1ft³. With the divider removed, the woofer sees the entire 2ft³. I hope this design makes your testing procedures easier.

Stephen A. Crosby
APO AE 09470

Speaker Builder

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Product Review

D.H. LABS SILVER SONIC AUDIO CABLES

By Gary A. Galo
Contributing Editor

D.H. Labs' Silver Sonic Audio Cables, D.H. Labs, 6633 Glenway Dr., West Bloomfield, MI 48322, (810) 851-1296. **T-14 Speaker Cable**. Conductors: 2 × 14AWG, silver plated oxygen-free copper, stranded. Dielectric: Pure Teflon® TFE. Capacitance: 21.5pF/ft. $3.00/ft. **BL-1 Balanced Interconnect Cable**. Conductors: 2 × 20AWG, silver plated oxygen-free copper, stranded. Dielectric: Teflon copolymer. Shield: Foil with stranded drain wire. Capacitance: Not specified (see text). $3.50/ft.

D.H. Labs is a newcomer to the audio cable scene, and its first products are sure to attract attention among those who believe that cables make a difference. High-end audio designers generally agree that Teflon® is the best solid dielectric material. Measurements introduced nearly fifteen years ago have shown the superiority of Teflon in critical performance areas, including dissipation factor and dielectric absorption (among other things, audio cables are capacitors).[1] D.H. Labs has recently introduced its new "Silver Sonic" cable line, including the T-14 loudspeaker cable and BL-1 interconnect (*Photo 1*).

The T-14 consists of two 14AWG conductors with pure DuPont Teflon TFE insulation. TFE is the highest grade of Teflon available. The stranded conductors are made of oxygen-free copper, and each is individually coated with silver. Silver is a superb conductor—but it oxidizes easily. To prevent this, the T-14's conductors are tightly sealed with the Teflon insulation. The T-14's silver coating has a thickness of 40µ, twice as thick as some silver coatings.

The insulated conductors are spiraled for low inductance, yet capacitance is also low—around 22pF/ft. These cables should mate well with most amplifier/loudspeaker combinations. The center conductors are wound with a vibration-damping filler and tightly covered with a tension wrap and heavy outer jacket, making the cable virtually immune to resonances and microphonics.

The BL-1 is a fully balanced interconnect cable consisting of two insulated center conductors, plus a foil shield and a separate drain wire. The conductors are made from slow-drawn, oxygen-free copper. Each of the fine strands is individually plated with silver. The dielectric is a new Teflon copolymer. D.H. Labs' president, Darren Hovsepian, claims the new copolymer has nearly identical electrical characteristics as the pure Teflon TFE used in the T-14 loudspeaker cable.

The copolymer requires less heat in the extrusion process, however, which is advantageous when manufacturing a finely stranded cable. The overall gauge of the conductors and drain wire is 20AWG, heavier than many interconnects. The BL-1 contains vibration-damping fillers similar to those used in the T-14 speaker cable. The outer jacket is extruded under pressure, further reducing the possibility of resonances and microphonics.

I evaluated the T-14 and BL-1 cables with my biamped Audio Concepts Sapphire II*ti*/Sub1 loudspeaker system, fed with a pair of Adcom GFA-585 power amps (the Sapphire II*ti*'s amp is modified) and an extensively modified Adcom GFP-565 preamp. The Sapphire II*ti* satellites are biwired (detailed in "Bi-Amping the Sapphire II Sub-I System," *SB* 3/92, p. 24). I fitted the T-14 loudspeaker cables with gold-plated Old Colony's fork lugs. I use Canare F-10 RCA connectors with the BL-1 cables (more on connectors below).

I had several weeks to evaluate the T-14 speaker cable before the F-10 interconnect arrived. My existing speaker cables were AudioQuest Indigo Blue; my interconnects were made from Grado Signature Laboratory Standard Audio Cable (no longer available in bulk form). Both of the D.H. Labs "Silver Sonic" cables made a significant improvement in my system. I was enormously impressed with the increase in detail and resolution.

Soundstage reproduction is also superior to my previous cables, being larger than I'm accustomed to. The D.H. Labs cables are extremely smooth and extended in the high frequencies, with plenty of the elusive air and space, yet free of any harshness. I have found them to be extremely musical over the long haul (I've used the BL-1 interconnects for several weeks, and the T-14 speaker cables for over two months).

If the BL-1 is used in an unbalanced application (e.g., most home installations), the shield should float at one end. On the source end, connect both the black center conductor and the drain wire to the shield of the RCA connector. On the load end, leave the drain wire unconnected. In this configuration, the BL-1 has a capacitance of about 56pF/ft. I recommend using a small piece of heat shrink tubing to make sure the drain wire remains insulated on the load end of the cable. The arrows printed on the cable should point toward the load. The cable isn't manufactured with directional characteristics, but the arrows provide an easy means of identifying the load end once the cable has been fitted with RCA connectors.

If you purchase the BL-1 interconnect cable, you'll have to decide which RCA connectors to use. My current favorite is the Canare F-10 (*Photo 2*). Canare has received little attention in the audiophile community; it is best known to audio professionals, primarily for

PHOTO 1: D.H. Labs' new Silver Sonic cables. From top to botton, the BL-1 balanced interconnect and the T-14 loudspeaker cable.

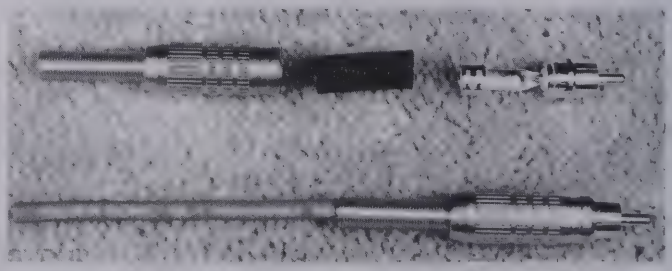

PHOTO 2: The Canare F-10 RCA connector shown disassembled, and fitted with D.H. Labs' BL-1 interconnect cable.

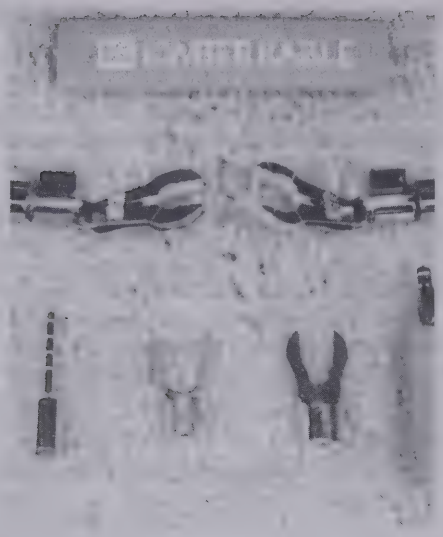

PHOTO 3: High-performance speaker cable terminations. Top, left to right: Kimber Kable's state-of-the-art PM-25 and PM-33 Postmasters, both supplied with heat-shrink tubing and Wonder Solder. Bottom, left to right: D.H. Labs' P-1 pin connector, SP-1, SP-2 spade connectors, and B-1C banana plug.

its excellent L4E6S balanced quad microphone cable. The F-10 RCA plug is a rugged gold-plated connector. The gold plating is extremely high quality, applied directly on brass. The best news is the insulator—it's pure Teflon. Canare's catalog simply lists the generic chemical name—Polytetrafluoroethylene, which is Teflon TFE.[2]

At around $3.50 each, the Canare F-10 connectors are the most economical Teflon-insulated RCA connector made. They compare favorably with high-end audio types costing several times as much—audiophiles usually pay at least four times this price for connectors of this quality.

Canare products are usually sold through pro dealers, including Hudson Audio-Video (listed below). HAVE carries many other items that might be of interest, including a wide selection of audio and video tape. Call for a free catalog; you might find enough other items to make up the $50.00 minimum order. Canare will send you a complete list of dealers with its catalog. Check the others for minimum order info. D.H. Labs was still deciding on which RCA connectors it intended to sell at the time of writing. I recommended the F-10.

You'll also need to purchase terminations for the T-14 speaker cables. The gold-plated fork lugs sold by Old Colony work well (cat. #SCFFF, $2.95/pair). D.H. Labs offers four different gold-plated terminators at a very reasonable cost. The SP-1 spade connector will accept 10AWG wire; cost is $2.20/pair. The SP-2 spade connector will accept up to 8AWG wire; cost is $5.00/pair. D.H. Labs also offers the B-1C banana connectors and P-1 pin connectors, at $5.00 and $3.00/pair, respectively.

Kimber Kable recently supplied samples of its Postmaster spade connectors (*Photo 3*). If cost is no object, these are the best speaker cable terminations I've seen. Postmasters are plated with Kimber's proprietary Ultraplate alloy. They have a compressed silicon wafer between the contact parts, which results in a gas-tight connection with only finger-tight torque. Two sizes are available. The PM-25s fit conventional ¼" binding posts; the PM-33s fit a 3/8" post. Cost is $14.00/pair.

Recently I evaluated the Kimber Kable KCAG interconnects and 4AG speaker cables. These cables are made with Kimber's proprietary Silver Varistrand conductors and Teflon insulation. The interconnects are terminated with Kimber's RCA-M plugs, and the speaker cables with Postmasters. The KCAG interconnect cables cost $350/meter pair. The 4AG speaker cables are $100/foot. While I found the Kimber Kables *slightly better* than the D.H. Labs cables in most areas of performance, it's clear that the law of diminishing returns is at work.

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Reader Service #22

A 1m pair of D.H. Labs BL-1 cables fitted with Canare F-10 connectors will cost around $30. The same length of Kimber KCAG with Kimber RCA-M plugs will cost nearly 12 times that amount. I think few listeners would say they were 12 times as good. Kimber's 4AG speaker cables are nearly 30 times the cost of the D.H. Labs T-14, if the T-14 is fitted with one of D.H. Labs gold-plated terminations. Again, the Kimber Kables are excellent, but, in my view, nothing like 30 times as good as the D.H. Labs cable. D.H. Labs offers unprecedented value, with performance excellent even when compared to cables selling for many times the price.

Darren Hovsepian also suggests trying the BL-1 as a digital interconnect. According to his measurements, with the shield floating on both ends (i.e., disconnected), the cable's impedance is 79Ω. I made a 1m digital interconnect with this cable and Canare F-10 connectors and compared its performance to my custom-made ones (described in *Ask TAA* 2/93). The BL-1 cable produced a considerably larger soundstage than mine, and superior inner detail.

The D.H. Labs cables are easy to work with. The BL-1 interconnect is flexible enough to warrant serious consideration for internal wiring in preamps, electronic crossovers, and other low-level circuitry. The use of a drain wire/foil shield combination makes the shield connection easy. D.H. Labs is expanding its dealer network, so ask for the name of a dealer near you. Old Colony Sound Lab carries both the T-14 and the BL-1 for $2.99/foot. Both Canare and D.H. Labs have brought the cost of high-quality Teflon-based cables and connectors within reach of any serious audiophile. These products are highly recommended.

SOURCES

D.H. Labs
6633 Glenway Dr., West Bloomfield, MI 48322
(810) 851-1296
T-14 Speaker Cable; BL-1 Interconnect

Canare Cable, Inc.
511 5th St., Unit G, San Fernando, CA 91340
(818) 365-2446
F-10 RCA Connectors

Kimber Kable
2752 South 1900 West, Ogden, UT 84401
(801) 621-5530.
Postmaster speaker connectors

Hudson Audio-Video Enterprises, Inc. (HAVE)
309 Power Ave., Hudson, NY 12534
(518) 828-2000
Canare products

Old Colony Sound Lab
PO Box 243, Peterborough, NH 03458
(603) 924-6371; FAX (603) 924-9467
dealer for D.H. Labs products

REFERENCES

1. W. Jung and R. Marsh, "Picking Capacitors," Parts 1 and 2, *Audio* (Feb. and March 1980).
2. T. Jones, *Electronic Components Handbook* (Reston Publishing, 1978): 4.

MANUFACTURER RESPONSE

I would like to thank SB and Gary Galo for reviewing our Silver Sonic audio cables. Our goal for the Silver Sonic line is to offer high-end performance at affordable prices, and we are very pleased that Gary was impressed with the performance and value of the cables.

One point that I believe would be of special interest to speaker builders is that the conductors in our speaker cable (and the interconnect) are tightly sealed in the Teflon insulation and will not oxidize over time. This is especially important for the wiring inside speaker enclosures, which cannot be easily replaced.

Also, the interconnect is now available in factory terminated lengths as well as by the spool.

Darren Hovsepian
D.H. Labs

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Reader Service #38

SMT PROJECTS — BKAA29

Owen Bishop — **$14.50**

This book is a practical introduction to Surface Mount Technology (SMT). SMDs (Surface Mount Devices) are miniaturized versions of traditional electronic components, soldered directly to the surface of the circuit board instead of connected by wire leads or terminal pins. Because of their small size, they are widely used in much of today's portable equipment, from calculators and camcorders to cellphones and smart cards. This volume describes the special features of SMT and how to work with SMDs on the workbench. It includes over 20 construction projects, all of which have many useful applications in and around the home and workshop. These include a warbler; blinker; passive I-R intruder detector; bicycle alarm; impatient and wailing sirens; clap switch, VHF transmitter; low-voltage AF amplifier; slave flash; hourglass timer; logic probe; digital die; mini radio receiver; two-tone (model) locomotive horn; sample-and-hold device; gunfire generator; sparkling pendant; live wire detector; multi-level resistance probe; and micro microwave (with sound and light for a dollhouse, but does not cook!). PCB layouts provided, for making your own boards. United Kingdom, 1993, 174pp., 5½″ × 8¼″, softbound.

CAD MUSIC AUDIO REFERENCE AND STUDIO TOOLS CD — CDCAD1 — $34.95

Contents: Reference Part—[1] Dramatic movie magic. [2] Soundscapes. [3] Monica's nylon dreams. [4] Dynamic flow. [5] Classical acoustic guitar. [6] Critical listening, *acoustic arpeggio*, 8 repeats, Q 1.8; (01) no equalizing; (02) +3dB 125Hz; (03) +3dB 250Hz; (04) +3dB 500Hz; (05) +3dB 1.0kHz; (06) +3dB 3.0kHz; (07) +3dB 6.0kHz; (08) +3dB 10.0kHz. [7] Chunky guitar jam I. [8] Chunky guitar jam II. [9] Splash in the wilderness. [10] Cave stream I. [11] Cave stream II. [12] Cave stream III. [13] Niagara Falls, falling rocks. [14] Stream. [15] Stream close up. [16] Rain and rolling thunder. [17]-[21] Grooves mix check, each approximately 2:05; funky swing, 103 bpm; silky house, 120 bpm; Minneapolis, 100 bpm; Ritz percussion, 113 bpm; L.A. beat, 104 bpm. [22] Guitar (Wah 1; Wah 2; Dance riff I A,C; Dance riff II A,C; Pick 1 A; Pick 2 C; Feedback C; C & C git stab; Metal lick; Dive bomber).

Studio Tools—[23] Left and right connection. [24] Digital zero, reference level 1kHz. [25] SMPTE code 25 fps, 8 min starts at 00:59:55:00. [26] Pink noise; 0:30 left channel; 0:30 right channel; 1:00 both channels. [27] Silence, 0:30. [28]-[30] Mastering calibration, 1kHz, 10kHz, 100Hz; 0:15 left and right channels; 1:00 both channels. [31]-[34] Frequency check, each frequency 0:15; 20, 32, 40, 64Hz; 120, 280, 420, 640Hz; 800Hz, 1.2, 2.8, 5.0kHz; 7.5, 12, 15, 20kHz. Produced by Claes Dahlström for CAD Music. Ref. #001g. Sweden, 1992, 61:03.

MCGRAW-HILL ELECTRONICS DICTIONARY — BKMH9

John Markus, Neil Sclater — **$49.95**

Reflecting the explosive progress in the field of electronics, the fifth edition of this best seller is now available. Currently the only dictionary that focuses on words and phrases specific to the field of electronics—plus such related fields as engineering, the physical sciences, materials science, and computer science—this reference remains a premier source for information in all areas of electronics. From CDs and VCRs to solid-state lasers and liquid-crystal displays, rapid expansion of electronic technology has left its mark on the language. Many of the dictionary's new terms have their origin in personal computing, on-line networks, and electronic media, while others are derived from the intersection of electronics with such specialized sciences as metallurgy, biophysics, and astrophysics. More than 14,000 entries, including more than 1200 new entries and 150 new illustrations, make this an essential reference for professionals and anyone who wants a better understanding of technical articles, books, catalogs, advertisements, and new product announcements on electronic subjects. 1994, 596pp., 7⅜″ × 9¼″, hardbound.

SILVER-PLATED HOOKUP WIRE

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MWR16BK	#16 Stranded Hookup, 25 feet	**$9.95**

AFTERTOUCH SAMPLER CASSETTE #4 — CASAT4 — $9.95

Creative Musicians Coalition

Fresh sounds from independent electronic musicians with a wide variety of exceptional talents and styles enable you to experience a fascinating cross-section of the best on the electronic scene today. #ATVOL4C, 50:00.

Contents: Side One–[1] "Electric City" (from *Electric City*), Jack Poley. [2] "The Maya" (*Angels, Aliens, & Archtypes*), Mark Dwane. [3] "Globe" (*Deeper in the Dreamtime*), Robert Scott Thompson. [4] "Return of the Ancient Children" (*On Time*), Robert Newcomb. [5] "Setting Sail" (*Inner, Stellar*), Fred Becker. [6] "Capriccio No.1, OP. 76–Brahms" (*Brahms: Neue Bahnen*), Norris Mills. [7] "Mars" (*Time and Space*), Mark Baechtle. [8] "White Browed Robin (The Aviary)" (*1992 Spring Sampler*), David Bagsby. Side Two–[1] "People Don't" (*Warm Places*), Neil Davis. [2] "Excerpt from Dot Black" (*Dot Black*), Petru Grajdian. [3] "Muladhara" (*The Seven Chakras*), D'Orlando. [4] "Solarwinds" (*Spatial Logic*), Michael Milazzo. [5] "Angelight" (*Angelight*), J.C. Sebastian. [6] "Triple Conjunction" (*Gardens*), Kelly Dowhower. [7] "The Tarn" (*Summit Approach*), Jeff Stover. [8] "Are You in the Computer?" (*Run the Gauntlet*), Hartley C. White.

NEW THIS ISSUE

AFTERTOUCH SAMPLER SPECIAL! — CASAT/S4

Creative Musicians Coalition — **$33.95**

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SCP-1	#10-#14 Pin Connector, ⅝″ pin Gold-plated, solder only.	**Each $2.95**

MICROPHONE MANUAL: DESIGN AND APPLICATION — BKB29

David Miles Huber — **$36.95**

Written for both the beginner and the professional, this book provides the latest techniques, tools, and tips needed to keep you abreast of the ever-changing technologies of microphone design and placement and environmental acoustics. Topics covered include: the microphone pickup and its method of operation; cable/connector interfacing; single-microphone placement; stereo "miking" fundamentals, including spaced-pair, coincident, and near-coincident techniques; applied placement techniques with the music production environment, video/film media, and speech and music reinforcement; and microphone accessories. 1988, 297pp., 7″ × 10″, softbound.

AUDIO ENGINEER'S REFERENCE BOOK — BKB28

Michael Talbot-Smith, editor — **$159.95**

This text covers a wide range of audio topics at a level equivalent to first year degree standard in electronic engineering/physics. Written by acknowledged experts in audio engineering, its contents include Basic principles; Acoustics and acoustic devices; Recording and reproduction; Digital equipment; Studios and their facilities; Distribution and audio signals; and Miscellaneous topics. United Kingdom, 1994, hardbound. IN PRODUCTION; PLEASE CHECK AVAILABILITY.

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NEW THIS ISSUE

AUDIO IC CIRCUITS MANUAL BKB24
R.M. Marston **$24.95**

A useful guide for the audio IC user, dealing with devices such as low frequency linear amps, dual preamps, audio power amps, charge coupled device delay lines, bar-graph display drivers, and power supply regulators. United Kingdom, 1988, 168pp., softbound.

SOUND RECORDING AND REPRODUCTION BKB20 $19.95
Glyn Alkin

This book is an exceptionally comprehensive overview of the whole field of current audio art and technology. It contains a wealth of vital information of interest to the professional sound recordist as well as the enthusiastic amateur wishing to improve his techniques or the keen listener interested in gaining a better understanding of the recording medium and sound quality in general. Beginning with a look at the nature of sound itself, the reader is able to follow the progress of a recording from the acoustic environment, through selection of equipment and microphone balance techniques to the various forms of audio reproduction including disc, tape, digital magnetic, optical, and thermomagneto-optic disk recording. United Kingdom, 1991, 267pp., 5½″ × 8½″, softbound.

ADVANCED AUDIO PRODUCTION TECHNIQUES BKB27
Ty Ford **$28.95**

This book explains how to create concepts and use current audio technology to achieve "audio pictures" in the mind of the listener. By using visual language, charts, and graphs, the author offers a multidimensional approach to the realization of practical concepts that are the basis of good audio. Key features include interviews with award-winning audio producers; step-by-step instructions on connecting and operating equipment; clear, easy-to-understand explanations of digital theories and operations; and insight into envisioning and executing the right sound. Through use of this book, even if you do not have a strong technical background, you will be able to operate complex audio systems successfully. 1993, 148pp., 6″ × 9¼″, softbound.

THE ART OF THE SOUND EFFECTS EDITOR BKB21
Marvin M. Kerner **$21.95**

In this volume directed toward the film student, beginning film professional, and apprentice sound editor, the author sets forth his personal approach to sound editing, based on his years of experience with vehicles such as *Ben Hur*, *Dr. Zhivago*, and "The Man from U.N.C.L.E." In an easy-to-read, anecdotal style, he describes the responsibilities of the sound editor and the assistant sound editor, taking the reader through the various steps of the sound editing process. 1989, 107pp., 6⅛″ × 9¼″, softbound.

NEWNES AUDIO & HI-FI HANDBOOK BKB25
Ian R. Sinclair, editor **$79.95**

This is the revised and updated second edition of the previously entitled *Audio Electronics Reference Book*, which has been substantially reduced in price and now has many new contributors. The aim is to present as wide a perspective as possible of high-quality sound reproduction, including reproduction under adverse circumstances (PA and in-car), from less conventional sources (such as synthesizers), and with regard to the whole technology from studio to ear. Chapters include: Sound Waves (by Dr. W. Tempest); Microphones (John Borwick); Studio and Control Room Acoustics (Peter Mapp); Principles of Digital Audio (Allen Mornington-West); Compact Disc Technology (Ken Clements); Digital Audio Recording (John Watkinson); Tape Recording (John Linsley Hood); Noise Reduction Systems (David Fisher); The LP Record (Alvin Gold); Disc Reproduction (Don Aldous); Tuners and Receivers (Hood); Preamps and Inputs (Hood); Voltage Amplifiers and Controls (Hood); Power Output Stages (Hood); Loudspeakers (Stan Kelly); Loudspeaker Enclosures (Kelly); Headphones (Dave Berriman); Public Address and Sound Reinforcement (Mapp); In-Car Audio (Berriman); Sound Synthesis (Mark Jenkins); Interconnections (Mornington-West); and The Future (Sinclair). United Kingdom, 1993, 820pp., 6¼″ × 9½″, hardbound.

EFFECTIVE AUDIO-VISUAL: A USER'S HANDBOOK BKB22
Robert S. Simpson **$42.95**

Here is a handy volume for anyone who needs to know more about audio-visual equipment and techniques. It not only discusses the traditional A-V approaches such as slide-tape, multi-image, film, and video, but also introduces the role of computer graphics and the application of interactive A-V. These varied approaches are discussed as they apply to the most commonly encountered A-V uses–business presentations, conferences, training sessions, museum exhibits, and visitors' center displays. This second edition has been extensively rewritten, with updated illustrations and new chapters on commissioning A-V programs and on computer-based multimedia. United Kingdom, 1992, 241pp., 7½″ × 10″, hardbound.

SOUND EFFECTS: RADIO, TV, AND FILM BKB23
Robert L. Mott **$42.95**

Created for students preparing for a career in radio, TV, or film, and professionals wishing to expand their knowledge of the field, this book offers invaluable information on the role sound plays in these media. All aspects of the craft are explored, ranging from the techniques used in Shakespeare's Globe Theatre to the challenges and expectations of today's electronic venues. The reader will learn what sound effects are, why they are used, and how they are created. Included are discussions of the history of sound effects; different types of sound effects; creating sound effects from scratch; recording sounds in the studio and field; the advantages of live sounds over tape; knowing why and when to use sound effects; Foleying and the Foley stage; and recording and editing equipment. 1990, 223pp., 6¼″ × 9½″, hardbound.

ACOUSTICS IN THE BUILT ENVIRONMENT: ADVICE FOR THE DESIGN TEAM BKB26 $97.95
Duncan Templeton, editor

Acoustics in the Built Environment is an invaluable work of reference for the building professional, covering all aspects of acoustics. It is unique in its range of topics: the environment, transport infrastructure, building design, building systems, and buildings in use. Each section has been contributed by an expert in the field, and presents information highly relevant to the day-to-day work of project design teams in a concise, readily accessible, and usable form. Chapters include: Environmental acoustics; Design acoustics; Services noise and vibration; Sound systems; and Technical information, comprised of both a detailed glossary and listings of equivalent, international, American, British, French, and German standards. United Kingdom, 1993, 158pp., 8⅜″ × 12″, hardbound.

VALKYRIE
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If you want to hear sound that reproduces the original as closely as possible, this hot new preamp is for you! Designed by Grayson King and introduced in the 1/94 issue of *The Audio Amateur*, the Valkyrie reveals many details that might not have been audible previously and is likely to bring to your ears much that you've been missing. Three buffers included (four possible), as well as state-of-the-art power supply. Top-quality parts throughout: Avel transformer, Panasonic caps, Roederstein resistors, Alps "Blue Velvet" pots. Article reprint included. Build it yourself to save, or buy it with the boards ready to go!

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◆ ◆ ◆

Vance Dickason, bestselling author of *The Loudspeaker Design Cookbook* (over 33,000 English copies sold, with German and Portuguese versions also available) brings almost twenty years of experience, research and hands-on knowledge to this new book on designing and building four tested and proven two-way loudspeaker systems.

◆ ◆ ◆

The best news, however, is that the directions, definitions and instructions for realizing these four demonstration systems, contain all the right questions and the clear paths to their answers which enable you to build excellent performing two-way systems with an almost endless combination of available woofers and tweeters.

◆ ◆ ◆

Dickason shows you how to look for the right driver characteristics to be paired successfully with the right crossover components and also how to fix any anomalies which may trouble your particular choices. Although the four two-way systems which are meticulously documented in this book, along with an outstanding general purpose subwoofer, may be built just as they are defined here, you are not limited to building these systems only.

◆ ◆ ◆

Loudspeaker Recipes, Book 1, lays a firm groundwork, both theoretical and practical, for building as many varied and successful two-way loudspeaker systems as you wish. The book benefits from the latest techniques for computer aided design, but is also rich in proven construction practices for building practical systems. Mr. Dickason is not only a published author and Editor of Voice Coil, the monthly newsletter for the loudspeaker industry, a Contributing Editor for *Speaker Builder*, but also is a professional consultant and product reviewer as well.

◆ ◆ ◆

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BUFFER 'N SPICE

I would like to thank Dick Campbell for his kind words regarding my buffer design ("Exploring BUF 124 with PSpice" *SB* 3/94, p. 39).

Although I have access to PSpice and try to simulate all of my circuits, simulation has not entirely replaced bench testing...for two reasons. First, as Mr. Campbell points out, PSpice has problems with the cascode-complementary topologies. In some more complex configurations PSpice can't handle the circuits, giving "convergence" errors, despite the fact that the circuit is working happily on the bench. I have to do much more work with PSpice to simulate my designs reliably.

The second reason is the lack of PSpice models for the discrete devices in my circuits. IC components are now very well characterized with PSpice models, and are available from the manufacturers. However, very few discrete devices have such models.

PSpice has a model generator, with which you can generate your own. But first you need reliable data for the devices. Unfortunately, discrete device datasheets rarely have the necessary data available. Model generation then becomes an intuitive process, very much dependent on one's experience.

Concerning Mr. Campbell's selection of devices, I have not tried these in the buffer. The 2N5457/59/60/62 have too large a spread in I_{DSS} to be applicable without testing and matching. The devices used in the original circuit are made by Toshiba of Japan, not by European manufacturers.

Erno Borbely
82205 Gilching, Germany

PC TO THE RESCUE

Is anyone trying to use low-cost PC sound boards to measure driver or box characteristics? It seems you could either use the "recording" capabilities of the boards (with the AGC turned off) along with a test CD of various frequency tones *or* use the sound generation capability along with a low-cost sound meter. Perhaps you could use two sound boards—one to generate sounds and the other to record them. Maybe you could use the different methods for different measurements, for example, Thiele/Small parameters, box losses, distortion, SPL versus frequency, and so on.

Of course, the better the sound board, the better the resolution of your measurements. Today's sound boards run the gamut from $40 to $1,000. How can you quantitize the precision of your measurements if you know the boards' parameters? I don't mind doing some programming, but I suspect that one of your editors could help me avoid some basic pitfalls along this path.

Lee Meador
Arlington, TX 76007

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Contributing Editor Richard Campbell responds:

Your question is certainly perceptive, since it seems obvious that multimedia-type record/play PC boards can be used for general data acquisition.

I am pleased to report that the first that I have seen of this breed of software has crossed my desk and is also currently advertised in SB *(see* SB *2/94, p. 67). It is called "PC Audio Lab" by Microacoustics Audio Software Products (2553 Carpenter St., Thousand Oaks, CA 91362, 818-719-2566).*

I am currently reviewing this product, so you'll get a chance to read about it in a future issue of SB. *I can say at this time that it works very well using the SoundBlaster 16 card (the Media Vision cards can also be used) and that it has many of the functions you mention plus many more. The software talks directly to the card without the need to load specific drivers, which the card might need for other applications.*

CAPACITOR GLUT?

I read with great interest Mark Seymour's article in *SB* 1/94 ("An Evolving Magnepan MG-1," p. 44) about adding capacitance to amplifier power supplies. I also obtained a copy of the Aug. 1980 *Audio* magazine article by Jung about this subject and now I have purchased the necessary parts to expand the capacitance of my Hafler DH120 amp from 6,600µF/channel with an additional 34,000µF/channel. Both Jung's and Seymour's articles convinced me that additional capacitance is one of the best sonic improvements to be made to any system.

I dug out the 4/91 issue of *Audio Amateur* and read the letter on p. 44 by Alan Douglas, "Megafarad Power Supplies." "Uh oh," says Douglas, additional capacitance might burn out the transformer/rectifier.

Jung and Seymour say nothing about this possibility. Although Seymour says his additional capacitance has operated well for ten years, I find myself wondering what might happen to my secondary listening system (the 120 powers a VMPS original sub and a pair of B&W DM 14s) if I go through with the extra capacitance.

Whom should I believe? Is the transformer/rectifier problem the reason that the high-end cost-no-object amp manufacturers don't offer the option of expanded capacity power supplies?

William H. Wallace
Stockton, NJ 08559

Mark Seymour responds:

You are right to be cautious about the modifications Mr. Jung and I have written about. Both warn readers of the dangers of adding capacitance without surge protection. Specifically, "Without such protection, the initial current surge will blow fuses, burn up the rectifier bridge, and degrade the transformer."

During operation, the current demand on an amplifier's power supply with an extended capacitor bank is a bit higher than without capacitance augmentation. Usually this can be handled within the rated current capability of the stock transformer and rectifier. It is wise, however, to measure quiescent current demand with and without extra capacitance in the circuit to be sure.

The bigger danger comes when the amp is first switched on. The discharged capacitors appear as a dead short to the rectifier bridge/transformer combination. The instantaneous current demand (the "On" current spike) is huge.

Temporary insertion of a 100W wirewound ceramic 10Ω resistor in series with the transformer limits the current surge to about 10A. This is within the current handling capability of the power transformer that came in my stock Hafler amp kit.

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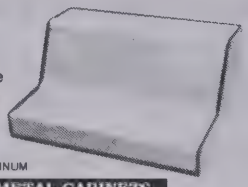

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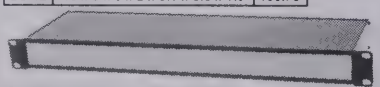

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Reader Service #23

After 100 seconds, when the capacitor bank is charged, this resistor is bypassed (effectively removed from the circuit) by means of a time-delay relay. If you have a smaller supply in your amp you may wish to use up to 20Ω resistor, which will further damp the initial current surge but will slow the charge time of the capacitor bank.

You wonder about use of augmented power supplies in "cost-no-object" amps. First, except in the private collection of audio designers who have a target market of one, I have never seen (heard) a true cost-no-object amp.

Second, even if people do appreciate the value ($/sonic enhancement), two other considerations rapidly become controlling: size and weight. The capacitor bank described in my SB *1/94 article occupies twice the volume of my power amp. If the devices were to be integrated, the new super-amp would be three times as large and weigh over 100 pounds.*

Finally, design of commercial amps requires a balance of many design alternatives to yield the best sound for the dollar. While extension of the capacitor bank improves the sound, other design options might yield an even better product.

Large electrolytic capacitors are expensive. The ones I used came from an electronic parts surplus house at about 10% of wholesale cost. My capacitor bank, if assembled from new parts, would cost over $1,000. If you had the extra $1,000 you might choose a different amp to begin with.

CIRCUIT SIMPLICITY

Silva's article, "Low-Frequency AC-to-DC Converter" in *SB* 1/94, presents a useful idea. However, his circuit implementation seems unnecessarily complex.

I have enclosed a circuit schematic (*Fig. 1*) that has the functionality of *Fig. A* in his article, but uses only about one third the parts. All specified components are available from Digikey (701 Brooks Ave. S., Thief River Falls, MN 56701, 800-344-4539). A step-by-step look at my circuit will explain the simplifications:

Input stage: Silva uses a times-two amp with an amplitude control. All sinewave and function generators I have seen or used have more than enough voltage range, amplitude controls, and low-impedance (typically 50Ω) outputs. Therefore, I have simply deleted this stage.

However, I have used AC coupling, because many sinewave generators don't have a precise zero offset output, or have offsets that can be easily misadjusted. For the safety of the circuit and the speaker under test, it is important not to allow DC in the speaker. Note that C1, although a large value, never sees much voltage, and only serves to block DC; therefore, it can be a cheap, low-voltage unit.

Voltage-to-current converter: Silva uses the classic Howland current pump, which is useful when the load (the speaker) must be grounded. Since no such requirement exists, a much simpler circuit is possible. Silva uses discrete components to provide the current needed to drive a speaker. I have used an inexpensive, readily available, power op amp, which eliminates more parts, and will drive the speaker at a minimum of ¼amperes, which is more than enough.

Current monitor: Silva uses a differential amp for this purpose. Unless his voltage-to-current converter is misbehaving, he should just be able to monitor the voltage at its input. For the converter I used, the question is moot because the current-monitoring resistor connects to ground, and, therefore, current can be measured by monitoring its voltage directly—no circuitry required.

Voltage monitor: Both circuits require a differential amp. I have left out the protection diodes (which you can add, if needed) because the possibility of zapping the op amp through the 100k resistors seems low. Silva neglects to mention (an oversight?) that the resistors in this stage must match. Specifically, the ratios R4/R3 and R6/R5 should be equal to within 1° for best accuracy.

Precision rectifier: Silva uses a full-wave circuit, but the average value of the positive and negative portions of an AC waveform that has no DC offset are by definition equal. Therefore, a half-wave rectifier is sufficient if DC offset can be set suitably low. Fortunately, that is easy.

For highest accuracy, U1's offset should be trimmed to less than 0.001V DC by installing whatever value of R_{OFF} is necessary to accomplish the task, to either the plus or minus supply as required. Using a low-offset op amp for U2 keeps offset of the other stages low.

I have also chosen not to install an amplitude control before this stage because of the availability of cheap DMMs that may not read low-frequency AC well, but can provide sufficiently high resolution at DC. If you wish an amplitude control simply install a pot in series with R7, or install whatever value of R7 you like.

In the circuit I used, the filtering is combined with the rectifier for further parts savings. I used a 100k resistor instead of a 10k unit so that the filter caps could be ten times smaller. For best accuracy and lower leakage, they should be tantalum rather than aluminum electrolytics.

Supply bypassing: Silva shows no bypass

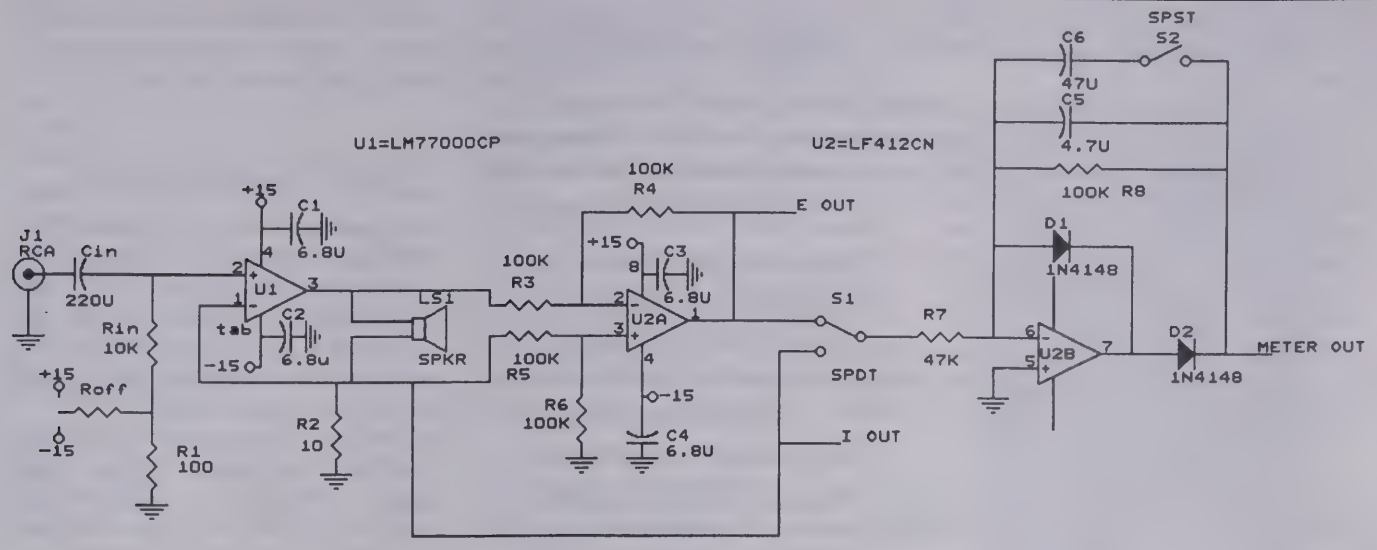


FIGURE 1: Suggested "Low-Frequency AC-to-DC Converter" circuit revision.

capacitors in his circuit. Unless the filter caps of your power supply are very close to your op amps, they must be used; otherwise, high-frequency oscillations will occur. It's good practice to include them, especially because of the low-impedance load that is being driven.

Note: A precision rectifier and filter determines the average, not the RMS, value of any AC input. Therefore, this circuit is not truly an AC voltmeter, won't give accurate absolute readings, and won't give accurate comparative readings on dynamic waveforms. It should only be used to compare static waveforms.

Roy Mallory
Bedford, MA 01730

Homero Sette Silva responds:

Thank you very much for your interest in my article in SB *1/94.*

First, I developed this dedicated circuit to measure T/S parameters of loudspeakers, and not as a general-purpose measurement circuit.

Second, many of those interested in loudspeakers aren't electronics experts, but they need a precise and foolproof tool for their measurements.

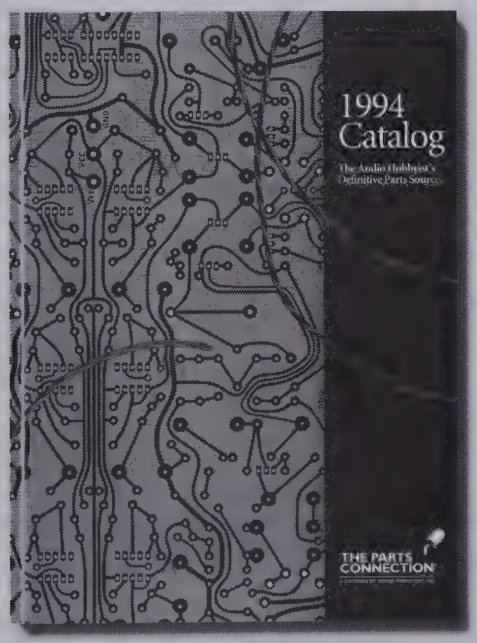
Input stage: The reason for the 50k pot and the times-two op amp at the input stage was the simple (but stable) oscillator I used, with a fixed and small amplitude output. About the capacitor coupling suggested, I avoided this due to increasing errors and great settling time, in the presence of very low frequencies.

Current-to-voltage conversion: Using a common ground to all measuring instruments (oscilloscope vertical and horizontal inputs, voltmeter, and so forth) and the device under test avoided a lot of trouble arising from wrong patching.

I'm aware about power op amps, but I don't have access to them here in Brazil. Unfortunately, Digi-key is far away.

Current monitor: The reason for a current monitor is very simple: to prevent disturbing the reference voltage drop in the sensing resistor, with low-impedance instruments or short circuits.

Voltage monitor: The lack of reference to



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Reader Service #26

this fact was not "an oversight," and I found it a necessity with 10% components. The same applies to the current source.

Precision rectifier: A half wave rectifier is not the best choice in the presence of very low frequencies, since the large capacitors needed to filter the AC components will result in large settling times. This fact is very annoying, when measuring Q_{TS} for example, since we are always changing frequency (from f_S to f_1 and f_2).

The gain adjust provided in this block was intended to calibrate the meter or modify its readings, without changing the loudspeaker current.

About reducing the filter capacitor, I said "increasing the resistor value allows you to reduce capacitor size proportionally, but temperature effect in the diode reverse current will be of concern." With aluminum capacitors a small value resistor is wise; using tantalum is a better choice.

Supply bypass: A very good practice. Thank you for reminding me.

Note: Throughout my text you'll find references warning that the circuit discussed was not a true RMS indicator, as in "you now have an AC voltmeter calibrated to RMS readings for distortionless sinusoidal input signals" or "Remember, this will hold only for sinusoids, since the averaging circuit is intrinsically a mean-value indicator."

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There are also a number of Elektor Electronics books geared to the electronics enthusiast – professional or amateur. These include data books and circuit books, which have proved highly popular. Two new books (published November 1993) are *305 Circuits* and *SMT Projects*. Books, printed-circuit boards, programmed EPROMS and diskettes are available from

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DINOSAUR POWER

Some years ago I purchased a pair of Z-40 Janzen electrostatic speakers. Although technology may long have passed them by, I have an affectionate attachment to them, and had the woofers reconed last year. From time to time the internal power supply in one of them fails and begins to blow fuses.

In past years, Soundmates in Minneapolis would repair the crossover/power supply boards at no charge. Unfortunately, Soundmates seems to have vanished in the years since my last encounter with the fuse problem. Can anyone out there help me locate a schematic for the power supply or put me in touch with Soundmates, the successor to Soundmates (if any), or suggest how I can repair these remarkable dinosaurs?

Bruce M. Chassy
Champaign, IL 61821

DRIVING DANIELLE

In my article "The Danielle II" (*SB* 2/94, p. 24), I mentioned limitations on the use of 4K111 drivers in cabinets which permit reflection of the rear wave. This has led to reader concerns about K2 cones in general. In order to avoid confusion, I offer the following:

1. The 4K111 drivers are excellent, with transparency, dynamics, and acceleration that few drivers can match. However, they would profit from a dipole arrangement (for its open sound and lack of reflections) or a transmission line to avoid reflections through the cone. Many other small, high acceleration, thin-coned drivers, including the Polydax MDA-100, would also profit from similar treatment.

2. K2 is a unique sandwich material with a core composed of hollow glass microballs. More rigid and better damped than simple Kevlar® cones, this product from FOCAL can better define transients and resist cone breakup.

As Dickason puts it, speakers can have quite a different "timbre soup" one from another. Successful speakers have been built around cellulose, K2 sandwich, Kevlar, honeycomb, aluminum, polypropylene, polyglass, ceramic, and other cone materials. The fun of speaker building is that of making music happen, not coldly applying technical know-how.

Marc Bacon
Montreal, Quebec H2H 2C3

SOLVING EQUATIONS

I am going crazy trying to figure out the mathematical sequence in "Stalking f_3" by Manning Redhill (*SB* 2/93, p. 24). Just above equation (1) the author indicates that the response of a fourth-order high-pass system is given by that equation.

First question: Where do you find that out? Are there some useful books to explain the responses of various systems? I make my living writing computer programs and have a degree in mathematics so they can be fairly technical.

Second, in that equation the response is given as a function of x. Does this mean that I could calculate SPL in decibels as a function of frequency (Hz) using this equation? Does the response correspond to the output of the vented box and driver? And, further, is this given as a function of frequency?

Third, equation (2) shows x as a function of f and calls x the "normalized frequency." What does it mean to "normalize" the frequency? Would my calculation (above) of SPL use f as the independent variable rather than x? Or would it just change the units along the bottom of the graph?

Fourth, and most important, what is the little d in the equations for the α_1 shown between equations (2) and (3)? It is also used later in the discussion, but I can find no definition for it.

Lee Meador
Arlington, TX 76007

Manning Redhill responds:

1. When the magnitude of the transfer function of a high-pass filter is given (this is what equation [1] is in my "Stalking f_3" article), then it is sufficient to examine how the function behaves when the independent variable x becomes very small. The denominator goes to 1 so it can be ignored, but the numerator goes to x^4, which labels the filter as fourth-order. In general, if the order is n, the amplitude response drops 6n decibels per octave.

2. and 3. The normalized frequency x is used only as a convenience. Whenever it appears in equation (1), substitute equation (2) for it. Now you have the amplitude response R(f) as a function of the ordinary frequency f in hertz. In a computer program, (a) select a frequency f in hertz, (b) use it to calculate x in equation (2), and (c) substitute x into equation (1) for the frequency response R(f) of the frequency you selected. Then (d) calculate $20Log_{10}[R(f)]$ to get the amplitude response in decibels. Now go back to (a).

4. In those equations between equations (2) and (3), the symbol d stands for denomi-

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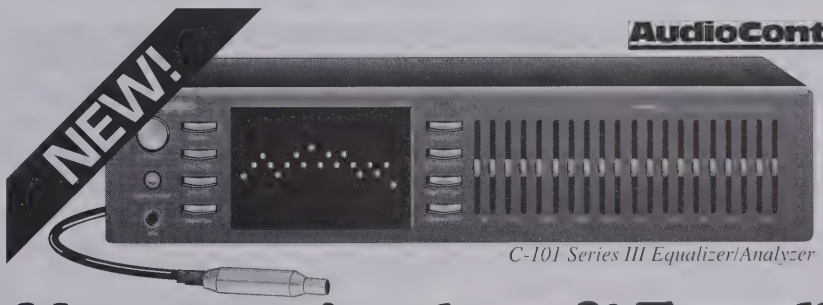

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nator—again, a notational convenience since the grouping of the symbols $Q_{LQTSh}^{1/2}$ appears in the denominator of the definition of all the α's.

A very useful, friendly book is Analog Filter Design *by M.E. van Falkenburg, published by Holt, Rinehart, and Winston (1982). It assumes you know your way around the complex plane.*

DO YOUR RESEARCH

I would like to respond to the editorial "Do We Need Postal Police?" in *SB* 2/94. What is an editorial like this doing in a loudspeaker journal? Please stick to the subject matter that we (the subscribers who pay for this) are entitled to. If you want a soapbox, write to the *Washington Post* or *Newsweek*.

Please do not mislead your readers with misinformation like "a government-subsidized agency." The USPS receives no tax subsidies; all revenues come from postage rates, which must rise from time to time like everything else due to inflation. Deregulating or privatizing the USPS is not the fix all people would like it to be. (Look at what happened to the trucking, airlines, and telephone industries, which are not now cheaper or in better shape.) Corporate greed can be as bad as bureaucracy.

I agree that the USPS is very poorly managed, and from the inside craft position (letter carrier) I hold, I can say this from experience (the editor cannot). Some say that we craft workers get paid too much. After seven years, I make only $33,000/yr., although I work very hard (it is a very difficult job that may seem easy on the outside, but until you do it, don't say anything).

One of the other purposes for federal protection of postal services is to provide universal service to everyone everywhere. UPS has already gone to a two-tier rate structure (residential delivery costs more than commercial delivery). It's easy to imagine a privatized post office providing better and cheaper service in large profitable urban areas and more expensive, but worse, service in smaller or rural markets (bad for Aunt Flo living in the middle of Iowa). Changing the current postal structure would also eliminate free forwarding and free return of first-class mail, and free forwarding of second-class publications like *SB*.

Also, a FAX is not a letter; it is a data transmission falling under FCC laws, not postal regulations. The USPS is one of the cheapest and most efficient postal services in the world. In Germany it costs 67.2 cents for a first-class letter; Italy, 62.3 cents; Norway, 54.8 cents; Austria, 47.9 cents; Japan, 46.6

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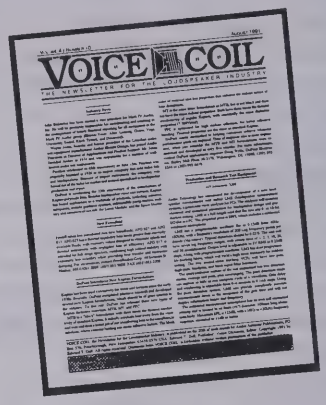

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cents; Great Britain, 42.5 cents; and Canada, 34.5 cents. In some of these countries, which may only be as big as some of our states, it takes longer for delivery.

The problem with USPS is not whether it needs a police service—it does, to take care of mail fraud, bombs, illegal drug traffic, dishonest employees, and so forth. The problem is an overbloated, extremely top-heavy management structure that has four craft workers for every manager/administrator position that does not handle or move mail.

Robert J. Suryan
Seattle, WA 98155

The health and existence of these magazines depends to a large extent on USPS performance, both in handling promotional mail and second-class rules. It is very much in the interest of all readers that their postal service function well. Mr. Suryan has not provided any information of which I have not been fully aware. The fiction that the USPS has no government subsidies is belied locally when a supervisor drives to his job daily in an automobile with a "US Government" tag on it. This editor's "research" includes over 30 years of dealing with the postal service.

The services which reader Suryan mentions are nice on paper, but they do not work well and sometimes not at all. Forwarding and address correction has, in some cases, taken six months. Missorting mail is a daily occurrence, ranging from one to 50 pieces each day. Our periodicals are routinely destroyed without notice and without compensation. The postal service is the only delivery service which does not have to be fiscally responsible for destruction of property, unless the sender pays for insurance. The latter is difficult to collect and takes a lot of time and paperwork.

The people who serve the USPS are to be commended, as always, for coping with an institution that must be modified for their sakes, as well as the public's. The postal system's problems are a matter of record with almost every periodical in this country, including The Reader's Digest. *Significant change in the system will benefit postal employees most of all, and they deserve better, as do those who depend upon the system.—Ed.*

CAVEAT EMPTOR

I was quite excited when I read Peter Manchev's letter in *SB Mailbox*, *SB* 5/92, p. 64, telling of his experiences with an 18″ Orevox Essence WC-18125 woofer. The specs looked interesting, and, when plugged into my MacSpeakerz program, yielded very promising performance graphs for a moderate-sized sealed box.

I contacted the supplier, and bought a WC-18125. Unfortunately, the actual speaker falls far short of the advertised specifications and, while it does produce a lot of subjectively impressive "boom," as Mr. Manchev enthuses, it does not go very low. The measured f_S is 34Hz, instead of the advertised 15Hz, so that the $f_B(4.3ft^3)$ turns out to be around 48Hz, instead of the computed and anticipated value of 28Hz. So much for my dream of –3dB at 25Hz in a small box.

Moreover, since I wrote to the supplier (Goldwood Products, of Northridge, CA) about my disappointment, I have received no reply. I think *SB* readers should be warned not to buy a "pig in a poke," but when pur-

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PREVIEW
Glass Audio

Issue 2, 1994

- Constructing Cascodes
- New Dynaco PAS Upgrade
- Lowe's Servo Bias System Revisited
- Output Transformers, Part I

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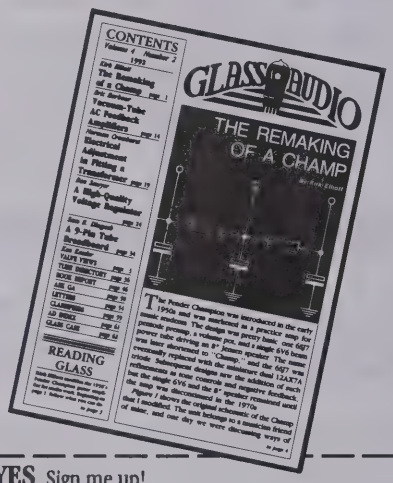

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chasing speakers from unknown suppliers (as opposed to the established and reputable dealers such as Madisound, Audio Concepts, and others), to do so on a "satisfaction guaranteed" contingency basis.

John F. Bundy
Etna, NH 03750

[*We have also queried Goldwood Products and have received no response.— Ed.*]

Crossover Response

Continued from page 11

response is not perfect, as illustrated by the ripples in the step response. For the greater cutoff rate obtained compared to, say, a first-order filter, these ripples might be acceptable. Other filter types, such as Bessel and Chebyshev were similarly analyzed, but in modifying the design for a more nearly flat SPL magnitude, the step response ended up little or no better than that of the Butterworth version. I did not consider fourth or higher order filters, because of the large number of components required.

In summary we could have concluded that the best options are first- and third-order Butterworth crossover filters. However, this study only serves to illustrate certain principles as far as time response is concerned. In practice drivers have a major effect on the crossover response, making it necessary to design with the aid of speaker design software.

SLM Modification

Continued from page 13

right and just above the battery compartment. The 47µF capacitor is soldered between the second (from the left) and third of these, with the case of the electrolytic connected to the latter (*Photo 1*). This should give a flat response to 20Hz. As a 0dBm reading from the Mitey Mike corresponds to an SPL of 120dB, the modified meter is compatible. Checked against a standard, the modified level meter measured at the line input within 0.5dB 20Hz–20kHz and within 1db –60 to 0dBm (not bad for an expenditure of less than $40).

I can't guarantee that the current model is exactly the same as the two that I have modified—manufacturers do make changes and improvements. I wouldn't anticipate major differences, however. Bear in mind, too, that the modifications will void the 90 day warranty by Radio Shack.

Microphone Correction

Continued from page 21

out in the display using about ⅓ or ⅙ octave. (Smoothing is under the Display Format menu.) *Figure 4* shows an example.

7. Be sure the displayed frequency range covers the range of correction interest (about 1–20kHz). The displayed range is all that will be written out in an ASCII file of frequency response. Save the frequency response to disk, in ASCII mode.

8. To use the file as a microphone correction file, you must change its name. Its "dot suffix" must be ".DAT," not ".FRD" as IMP/M will assign. You can do this via the DOS "REN" command. The file *must* be located in the same subdirectory from which you run IMP/M. Move it if necessary.

9. Look at the file of your original reference microphone correction data. In its top lines, there may be a reference to some number as so many "mV/Pa." This is the reference sensitivity which you may have modified if you use a microphone attenuator. You should copy this entire line, including quotes, with a text editor to the top of the correction file you just made for your new microphone. That way, the microphone can be used for absolute SPL-based measurements also. (See IMP/M help for some details.)

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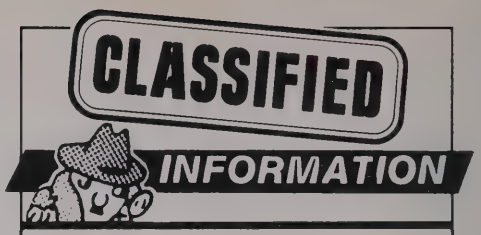

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ALL PERSONS INTERESTED IN STARTING AN AUDIO/SPEAKER BUILDER CLUB IN THE SOUTHWEST MISSOURI-NORTHWEST ARKANSAS AREA, please send your name, address, phone number, and something about yourself to: Greg McKinney, 900 S. Roanoke, Apt. 2, Springfield, MO 65806.

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ARIZONA AUDIOPHILE SOCIETY located in metropolitan Phoenix is a growing and active club in the pursuit and reproduction of recorded music. New members are welcome. Meetings are last Tuesday of each month. Receive monthly newsletter and biannual journal. Club discounts with local high-end audio dealers. Send inquiry to Arizona Audiophile Society, PO Box 13058, Scottsdale, AZ 85267, or call Bob Williams, (602) 944-5929.

THE ATLANTA AUDIO SOCIETY is dedicated to furnishing pleasure and education for people with a common interest in fine music and audio equipment. Monthly meetings often feature guest speakers from the audio manufacturing and recording industry. Members receive a monthly newsletter. Call Chuck Bruce, (404) 876-5659, or Eddie Carter, (404) 847-9296, or write A.A.S., 4266 Roswell Rd. N.E., K-4, Atlanta, GA 30342-3738.

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THE BOSTON AUDIO SOCIETY, the nation's oldest (founded 1972), seeks new members. Dues include the monthly meeting notice and our newsletter, the *BAS Speaker* (6/year). Recent issues cover Carver, a/d/s; the founder of Tech Hi-Fi, Photo CD. Plus visits from famous speaker designers, listening tests, measurement clinics, research investigations, and more. Back volumes available. Membership includes engineers, journalists, consultants, and music-loving audiophiles like yourself. For information write to PO Box 211, Boston, MA 02126-0002, USA.

THE CATSKILL AND ADIRONDACK AUDIO SOCIETY invites you to our informal meeting. Join our friendly group of audio enthusiasts as we discuss life, the universe and everything! Toobers, Tranzzeestors, vinyl canyons, or digital dots. No matter what your level of interest, experience, or preferences, you are welcome. Contact CAAS at (518) 756-9894 (leave message), or write CAAS, PO Box 144, Hannacroix, NY 12087.

THE COLORADO AUDIO SOCIETY (CAS) is a group of audio enthusiasts dedicated to the pursuit of music and audiophile arts in the Rocky Mountain region. We offer a comprehensive annual journal, five newsletters, plus participation in meetings and lectures. For more information, send SASE to CAS, 1941 S. Grant St., Denver, CO 80210, (303) 733-1613.

CONNECTICUT AUDIO SOCIETY is an active and growing club with activities covering many facets of audio–including construction, subjective testing, and tours of local manufacturers. New members are always welcome. For a copy of our current newsletter and an invitation to our next meeting, write to: Richard Thompson, 129 Newgate Rd., E. Granby, CT 06026, (203) 653-7873.

ESL BUILDERS GROUP is a new address for people who have built or want to build electrostatic loudspeakers and associated (tube) drivers, or are just interested? An answer is ensured, if you include some kind of compensation for postage and handling. Write to: Gunter Roehricht, Bühlerstr. 21, 7030 Böblingen, Germany.

THE HI-FI CLUB of Cape Town in South Africa sends a monthly newsletter to its members and world-wide subscribers. To receive an evaluation copy of our current newsletter, write to PO Box 18262, Wynberg 7824, South Africa. We'll be very pleased to hear from you.

HI-FI COLLECTOR/HOBBYIST seeks "living letters"/audio penpals from other states to correspond via reel-to-reel tape. Noncommercial strictly; make up short monologues on subjects from vintage technology, with regional FM excerpts for background or equipment samples, from personal tales of yard-sale scavenging success, repair/restoration tactics and strategies, favorite service centers, general ways to handle the burgeoning obsession with arcane hi-fi gear. All correspondence on 3″, 5″, 7″ reels (¼″ tape) will be cheerfully answered and tapes returned via parcel post. James Addison, 171 Hartford Rd., Apt. #7, New Britain, CT 06053.

THE PRAIRIE STATE AUDIO CONSTRUCTION SOCIETY (PSACS) meets every other month. Meetings feature audio construction, design, and analyses, blind listening tests, equipment clinics, auto sound, lectures from manufacturers and reviewers. PSACS, PO Box 482, Cary, IL, 60013, or call Tom, (708) 248-3377 days, (708) 516-0170 eves.

MEMPHIS AREA AUDIO SOCIETY being formed. Serious audiophiles contact J.J. McBride, 8182 Wind Valley Cove, Memphis, TN 38125, (901) 756-6831.

LONDON LIVE D.I.Y. HI-FI CIRCLE meets quarterly in London, England. Our overall agenda is a broad one, having anything to do with any aspect of audio design and construction. We welcome everyone, from novice to expert. For information contact Brian Stenning, 081-748-7489.

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MONTREAL SPEAKER BUILDER CLUB. Meets when it can, BYOB, discussions range from speaker design and testing to equipment modification. All welcome. Contact Andrew McCree, 4701 Jeanne Mance, Montreal, PQ, H2V 4J5 Canada, (514) 281-7954.

NEW JERSEY AUDIO SOCIETY meets monthly. Emphasis is on construction and modification of electronics and speakers. Dues include monthly newsletter with high-end news, construction articles, analysis of commercial circuits, etc. Meetings are devoted to listening to records and CDs, comparing and A-B-ing equipment. New members welcome. Contact Frank J. Alles, (908) 424-0463, 209 Second St., Middlesex, NJ 08846; or Bob Young, (908) 381-6269; or Bob Clark, (908) 647-0194.

THE INLAND EMPIRE AUDIO SOCIETY—soon to become **THE SOUTHERN CALIFORNIA AUDIO SOCIETY (SCAS)**—is now inviting audiophiles from all areas of Southern California and abroad to join our serious pursuit for that elusive sonic truth through our meetings and the **IEAS** official speaker, *The Reference* newsletter. For information write or call Frank Manrique, President, 1219 Fulbright Ave., Redlands, CA 92373, (714) 793-9209.

NORTHERN VIRGINIA BUSINESS CLUB. I am interested in turning my speaker-building interests into a profitable business. So I'm organizing this business club to attract other speaker builder entrepreneurs with the same interests. Call Frank Troy, (703) 912-8226, M-F, 7:30 a.m. to 4:30 p.m.

PACIFICNORTHWEST AUDIO SOCIETY (PAS) consists of 60 audio enthusiasts meeting monthly, second Wednesdays, 7:30 to 9:30 p.m., 4545 Island Crest Way, Mercer Island, WA. Write Box 435, Mercer Island, WA 98040 or call Bob McDonald, (206) 232-8130 or Nick Daniggelis, (206) 323-6196.

TUBE AUDIO ENTHUSIASTS. Northern California club meets every other month. For next meeting announcement send #10 SASE to Tim Eding, PO Box 611662, San Jose, CA 95161.

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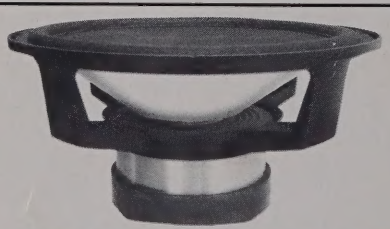
VINTAGE AUDIO LISTENERS AND VALVE ENTHUSIASTS (VALVE) meets the first Sunday of every month to swap vintage audio gear, audition rare and collectible equipment, and evaluate modifications and scratchbuilt projects. Dues provide a monthly newsletter with current reviews of vintage components and modification information; vintage service data; and access to an active network of serious collectors. For information, call (206) 697-1936 or write to 1127 NW Brite Star Ln., Poulsbo, WA 98370.



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WASHINGTON AREA AUDIO SOCIETY meetings are held every two weeks, on Fridays, from 19:00 hours to 21:30 hours at the Charles Barrett Elementary School in Alexandria, VA. Prospective members are welcome but must register in advance in order to be admitted to the meetings. No exceptions please. Call Horace Vignale, (703) 578-4929.

WASATCH AUDIO, located in Salt Lake City. Our club is interested in construction, modifications, design, and listening to music. We are looking for members and ideas for our new club. Contact Edward Aho, (801) 364-4204.

WEST VALLEY AUDIO SOCIETY. We are starting a group interested in all aspects of high performance audio. West San Fernando Valley, CA. Contact Barry, (818) 225-1341.

THE WESTERN NEW YORK AUDIO SOCIETY is an active, long-established club located in the Buffalo area. We issue a newsletter and hold meetings the first Tuesday of every month. Our meetings attract many prominent manufacturers of audio-related equipment. We are involved in all facets of audio from building/modifying to exposure to the newest high-end gear, and the chance to hear more types of music. For information, write to WNY Audio Society, PO Box 312, N. Tonawanda, NY 14120.

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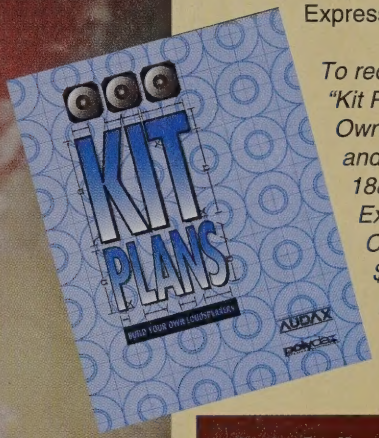

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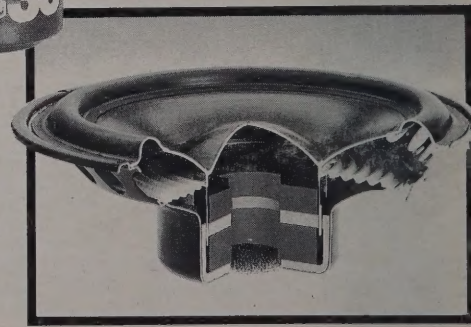

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